

# **Development of MEMS Piezoelectric Energy Harvesters**

by

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## Abstract

The research of powering devices in a microwatt range has been activated and developed by the emergence of low-power Very Large Scale Integration (VLSI) technology in the past few years. The powering devices require a size that is compatible with the application, sufficient power, and extended lifetime using permanent and ubiquitous energy sources. The piezoelectric energy harvester using vibration sources is attractive due to its high conversion efficiency, simple design for miniaturizing, and lack of external voltage source. While bulk piezoelectric energy harvesters produce enough power for a few tens of mW, the insufficient power is still a major issue during miniaturizing into micro size.

The piezoelectric energy harvester was fabricated by micro-electro-mechanical systems (MEMS) and developed to enhance its output power. It was designed to be resonated at the frequency range of ambient vibration source (50~300 Hz) and convert the mechanical stress to electricity by piezoelectric thin film. The cantilever structure was chosen in this study due to its large strain, and a big proof mass at the end of tip was integrated for the same reason. This study focuses on three specific issues related to the robust fabrication process, including the integration of piezoelectric thin film, structure design for high power density, and the reliability of device. The Lead Zirconate Titanate (PZT) thin films were prepared by a sol-gel process and were used to fabricate energy harvesters by an optimized MEMS process. The properties of PZT thin film

were studied considering the substrate effect, heat treatment, and thickness effects. The fabricated energy harvester produced 769 mVpk-pk, and 6.72  $\mu$ W with the optimal resistive load of 11 k $\Omega$  at 127 Hz of resonant frequency. The device had dimensions of about 4 mm(L) x 2 mm(w) x 0.021 mm(H), and the Si proof mass had dimensions of 3 mm(L) x 2 mm(W) x 0.5 mm(H). Beyond this result, the technical platform for the robust fabrication process was established on a Deep Reactive Ion Etcher (DRIE). The plasma etching using DRIE was optimized to prevent damage of the PZT film and to obtain uniform and precise dimension control.

The trapezoidal shape of the cantilever was demonstrated to enhance the power density by stress distribution on the PZT film. The geometry change in cantilever shape distributed the strain on piezoelectric film and improved the output power  $\sim$ 40% higher than that of the rectangular shape due to nonlinear piezoelectric properties. The multi-beam arrays were designed to obtain a multiplied electric power effect as if as number of cantilevers was used. The multi-beam arrayed design requires the uniform machining to match the unified resonant frequency of each cantilever structure. Based on the optimized fabrication process, the cantilever array that consists of four cantilevers generated 18.39  $\mu$ A and 1.352  $\mu$ W with 4 k $\Omega$  of optimized resistive load in parallel connection under 1 G of acceleration force. The result was exactly four times higher power and current than that of individual cantilever.

Finally, reliability tests were performed for the piezoelectric MEMS energy harvester considering the number of cyclic loads and temperature, and the degradation of PZT during fabrication was also investigated.

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## Table of Contents

Abstract.....	ii
Acknowledgments.....	iv
List of Figures.....	ix
List of Tables.....	xiv
1. Introduction	
1.1 Motivation .....	1
1.2 Energy Harvesting .....	2
1.3 Competing Power Sources .....	3
1.4 Converting Vibration to Electricity .....	6
1.5 Research Objectives .....	8
2. General Background	
2.1 Piezoelectricity and PZT films .....	10
2.1.1 Piezoelectricity .....	10
2.1.2 Piezoelectric Materials for Energy Harvesters.....	14
2.1.3 Lead Zirconate Titanate (PZT) .....	16
2.1.4 Poling & Depoling .....	18

2.1.5 Piezoelectric Mode (d33 & d31) .....	21
2.2 Electrical Power Management .....	23
2.3 Piezoelectric MEMS Energy Harvester .....	26
2.3.1 Single Piezoelectric MEMS Energy Harvester .....	26
2.3.2 Multi-beam Arrayed Piezoelectric MEMS Energy Harvester .....	28
3. Integrating Piezoelectric Thin Film on Silicon Wafer	
3.1 Film Characterization .....	31
3.2 Thin Film Deposition .....	32
3.3 Chemical Solution Deposition(CSD) of PZT film.....	33
3.3.1 Analysis of PZT Thin Film on Pt Substrate.....	36
3.3.2 Adhesion Layers Effects on The Electrical Properties PZT.....	38
3.3.3 Heat Treatment Effects.....	42
3.3.4 PZT Thickness Effects .....	45
4. Design and Fabrication of MEMS Energy Harvester	
4.1 Design.....	49
4.1.1 Resonant Frequency .....	50
4.1.2 Strain on PZT Film & Safety Factor .....	53
4.1.3 Design Consideration for Low Resonant Frequency.....	54
4.1.4 Analytical Output Power .....	60
4.2 MEMS Fabrication. ....	62
4.2.1 Lithography and Mask for DRIE .....	65
4.2.2 Silicon Bulk Machining (ASE).....	69
4.2.3 Surface Micromachining (AOE).....	76

4.2.4 Designing Masks Based on Optimized pMEMS Process.....	80
5. Device characterization and Analysis	
5.1 Experimental Setup for Characterization.....	92
5.2 Performance of Single Cantilever .....	93
5.2.1 Device on Silicon Wafer .....	93
5.2.2 Device on SOI Wafer .....	101
5.3 Performance of Cantilever Array .....	104
5.3.1 Device on Si and SOI Wafer .....	104
5.3.2 High Power Device .....	105
5.3.2.1 Parallel Connection .....	107
5.3.2.2 Serial Connection .....	115
5.4 Uniform Stress Contribution .....	116
5.5 Comparison Between 31 Mode and 33 Mode.....	125
5.5.1 Poling Effect on 31 Mode and 33 Mode .....	128
5.5.2 Outputs on 31 Mode and 33 Mode.....	134
5.6 Reliability.....	136
5.6.1 Structure Failure by Fatigue.....	137
5.7 Degradation of PZT.....	140
5.7.1 Degradation During Fabrication .....	140
5.7.2 Reliability Under Cyclic Load .....	142
5.7.3 Reliability under High Temperature .....	143
6. Conclusions and Future Work	
6.1 Summary and Conclusions .....	148

6.2 Future Work .....	150
Bibliography.....	152

## List of Figures

1.1	Energy consumption of VLSI wireless sensors.....	2
1.2	The schematic architecture of a self-powered wireless sensor network.....	3
1.3	Comparison of the available power density and the lifetime among potential energy sources.....	4
1.4	Illustration of the electromechanical transducer systems: (a) electrostatic, (b) electromagnetic, and (c) piezoelectric.....	7
2.1	The piezoelectric effect and their variables.....	11
2.2	Perovskite structure.....	17
2.3	Phase diagram of PZT.....	18
2.4	The dipole moment during poling process .....	19
2.5	The shape change by the polarization as the crystallographic structures.....	20
2.6	Electrode configurations of $d_{31}$ and $d_{33}$ for cantilever beam.....	21
2.7	A schematic diagram of a standard energy harvesting system.....	23
2.8	The signals during a piezoelectric energy harvester system.....	24
2.9	A standard energy harvesting circuit .....	24
3.1	Typical sol-gel chemical solution deposition (CSD) process. ....	33
3.2	The schematic changes of coupling coefficient and piezoelectric constant.....	34

3.3	Process flow of PZT chemical deposition.....	35
3.4	XRD data of as deposited Pt/Ti (120/10nm) by sputtering.....	36
3.5	PZT/Pt film analysis by XRD .....	37
3.6	SEM image of PZT film .....	38
3.7	PE hysteresis loops of PZT films deposited on different thickness of adhesion layers.....	39
3.8	Substrate effect on piezoelectric properties. Adhesion layers deposited between SiO <sub>2</sub> and bottom Pt, and their effects described by: (a-b) Ti, Cr, and Ta have different effects on PZT.....	40
3.9	XRD and SEM analysis of PZT film on different adhesion layer: (a) XRD analysis, (b) Ti, (c) Cr, and (d) Ta adhesion layers .....	41
3.10	Angle view of PZT thin film on Cr and Ti adhesion layers: (a) Cr and (b) Ti .....	42
3.11	Electric properties of PZT films using different heat treatment method: (a) as temperature on hotplate (b) P <sub>r</sub> , (c) dielectric constant, and (d) PE loop as heat sources .....	44
3.12	XRD analysis of PZT film pyrolysed using (a) furnace and (b) hotplate .....	45
3.13	XRD analysis of PZT films with different thickness.....	46
3.14	PE hysteresis loop and dielectric constant as frequency for the PZT films with different thickness.....	47
4.1	Illustration of common configurations of pMEMS structure.....	49
4.2	Schematic of cantilever energy harvester with proof mass and the notation of dimension.....	50
4.3	Resonant frequency vs. weight of proof mass in various lengths of cantilevers. The widths of the beam and mass are fixed as 2 mm.....	56
4.4	Resonant frequency vs. thickness of cantilever in various lengths of cantilevers whose width is fixed at 2 mm .....	57
4.5	Safety factor vs. thickness of Si at different lengths of cantilevers. The width of the beam was fixed at 2 mm and the weight of mass is constant .....	58
4.6	Average strain on PZT vs. thickness of Si at different lengths of cantilevers when the width of beam was fixed as 2 mm and the weight of mass is constant .....	59

4.7	Equivalent lumped spring mass system.....	60
4.8	Process flow of piezoelectric MEMS energy harvester using SOI wafer. ....	63
4.9	Edge of negative photoresist patterning. ....	67
4.10	Damages on edge of Pt/PZT film after removing Al metal mask. ....	68
4.11	The illustration of DRIE.....	71
4.12	Cyclic etching and passivation of ASE process.....	73
4.13	SEM image of side wall angle in silicon etching.....	74
4.14	Different etching rate than window size. The side of mass was not released, while the other area started to etch out.....	76
4.15	Cracks were created on PZT due to strong ion bombardment.....	77
4.16	SEM image of etched PZT film surface in cross section view.....	79
4.17	1 <sup>st</sup> generation photolithographic mask for prototype fabrication ....	82
4.18	2 <sup>nd</sup> generation photolithographic mask for prototype fabrication.....	84
4.19	3 <sup>rd</sup> generation photolithographic mask for prototype fabrication ....	86
4.20	Prototype cantilever array ....	87
4.21	Photolithographic mask for cantilever array prototype ....	88
4.22	4 <sup>th</sup> generation photolithographic mask for prototype fabrication ....	89
4.23	5 <sup>th</sup> generation photolithographic mask for prototype fabrication.....	91
5.1	Schematic of experimental setup for energy harvester.....	92
5.2	Cantilevers have different average thicknesses on Si under layer t (a) 100 $\mu\text{m}$ , (b) 20 $\mu\text{m}$ , and (c) 10 $\mu\text{m}$ (cantilever was bent due to internal stress).....	94
5.3	Geometric graph of cantilever's thickness: (a) Sampled cantilevers and their location, (b) Thickness at the joint on proof mass, (c) Thickness at the middle of cantilever, and (d) Thickness at the joint on anchor.....	96
5.4	The resonant frequency of cantilever fabricated on Si wafer.....	97
5.5	Resonant frequency is very sensitive to the thickness of the beam.....	98
5.6	(a) Output peak voltage (b) Power from single cantilever on Si.....	99

5.7	Concave etching profile after silicon etching.....	101
5.8	Comparison between resonant frequencies of estimated and measured values.....	102
5.9	Output peak voltage and power of device fabricated on SOI wafer.....	103
5.10	Resonance frequencies of cantilever array fabricated on (a) Si wafer, and (b) SOI wafer, and output voltage from device in sweep mode for (c) Si wafer, and (d) SOI wafer. The acceleration is 1 g.....	105
5.11	Prototypes of pMEMS cantilever arrays. ....	107
5.12	The resonant frequency of a cantilever array fabricated on an SOI wafer.....	107
5.13	(a) The output voltage vs. frequency, and (b) the phase shift of C1, C2, and C3 from the phase of their parallel connection (C1+C2+C3) .....	109
5.14	The signals from four cantilevers and their resonant frequency at 229.1 Hz; (a) Peak-to-peak voltage, and (b) output power with resistive load.....	110
5.15	The generated current at the optimal resistance load as number of cantilevers in parallel connection.....	113
5.16	AC output of four cantilevers and their parallel connection.....	114
5.17	Output peak voltage and power as resistance load from serial connected array.....	115
5.18	The generated current at the optimal resistance load as number of cantilevers in serial connection.....	116
5.19	The structure for uniform stress distribution.....	118
5.20	The configuration of electrode for geometric stress concentration: (a) rectangular, (b) trapezoidal. ....	119
5.21	The configuration of electrode and mesh elements on (a) rectangular and (b) trapezoidal.....	120
5.22	The output powers of different electrodes for (a) the rectangular and (b) the trapezoidal cantilever across the resistive load. Their maximum power was compared with the FEM modeling strains for (c) the rectangular and (d) the trapezoidal cantilever.....	122
5.23	The strain profiles from anchor to mass for the rectangular and the trapezoidal-shaped cantilevers. ....	123
5.24	The generated total powers from rectangular and trapezoidal. ....	124

5.25	The output maximum power vs. cantilever shape.....	125
5.26	Schematic illustration of 31 mode (a) and 33 mode (b) cantilevered piezoelectric energy harvesters.....	126
5.27	Schematic of IDE.....	127
5.28	Poling directions on IDE that show dead zone as grey color .....	128
5.29	Polarization hysteresis loop for the PZT film on PEH.....	130
5.30	Output power as the poling conditions.....	131
5.31	(a) P-E loop after DC biased poling at high temperature (b) the shift of coercive field as direction of poling.....	132
5.32	The performance of PEH as poling conditions: (a) output voltage and (b) output power.....	132
5.33	The characterization of the PZT film in d33 mode PEH: (a) hysteresis loop & (b) XRD analysis.....	133
5.34	(a)output voltage as different poling conditions & (b) output voltage at open-circuit as poling field.....	134
5.35	The comparison between 31 and 33 mode conversion based on cantilever displacement: (a) output from 31 mode, (b) output from 33 mode.....	135
5.36	SEM image of fractured surface of cantilever.....	139
5.37	The piezoelectric properties of PZT after plasma etching.....	141
5.38	Output voltage and resonant frequency as number of cyclic load.....	142
5.39	Output voltages vs. frequency with different temperature.....	144
5.40	Resonant frequency shift with temperature.....	145
5.41	The output voltage vs. temperature with open circuit.....	146

## List of Tables

1.1	Various potential power sources from ambient and comparison of meso size energy harvesting devices. The power density has unit in $\mu\text{W}/\text{cm}^3$ or $\text{cm}^2$ .....	5
1.2	The potential vibration sources and their characteristics.....	6
1.3	Comparison of three types of transducers.....	7
2.1	Relationship between tensor notation and matrix notation.....	13
2.2	Coupling coefficient values of typical piezoelectric materials.....	15
2.3	Comparison of piezoelectric coefficients of the representative materials in thin film scale.....	16
2.4	Comparison between $d_{33}$ and $d_{31}$ mode for energy harvester.....	22
2.5	The comparison of reported MEMS energy harvester.....	28
3.1	Comparison of deposition method for ferroelectric materials.....	32
3.2	$P_r$ , $\epsilon_r$ , and $d_{31}$ of PZT film deposited on different adhesion layers.....	40
4.1	The material constants.....	53
4.2	Specification and issues of individual layers in fabricating pMEMS energy harvesters .....	64
4.3	Parameters of lithography processes .....	69
4.4	The process variables and their effect on etching.....	71
4.5	ASE recipe .....	74
4.6	The reported recipes of PZT plasma etching.....	76

4.7	Investigation of process parameters of AOE system for etching PZT thin film .....	78
4.8	AOE recipe.....	79
4.9	Dimension and resonant frequency of single cantilever in 2 <sup>nd</sup> mask.....	85
4.10	Dimension and resonant frequency of cantilever array in 2 <sup>nd</sup> mask.....	85
4.11	Dimension of devices on 4 <sup>th</sup> mask.....	89
5.1	Characteristic of energy harvester on Si wafer .....	100
5.2	Characteristics of energy harvester fabricated on SOI wafer .....	103
5.3	Cantilever array and their power management for high power device.....	106
5.4	Resonant frequency, optimal resistance, generated current, and maximum output power from the array device with the number of cantilevers.....	112
5.5	Comparison between $d_{31}$ and $d_{33}$ mode.....	135
5.5	Reliability issued on Si-based PZT cantilever.....	137

# CHAPTER 1

## INTRODUCTION

### 1.1 Motivation

The recent development of Integrated Chip (IC) technology has enabled the appearance of miniaturized electronics with their advantages of integration, light weight, low cost, and high performance [1]. The wireless sensors are attractive for monitoring dispersed conditions of mechanics, structures, and environments by deploying large quantities of them various positions. While the wireless nodes have been shrunk to sub-centimeter sizes and have decreased their energy consumptions, the traditional chemical batteries have been restricted by their large size, manufacturing cost, depleted power sources, and expense of replacement.

The wireless sensor nodes perform four main functions that require electric power: transmitting, receiving, polling, and sleeping (standby) [2]. The maximum power consumptions of the wireless devices ranges from 500  $\mu$ W to 50 mW in transmit mode depending on the distance and data transfer rate in the current technology. The transmit time, however, is less than 0.1% of the total lifetime of the device, and the power consumption of other modes are a few hundred times less than that of the transmit mode [3]. The power trace is scaled in Figure 1.1.

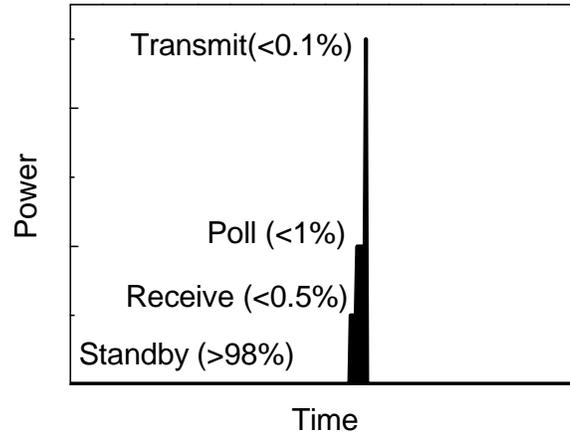


Figure 1.1 Energy consumption of VLSI wireless sensors

The research of sustained micro powering devices for wireless devices was activated and developed in the past decade to replace traditional chemical batteries. The promising energy harvester should be a compact power supply, and it should permanently provide sufficient electricity to the wireless sensor node, which consists of sensors, wireless communication components, and control electronics. The low powering device with the wireless sensor is expected to emerge for wider applications such as environmental monitoring, automation, diagnostic, medical, homeland security, military, and gaming, if it successfully provides enough power in a compatible size.

## 1.2 Energy harvesting

Human beings require energy for a better quality of life and have built up power plants using natural energy sources such as combusting materials, the sun, wind, or gravitation.

Attributing to the appearance of small electronics that consume very low power, people have tried to find new energy sources that are more effective in size, cost, and lifetime. In the past few decades, researchers have found some wasted energy sources or secondhand sources such as light, vibration, and heat. They are ubiquitous and it is possible to convert them into micro electric power, which is required in small electronics. Figure 1.2 shows a schematic architecture of a micro powering device for a low power device.

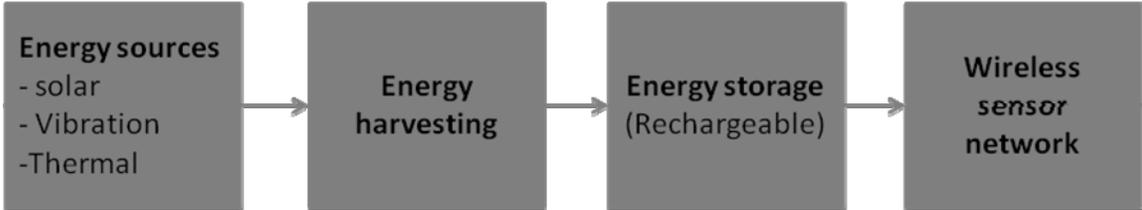


Figure 1.2 The schematic architecture of a self-powered wireless sensor network

To realize the micro energy harvester as an engineer, the target energy source and its efficiency should be considered within the size of the application. Since the energy harvester is required to generate sufficient power for electronics using ambient energy sources, energy density rather than energy efficiency should be considered to fit to the small electronics. The generated electric power through the energy harvester will be stored in micro batteries or a capacitor.

### 1.3 Competing power sources

Various ambient energy sources, such as solar, thermal, acoustic, and mechanical energies, have been investigated for micro-power generating devices. The potential energy

sources should have high power density with long lifetimes (wireless sensors have a 10-year lifetime in its roadmap) and low product cost in a subcentimeter size. Solar energy and mechanical vibration were considered, and their estimated power density and lifetime are compared with the conventional chemical batteries in Figure 1.3.

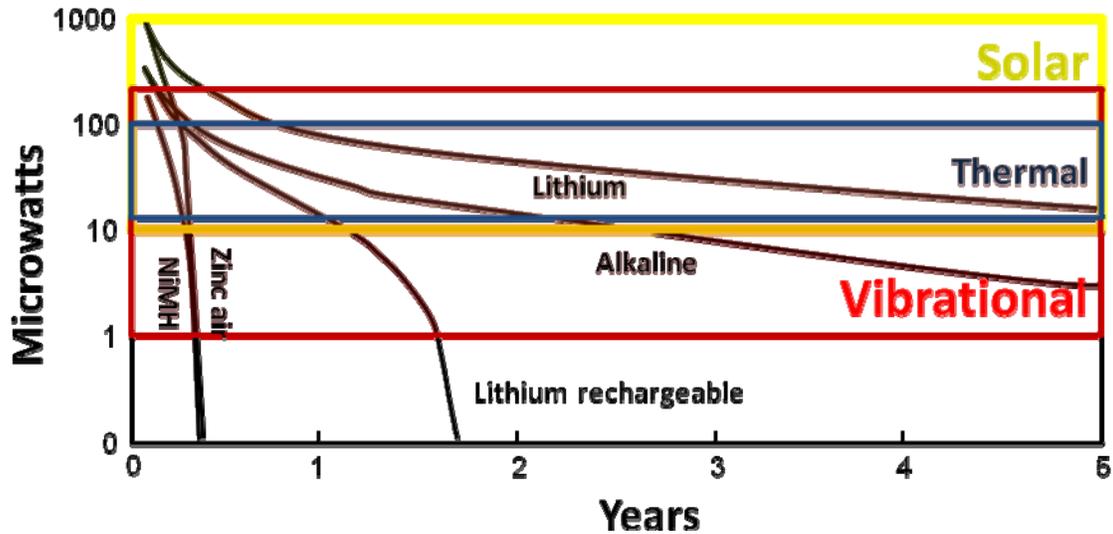


Figure 1.3 Comparison of the available power density and the lifetime among potential energy sources [4]

Light, temperature, and vibration energies have been widely studied in the past decade due to their potential conversion efficiency, ubiquitous energy sources, and simple fabrication. The solar cell, typically using amorphous silicon, was widely introduced as a power source for the wireless sensor, since the solar cell is the most developed and has the highest power density and longest lifetime [5, 6]. But the photonic device is limited by the time and place of its use: night time, cloudy weather, indoor, or shadows all impair the device. The decrease of output power by incidence of light source is also a challenge on this type of energy harvester.

Like photovoltaics, thermoelectric devices have been largely in development in the past decade to acquire electric power for low power devices by using Bismuth telluride ( $\text{Bi}_2\text{Te}_3$ ) and doping [7, 8]. Stordeur et al. developed a thermoelectric device that generated 45 mW with a 28 mm<sup>2</sup> device size [9], and Schneider et al. generated 60 mW with 16 cm<sup>2</sup> of area at the 5 °C of temperature gradient [10]. While the micro size of the device is restricted to how large a temperature gradient it can achieve and to how efficient it can be because of its size, Bottner et al. have obtained about 60  $\mu\text{W}/\text{cm}^2$  at 1 mm<sup>2</sup> size of device [11].

Table 1.1 Various potential power sources from ambient and comparison of meso size energy harvesting devices. The power density has unit in  $\mu\text{W}/\text{cm}^3$  or  $\text{cm}^2$ .

Energy sources	Power density ( $\mu\text{W}/\text{cm}^3$ or $\text{cm}^2$ )	Ref.
Light	10~15,000 as amount of lux	[12, 13]
Temperature	60 at 5 °C gradient	[11]
Vibration	250* with piezoelectric	[4]

\* Device scale is large over few  $\text{cm}^3$

When solar energy is not available, a vibration source is a good candidate ,due to its ubiquitousness and high-efficiency conversion methods. The characteristics of potential vibration sources are listed in Table 1.2. They range up to 200 Hz of frequency and acceleration under 10  $\text{m}/\text{s}^2$ .

Table 1.2 The potential vibration sources and their characteristics [4]

Vibration sources	Frequency (Hz)	Acceleration (m/s <sup>2</sup> )
Blender casing	121	6.4
Clothes dryer	121	3.5
Washing machine	109	0.5
Car instrument panel	13	3
Small microwave oven	121	2.5
HVAC vent in office building	60	0.2-1.5
Windows next to a busy road	100	0.7
DC on notebook computer	75	0.6
Second story floor of busy office	100	0.2
Refrigerator	240	0.1
Vehicles	5-2000	0.5-100

#### 1.4 Converting vibration to electricity

There are currently three types of electromechanical transducers being reported. Figure 1.4 shows the schematic diagram of conversion mechanisms such as electrostatic, electromagnetic, and piezoelectric.

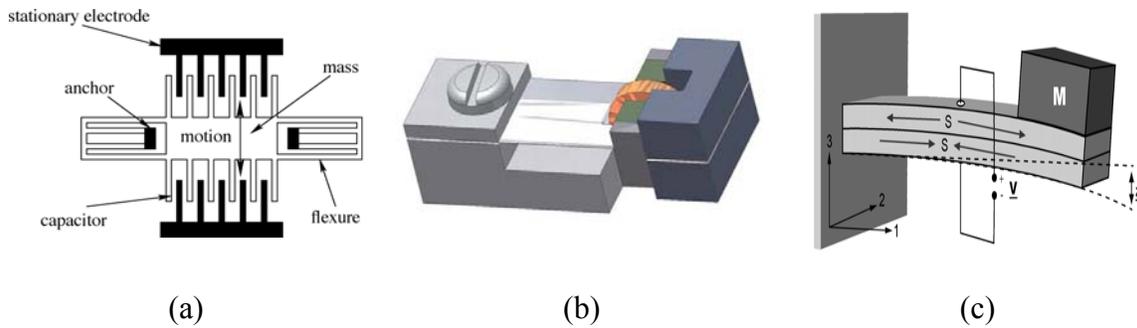


Figure 1.4 Illustration of the electromechanical transducer systems: (a) electrostatic, (b) electromagnetic, and (c) piezoelectric [14-16]

The power densities of these three types are compared and their current progress are simply commented on in Table 1.3. The piezoelectric conversion shows a higher mechanical-to-electrical transduction efficiency than the other types such as electrostatic and electromagnetic transducers [17, 18]. The piezoelectric conversion can be precisely controlled due to its simple configuration. The electromagnetic energy harvester can generate high power using an induced magnetic field in the device schemes. However, generating a magnetic field with MEMS is not easy because of miniaturization. The electrostatic method has an advantage in its device size, but it requires a high frequency and external voltage sources. The piezoelectric energy harvester is relatively easy to fabricate by MEMS technique and can be designed to obtain the target frequency with a high conversion efficiency.

Table 1.3 Comparison of three types of transducers

	Energy density	Fabrication	Current size	Disadvantages
Electromagnetic	24.8 mJ/cm <sup>3</sup>	Complex	Macro	Miniaturizing

Electrostatic	4 mJ/cm <sup>3</sup>	Simple	Integrated	External voltage & high frequency
Piezoelectric	35.4 mJ/cm <sup>3</sup>	Moderate	Macro & micro	Integrating piezo. thin films

## 1.5 Research objectives

The main objective of this research is to develop a piezoelectric energy harvesting device in a micro-electro-mechanical system (pMEMS) to provide sufficient power at the ambient frequency range. The objective will be addressed by developing the micromachining process through the integration of piezoelectric film, device fabrication, and advanced design. This research will be focused on the following specific tasks:

### 1. Integration of piezoelectric thin film on microsystems

Lead zirconate titanate (PZT) thin film is deposited on the silicon based microsystem. This integration is a most important issue on the piezoelectric sensor and actuator as well as the energy harvester. The characterization addresses microstructural analysis and electrical properties. The reliability study of piezoelectric thin film under cyclic movement is accompanied by the structural failure of the silicon based cantilever.

### 2. Development of MEMS technologies

The advanced technology of the MEMS process produces the robust reliability in fabricating the energy harvester with the successful integration of PZT thin films. The optimal

fabrication process is developed with regard to prevention of damage on the PZT film and electrodes and control of the dimension precisely for the tuning of resonant frequency. The etching and releasing of the cantilever is performed by dry etching using plasma, and their process conditions are studied.

### 3. Development of the energy harvester

High conversion efficiency is the main issue for piezoelectric energy harvesters in current technology. Based on robust fabrication technology, the single cantilever is designed to develop the piezoelectric MEMS energy harvester targeting ambient vibration sources. The shape of the cantilever is tailored to improve the power density by distributing the strain on the piezoelectric film.

Besides the improvement of power density, the cantilever arrays are fabricated to obtain high power and broadband working frequencies. The cantilever array design shows the feasibility of multiple cantilevers being integrated for a single application. The array device can be enabled by precise frequency control through the advanced fabrication and design.

## CHAPTER 2 GENERAL BACKGROUND

### 2.1 Piezoelectricity and PZT film

The performance of the piezoelectric energy harvester mainly relies on the properties of the piezoelectric materials. MEMS devices restrict the piezoelectric layer in a thin-film scale. Since the properties of piezoelectric thin film are different from those of bulk films due to scaling and substrate clamping effects, a film study should be included to increase the efficiency of the piezoelectric conversion. In this chapter, the theoretical background of the piezoelectricity and the thin film PZT will be introduced for a better understanding of the pMEMS device.

#### 2.1.1 Piezoelectricity

Piezoelectricity occurs when an electric surface charge develops on a crystalline material by mechanical stress. It was discovered by Jacques and Pierre Curie in 1880, and the inverse phenomenon, named the converse piezoelectric effect, was discovered one year later [19]. These charges, named by polarization ( $P$ ), are proportional to the applied stress ( $\sigma$ ) and are expressed by equation 2.1 or equation 2.2 for the converse piezoelectric effect. The material constant for piezoelectricity is called the piezoelectric strain constant ( $d$ ) and units are defined as  $pC/N$  or  $pm/V$ .

$$P = d\sigma \quad (2.1)$$

$$\varepsilon = dE \quad (2.2)$$

where  $\varepsilon$  is strain and  $E$  is electric field. If applying Hook's law to equation 2.1

$$P = dc\varepsilon = e\varepsilon \quad (2.3)$$

$$\sigma = c\varepsilon = eE \quad (2.4)$$

where  $c$  is the elastic modulus (Pa) and the piezoelectric stress constant ( $e$ ) is  $dc$  with  $\text{pC/m}^2$  of unit. The relationship between the mechanical variables and the electrical variables is shown in Figure 2.1.

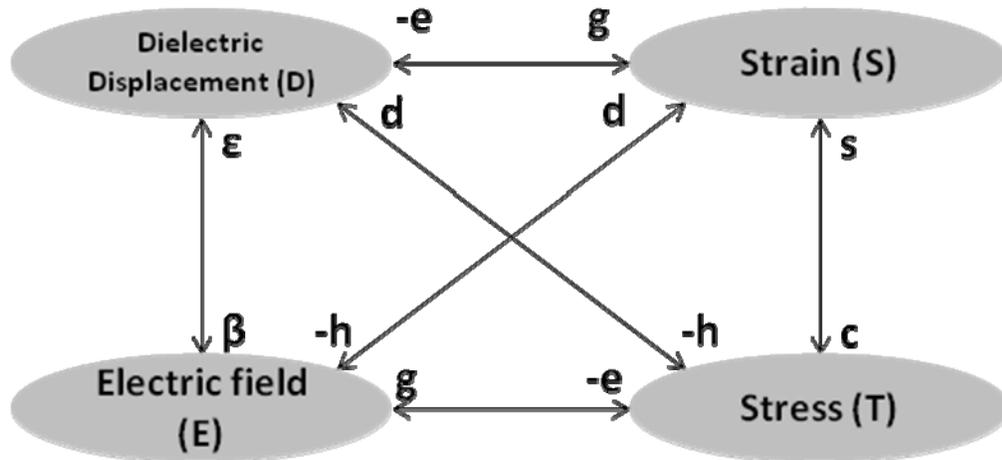


Figure 2.1 The piezoelectric effect and their variables

An additional constant related to piezoelectricity is the electromechanical coupling coefficient ( $k$ ), which quantifies the fraction of the electrical energy converted from mechanical energy or vice versa. It is defined by:

$$k^2 = \frac{\text{mechanical energy converted from electrical energy}}{\text{input electrical energy}} \quad (2.5)$$

$$k^2 = \frac{\text{electrical energy converted from mechanical energy}}{\text{input mechanical energy}}$$

By introducing Barium Titanate Oxide (BTO), piezoelectric ceramics were developed, and their commercialized applications have expanded to include high voltage sources, wave filters, delay line transducers, underwater sound and ultrasonic power, acoustic speakers, ink-jet printer heads, and various types of sensors [19, 20].

Piezoelectric materials have several forms, including polycrystalline, single crystalline ceramics, polymers, thin films, and composite films between polycrystalline and polymer. These materials should be chosen considering the desired frequency and fabrication process. The single crystalline ceramics include  $\alpha$ -SiO<sub>2</sub> (Quartz), LiTaO<sub>3</sub>, LiNbO<sub>3</sub>, etc. and have high piezoelectric properties, but their synthesis is limited by size and the cost of the substrate [21]. The piezoelectric polymer and composite polymers include the PVDF families, PZT-PVDF ( $d_{33}=25$  pC/N) [22], PZT-epoxy ( $d_{33}=80$  pC/N) [23], etc. While they show high piezoelectric properties, the synthesis of the materials is limited by size and cost. Now, most material studies for piezoelectric applications are focused on polycrystalline ceramics such as ZnO, AlN, PZT, doped PZT, and BaTiO<sub>3</sub>.

The piezoelectric effect originates from the symmetry of the crystal structure. Only non-symmetric structures can make displacement of gravity in the negative or positive charge under stress. The 32 crystallographic point groups have 11 centrosymmetric and 21 non-centrosymmetric structures, and 20 piezoelectric groups are involved in non-centrosymmetrics.

Piezoelectricity is a directional property between stress and discharge, so piezoelectric constants ( $d$ ) have different values in different directions. The relationship should be considered in three dimensions (3D) with tensor notation. The Hook's law in three dimensions can be expressed as

$$\sigma_{ij} = c_{ijkl} \varepsilon_{kl} \quad (2.6)$$

where  $\sigma$  is stress,  $\varepsilon$  is strain, and  $c$  is elastic modulus. Since the stress and strain can be expressed by a matrix, the tensor notation will be exchanged by a matrix notation. For example, the  $ij$  and  $kl$  will be substituted by  $p$  and  $q$ . When  $i, j, k,$  or  $l$  has 1,2, or 3,  $p$  and  $q$  have 1, 2, 3, 4, 5, and 6. Their relationships are shown in Table 2.1.

Table 2.1 Relationship between tensor notation and matrix notation

$j \backslash i$	1	2	3
1	1	6	5
2	6	2	4
3	5	4	3

Piezoelectricity can be mathematically described by a constitutive equation given by equation (2.7) [19]

$$S_p = s_{pq}^E T_q + d_{kp} E_k \quad (2.7)$$

$$D_i = d_{iq} T_q + \varepsilon_{ik}^E E_k$$

$$S_p = s_{pq}^D T_q + g_{kp} D_k \quad (2.7)$$

$$E_i = -g_{iq} T_q + \beta_{ik}^T D_k$$

where  $S$  is the strain component,  $T$  is the stress component ( $\text{N/m}^2$ ),  $E$  is the electric field component ( $\text{V/m}$ ),  $D$  is the electric displacement component ( $\text{C/m}^2$ ),  $s$  is the elastic compliance constant ( $\text{m}^2/\text{N}$ ),  $d$  is the piezoelectric constant ( $\text{m/V}$  or  $\text{C/N}$ ),  $\varepsilon$  is the dielectric constant of the piezoelectric material ( $\text{F/m}$ ),  $g$  is the piezoelectric constant ( $\text{V}\cdot\text{m/N}$  or  $\text{m}^2/\text{C}$ ), and  $\beta$  is the impermittivity components ( $\text{m/F}$ ). The superscript denotes a constant and zero condition of notation for measurement.

A cantilevered energy harvester has in-plane strain on the piezoelectric film coated on the structure in this study. The piezoelectric film has top and bottom electrodes that collect the discharge in the normal direction of the strain. The constitutive equations used in this configuration become

$$S_1 = s_{11}^E T_1 + d_{31} E_3 \quad (2.8)$$

$$D_3 = d_{31} T_1 + \varepsilon_{33}^E E_3$$

and the sandwiched electrode between piezoelectric film with a cantilever structure use the 31 mode of piezoelectricity. We mainly consider constants of  $d_{31}$  and  $\varepsilon_{33}$ .

### 2.1.2 Piezoelectric materials for energy harvesters

The energy harvesters require very high piezoelectricity, which is an inherent property of materials, and the performance of the device is seriously affected by piezoelectric materials. When the energy harvester is miniaturized by a MEMS process, the successful integration of

piezoelectric materials in thin film scale should be follow. The common piezoelectric thin films such as the lead zirconia titanate (PZT) family, zinc oxide (ZnO), aluminum nitride (AlN), PVDF, and BaTiO<sub>3</sub> (BT) have been used for piezoelectric applications. Piezoelectric properties are indicated by a piezoelectric coefficient (*d*) and a coupling coefficient (*k*), which relates strain to electric field and the ability to convert mechanical energy to electrical energy or vice versa. The typical values of *k* for unbounded piezoelectric materials are listed in Table 2.2.

Table 2.2 Coupling coefficient values of typical piezoelectric materials

<i>Materials</i>	<i>Coupling coefficient (k)</i>
Quartz	0.1
ZnO	0.33
BTO	0.4
PZT	0.5 ~ 0.7

The piezoelectric constant in thin film scale, however, has different values when compared to bulk materials, since the film is clamped on the substrate. The effective  $e_{31}$  and  $d_{33}$  for the thin film can be estimated by equations 2.9 and 2.10 [2].

$$e_{31} = \frac{d_{31}}{s_{11}^E + s_{12}^E} = e_{31} - \frac{c_{13}^E}{c_{33}^E} e_{33} \quad (2.9)$$

$$d_{33} = \frac{e_{33}}{c_{33}^E} = d_{33} - \frac{2s_{13}^E}{s_{11}^E + s_{12}^E} d_{31} \quad (2.10)$$

PZT films have been chosen as a transducer layer for energy harvesting devices due to their excellent electromechanical properties with high piezoelectric constants and electromechanical coupling coefficients. They have been compared with other materials such as ZnO, AlN, and poly(vinylidene fluoride) (PVDF), as shown in Table 2.3 [24-29]. The piezoelectric constants in Table 2.3 are typical values of films less than 1  $\mu\text{m}$ . Under a given force on a cantilever, the stress on the surface of the structure is different depending on the Young's modulus of the materials. For example, polymer PVDF has a low Young's modulus and makes a larger displacement on structures as compared to others. In this study, a thick Si membrane supports the piezoelectric film so that the stress is not changed much depending on the piezoelectric materials. Stiff silicon layers govern the displacement and stress on the surface. PZT was chosen only considering its conversion efficiency.

Table 2.3 Comparison of piezoelectric coefficients of the representative materials in thin film

	ZnO	AlN	PVDF	PZT (x=0.48)
$d_{31}$ (pC/N)	-2.3~-4.7	-2	-8	-40~-94
$k_{31}$ (CV/Nm)	0.19		0.12	0.22~0.31
$d_{33}$ (pC/N)	7.5~12.4	3.4~5.1	30~33	90~223
$k_{33}$ (CV/Nm)	0.28~0.41	0.21	0.16	0.49~0.67

The values vary in terms of additives, process conditions, and thickness ( $< 1 \mu\text{m}$ )

### 2.1.3 Lead Zirconate Titanate (PZT)

The perovskite structure of PZT consists of oxygen and small Ti or Zr atoms in symmetrical octahedral sites, and Pb cations occupying dodecahedral sites, as shown in Figure 2.2.

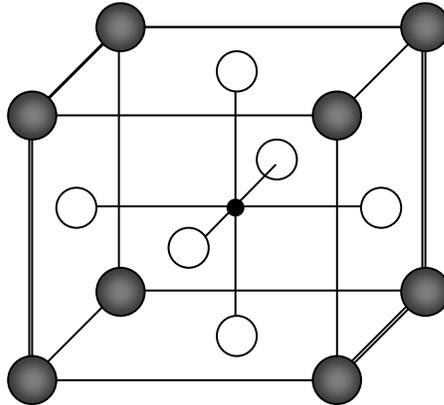


Figure 2.2 Perovskite structure

The PZT shows that its crystal structure depends on the temperature, as well as the ratio between Ti and Zr, as shown in Figure 2.3. Above the Curie temperature, the solid solution of  $\text{PbZrO}_3$  and  $\text{PbTiO}_3$  shows a paraelectric cubic structure. The Curie temperature is the limited temperature of the piezoelectric PZT material, and it ranges from 510 K to 800 K. The cubic perovskite structure is distorted below the Curie temperature into a rhombohedral phase for Zr-rich compositions or a tetragonal phase for Ti-rich compositions. The Curie temperature ( $T_c$ ) is a transition temperature of ferroelectricity and is defined as  $\epsilon = C/(T-T_\theta)$ , where  $\epsilon$  is the permittivity,  $C$  is the Curie constant, and  $T_\theta$  is the Curie-Weiss temperature. As the composition of Zr or Ti, the PZT has a different phase between the rhombohedral and tetragonal ones. The boundary of those two phases is called the morphotropic phase boundary (MPB) around the Zr/Ti composition of 0.52/0.48. The tetragonal structure has six  $\langle 100 \rangle$  polarization directions and the rhombohedral structure has eight  $\langle 111 \rangle$  polarization directions. The maximum piezoelectric and dielectric constants are commonly observed at MPB due to high probability of

polarization, and they are more sensitive to the composition than to the temperature because the MPB is very vertical.

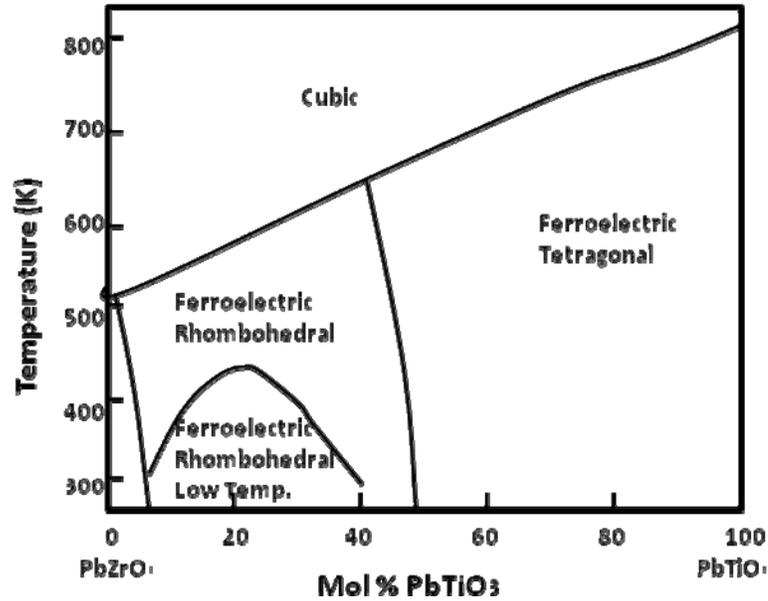


Figure 2.3 Phase diagram of PZT [30]

#### 2.1.4 Poling & depoling

When the material becomes ferroelectric with losing its symmetry, small-sized cations ( $\text{Ti}^{4+}$  or  $\text{Zr}^{4+}$ ) move to the stable sites, which are determined by the minimum free energy through the distortion of the structure. These movements enable charge separation and spontaneous polarization.

The fabricated ferroelectric films have spontaneous polarization, but they do not show ferroelectric behavior. Since the electric dipoles are randomly oriented before applying an external electric field on PZT, the net polarization is zero. The uniform electric field or strain is

applied to rearrange the dipoles and obtain ferroelectric behavior. This is called “poling,” and the dipole movement with poling is shown in Figure 2.4.

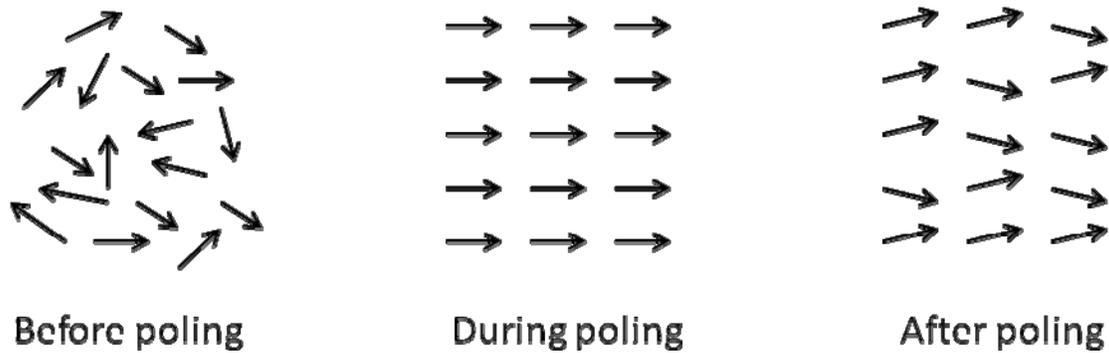


Figure 2.4 The dipole moment during the poling process

The localized dipole with uniform alignment is called the domain, and their boundary is the domain wall. The neighbor domains should have different polarization directions, and the motion of the domain wall is the origin of the ferroelectric mechanism, and the non-180° domain wall motion can be the origin of the nonlinear piezoelectric properties. It is known that the polarization directions depend on the crystallographic direction as shown in Figure 2.5. The tetragonal structure makes two kinds of polarization in antiparallel and perpendicular directions that refer to the 180° and 90° domain walls, respectively. The rhombohedral structure also provides more spontaneous polarization in the diagonal directions that refer to the 71°, and 109° domain walls. It is known that the domain and domain walls are created to reduce the free energy of the electrostatic and elastic energy in materials. Only the non-180° domain walls contribute to the reduction of the elastic energies so that the piezoelectric properties are dominated by them [31, 32]. This is called the extrinsic contribution. The intrinsic contribution is the lattice contribution from the displacement of each ion.

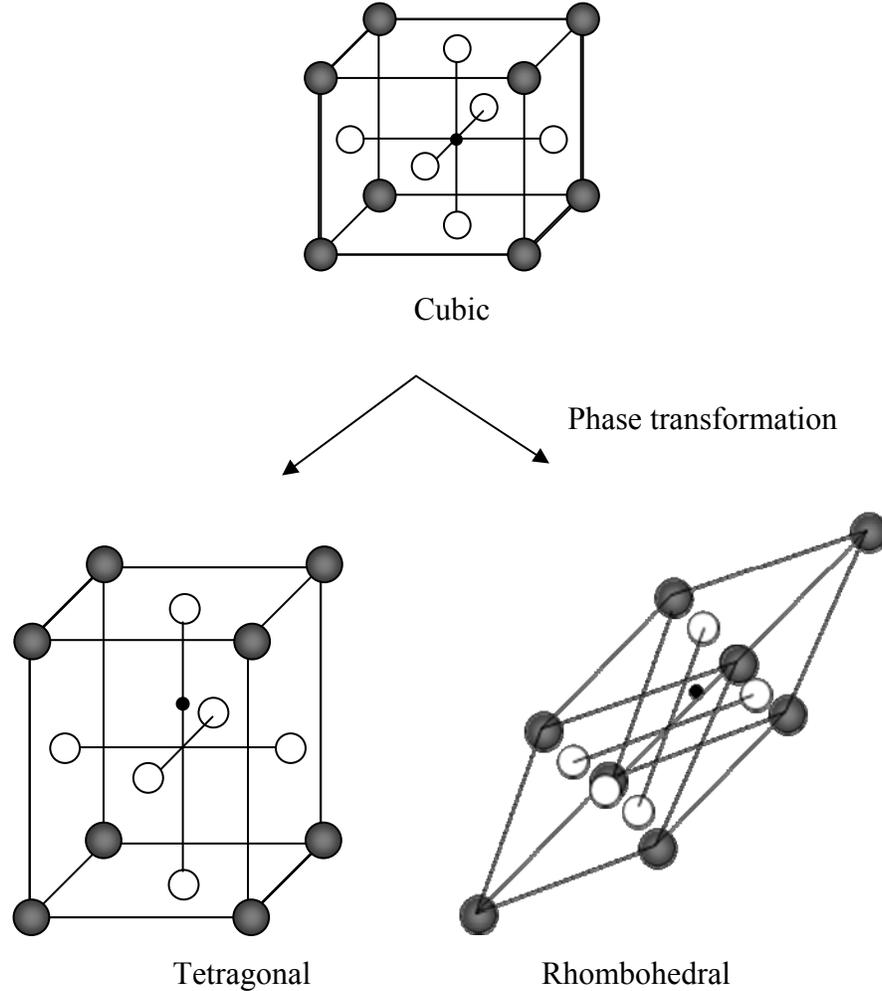


Figure 2.5 The shape change by the polarization as the crystallographic structures

Contrary to the poling of the materials, the dipoles rearrange to an equilibrium state as time goes by without the electric fields or mechanical stress. This is called “depoling,” and the piezoelectricity becomes degraded. The depoling of ferroelectric thin films is a factor to determine their lifetime and functional degradation of piezoelectric applications. For example, Kholkin et al. poled sol-gel deposited PZT film by UV. The piezoelectric constant and aging (depoling) improve with poling time, temperature, and electric field [33]. They have saturated

values when poled over 100 °C and 600 seconds. The electric field should be high enough to increase the active domains, but not breakdown the voltage because of the thickness of the films.

### 2.1.5 Piezoelectric mode ( $d_{33}$ & $d_{31}$ )

Randomly oriented PZT film can be polarized and produce electricity in the direction of the electric field. The  $d_{33}$  (longitudinal) mode has the motion parallel to the electric field. The  $d_{31}$  (transverse) mode utilizes the motion perpendicular to the electrical field. In the  $d_{15}$  (shear) mode, the electric field is perpendicular to the polarization and induces shear strain.

In bulk PZT (52/48) at MPB,  $d_{33} = 223$  pC/N,  $d_{31} = -94$  pC/N, and  $d_{15} = 494$  pC/N were measured [34]. Although the largest motion can be generated from  $d_{15}$ , the structure using  $d_{15}$  is hard to build up. Therefore, the  $d_{33}$  and  $d_{31}$  mode is usually developed, as shown in Figure 2.6. When the cantilever bends up and down, the strain and polarization occur in the direction of right and left. The electrodes are placed in a transverse direction that is perpendicular to the polarization direction. Figure 2.6(a) has the 31 mode. An Interdigital Electrode (IDE) is available for the 33 mode where the electrodes collect polarization in a normal direction.

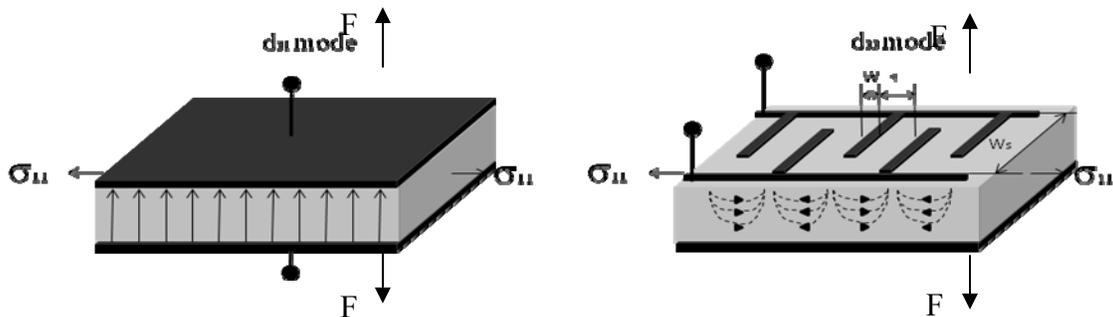


Figure 2.6 Electrode configurations of  $d_{31}$  and  $d_{33}$  for cantilever beam

The energy of the capacitor is expressed by  $E=CV^2/2$ . Although the  $d_{33}$  mode has a dielectric constant ( $d_{33}$ ) that is twice as large, the IDE design has a smaller charge density than that of  $d_{31}$ . The longitudinal mode has an incompletely coupled region with the  $d_{33}$  constant under the electrode and induces a different stiffness due to non-uniform poling. While the output voltage of the  $d_{31}$  mode is decided by the thickness of the PZT film, the configuration of IDE can regulate the output voltage in the  $d_{33}$  mode. Table 2.4 shows the comparison between the  $d_{33}$  and  $d_{31}$  modes.

Table 2.4 Comparison between the  $d_{33}$  and  $d_{31}$  modes for energy harvesters

	$d_{33}$	$d_{31}$
Pro	<ul style="list-style-type: none"> <li>- High dielectric constant</li> <li>- Easy output voltage regulation</li> </ul>	<ul style="list-style-type: none"> <li>- Large charge density</li> <li>- Easy control of PZT microstructure on Pt electrode</li> </ul>
Cont.	<ul style="list-style-type: none"> <li>- High % loss factor (wasted area under electrode)</li> <li>- Requires optimization of IDE</li> <li>- Complicated modeling due to IDE</li> </ul>	<ul style="list-style-type: none"> <li>- Low dielectric constant</li> </ul>

The optimization of IDE should be executed considering the longitudinal poling to prevent electrode loss. For example, it is intended that a narrow width of electrode is adequate due to the small loss under the electrode area. The simulations for optimized IDE design were developed by Finite Elements Analysis (FEA) [35, 36]. Knight simulated the maximum  $d_{33}$  polarization of IDE, which is obtained when large spacing ( $s$ ) between electrodes exists and when the polarization is decreased by increasing the space between the finger electrodes/thickness of PZT after a maximum value at 0.8. The width ( $W_s$ ) between the electrode strips should be four times

longer than the spacing ( $s$ ), and the polarization is dramatically decreased if spacing ( $s$ ) is less than  $10\ \mu\text{m}$  [36].

## 2.2 Electrical power management

The total system of self-power is divided between the piezoelectric energy harvester and the storage circuit system as shown in Figure 2.7. When designing the energy harvester, its efficiency is a major issue, and the link between the energy harvester and the managing circuit should also be considered.

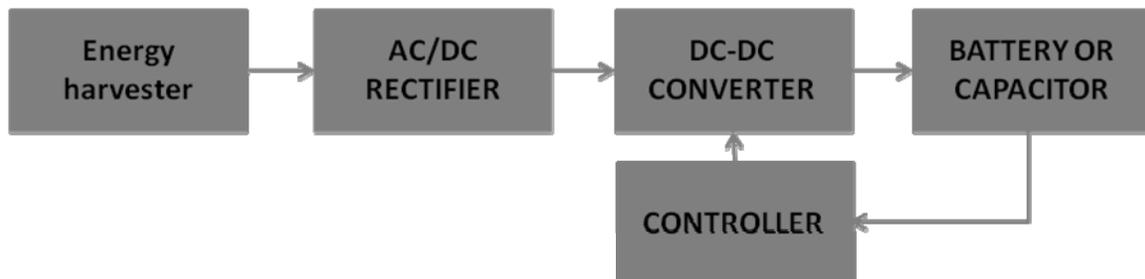


Figure 2.7 A schematic diagram of a standard energy harvesting system

The vibration source resonates the cantilevered energy harvester and makes cyclic stress on the piezoelectric layer. Piezoelectric energy harvesters generate an AC current parallel to a complex output impedance while the device structure resonates at the resonant frequency. Since electronics requires stable DC voltage, the output from the energy harvester should be rectified and regulated for the target electronics. The process signals are illustrated in Figure 2.9.

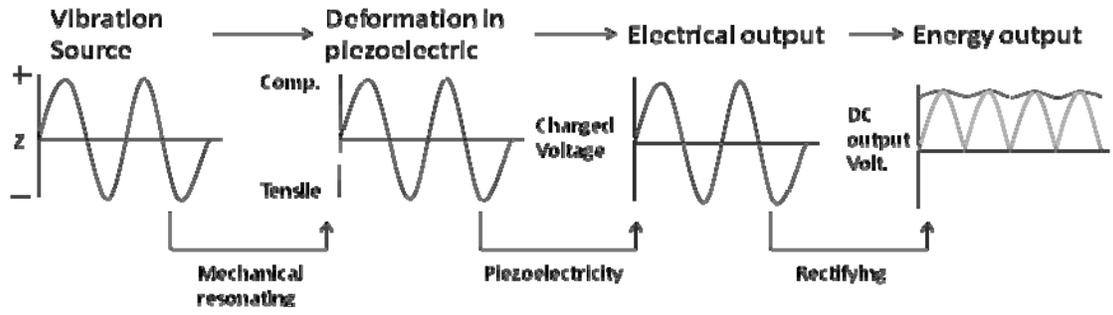


Figure 2.8 The signals during piezoelectric energy harvester system

The standard energy harvesting circuit is shown in Figure 2.9. The AC signal is rectified by full-wave by Wheatstone Bridge and the filtering capacitance is followed to smooth the DC voltage. The resistive load regulates the voltage. Since the generated electric output is mostly low and intermittent to be provided to the electronics directly, accumulation or storage of the electrical energy is required.

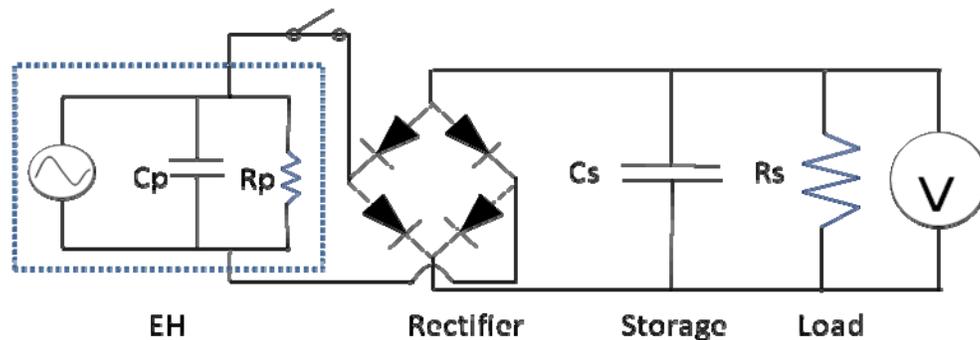


Figure 2.9 A standard energy harvesting circuit

The output voltage ( $V_p$ ) generated from the energy harvester (EH) should be high enough not to be blocked on the rectifier. The diode rectifier has its own threshold voltage, and the

power loss occurs through the diode proportional to its threshold voltage. The diode should be selected considering power losses with low dissipation. The MOS transistor is a possible application with the advantages of scaling down [37]. In the mean time,  $V_p$  should be high enough but not too high to breakdown the circuits. The power losses occur through all circuit elements and are constituted as a portion of  $V_p$ .

To maximize the power delivery, the impedance of the energy harvester should be optimized to match that of circuits or vice versa. Piezoelectric energy harvesters produce low current compared to the generated voltage. The matching impedance might reduce the power losses on circuits. A higher current, moreover, will reduce the charging time on capacitors.

The electric circuit for piezoelectric energy harvesters has been developed considering the energy harvester [38-40]. The circuits study predicts the electrical behavior of the piezoelectric energy harvesting system considering the electrical load, the resonant frequency, the damping effect, and the electromechanical coupling coefficient. One of the electric circuits for the piezoelectric energy harvesting system called “Synchronized Switch Harvesting on Inductor” (SSHI) was proposed [41]. SSHI consists of a switch and an inductor in serial connection. The switch has a closed state when the transducer makes the maximum displacement, and the inductance and capacitance of the piezoelectric film produces an oscillator. The SSHI increases the output power effectively, but the quality factor of this nonlinear processing circuit is still an issue on efficiency.

## **2.3 Piezoelectric MEMS energy harvester**

### **2.3.1 Single piezoelectric MEMS energy harvester**

The piezoelectric energy harvester in a bulk scale has been successfully verified as a possibility to be a power source for self-sustained wireless sensors. Their large size, however, restricts their integration into miniaturized wireless sensors.

In the early 2000s, researchers started to focus on the modeling and fabrication of a piezoelectric energy harvester in a micro scale. Demir et al. reported the strain, deflection, and generated output power for a power generator with a PZT membrane. They also investigated the residual stress and compositional function of PZT thick film [42]. Lu et al proposed the modeling of 31 mode micro power generators with films of both PZT and PZN-PT single crystal [43].

The reported prototype of MEMS energy harvesters are compared in Table 2.5. The piezoelectric energy harvester based on a cantilever structure with a proof mass is widely used due to its large strain at a given acceleration force and easiness to build up a low resonant frequency. Marzencki et al. fabricated a cantilever-based energy harvester with an Si proof mass. It seems that their energy harvester generated a relative low power of  $0.038 \mu\text{W}$  due to the low piezoelectric properties of AlN film [44]. Later, they integrated the power management circuit on multiple cantilevers. Multiplied voltage was generated throughout the capacitance load, but power was low, in the tens of nW at 0.4 g acceleration condition [37, 45]. Jeon et al. first developed a PZT-based piezoelectric energy harvester using 33 mode by using interdigitated electrodes. They controlled the residual stress by selecting an oxide buffer layer to prevent bending of the cantilever. Their device generated about  $1 \mu\text{W}$  power and 2.4 DC voltage after rectifying and storing at 13.9 kHz and 10 g of resonant frequency and acceleration [46]. The

optimal resistance was  $5.2 \text{ M}\Omega$  due to a small area of IDE configuration as compared to that of the electrode plate in 31 mode. Fang et al. aligned a Ni proof mass at the end tip of the PZT cantilever for a larger displacement. A thick PZT transducer layer with top and bottom electrode plates generated  $2.16 \text{ }\mu\text{W}$  of output power at 1 g and 609 Hz conditions [47]. The resonant frequency of their energy harvester was reduced as compared to previous ones due to the thick beam structure with the silicon layer, but it is still higher than ambient vibration sources. The alignment of the Ni mass is not compatible for the MEMS process and restricts further miniaturization. Lee et al. compared the 31 mode and 33 mode using  $11 \text{ }\mu\text{m}$  thick aerosol deposited PZT film [48]. While the 33 mode energy harvester generated 2.3 V with resistive load and  $0.35 \text{ }\mu\text{W}$ , the 31 mode device generated 1.6V with resistive load and  $1.1 \text{ }\mu\text{W}$  under 1 g acceleration force and 256 Hz of resonant frequency. The aerosol deposited thick PZT induced serious bending on the cantilevered structure, and it seems that the film has a large leakage due the microstructure of the aerosol deposited films based on its remanent polarization value of less than  $10 \text{ }\mu\text{C}/\text{cm}^2$ . Elfrink et al. generated a high output power of  $60 \text{ }\mu\text{W}$  using a large silicon mass and an AlN film under 2 g acceleration force and 572 Hz of resonant frequency [49]. Their energy harvester generated about 2 V through a load of  $500 \text{ k}\Omega$ . They show that AlN is a viable choice for piezoelectric energy harvesters due to its low dielectric constant and simple deposition techniques. The low dielectric constant induced a high optimal resistive load and a relative high voltage as compared with that of the PZT film, but it has a large discrepancy when matching impedance with circuits. Shen et al. [50, 51] generated a moderate level of power of  $2.15 \text{ }\mu\text{W}$  and 160 mV under 2 g acceleration force and 465 Hz. Although the low impedance of  $6 \text{ k}\Omega$  generated a relatively high power, its low output voltage is still a limitation for circuitries under the typical threshold voltages of MOFET transistors. The later report of a similar structure shows

less output power. It seems that this discrepancy is caused from unstable film properties of PZT. A laser machined power generator in a few mm size of scale was tried, but the process is inherently limited to scaling down and showed a low power density at  $301 \mu\text{W}/\text{cm}^3$  [52].

Table 2.5 The comparison of reported MEMS energy harvesters

<i>Author</i>	<i>Device cantilever/mass</i>	<i>Mode</i>	<i>Conditions (<math>\text{ms}^{-2}</math> / Hz)</i>	<i>Power (<math>\mu\text{W}</math>)</i>	<i>Power density (<math>\mu\text{W}/\text{cm}^3</math>)</i>
Marzencki et al.	AlN / Si	31	4.9 / 204	0.038	10
Marzecki et al.	AlN / Si	31	39.2 / 1368	1.97	3569
Y.B.Jeon et al.	PZT / (Su-8)	33	106 / 13.9 K	1	37,037
Fang et al	PZT / Ni	31	9.8 / 608	2.2	10,843
B S Lee et al.	PZT / Si	31 33	9.8 / 255.9 9.8 / 214	2.765 1.288	6513 2105
Elfrink et al.	AlN/Si	31	19.6 / 572	60	~5214
Shen et al.	PZT / Si	31	19.6 / 465 7.36 / 184	2.15 0.32	3,272 416

### 2.3.2 Multi-beam arrayed piezoelectric MEMS energy harvester

The single cantilever has a limitation of low output power, low voltage for the threshold

of circuitries, or low current in recent technology, which is insufficient for wireless sensor networks. Moreover, their narrow operation frequency range also restricts their uses under the uncertainty of the frequency of ambient vibration sources. The cantilever array was proposed to solve these issues, and the later one has been introduced in bulk and MEMS scale to widening working frequency regions.

The cantilever-based energy harvesting arrays were introduced in bulk scale by Ferrari et al. At the certain frequency of resonating, the measured signals from the open circuit definitely created interference. For three cantilevers designed to be resonated at 113 Hz, 183 Hz, and 281 Hz of frequency, their output signals were rectified and charged to the capacitor individually [53]. Liua et al. first introduced micro multi-array power generators for the broadband working frequency. They used PZT thin film and aligned an Ni proof mass at the end of the tip. Three cantilevers generated 2.55  $\mu\text{W}$ , 2.1  $\mu\text{W}$ , and 1.87  $\mu\text{W}$  at their own resonant frequency. Their MAPG covers the frequency range from 226 to 234 Hz, and they generated 3.98  $\mu\text{W}$  and 3.93 DC output voltage charge in capacitor at the 229 Hz of frequency [54].

With the output power, the voltage and current should be regulated when connecting to the power management circuit. In bulk scale, multi layers of piezoelectric transducer layers may regulate the output voltage and current, but its scheme is hard to realize due to limitations of film technology, such as internal stress and keeping piezoelectric properties for multi layers. In this study, a cantilever array design is used for regulating the output voltage or current with the increase of output power.

## CHAPTER 3 INTEGRATING PIEZOELECTRIC THIN FILM ON SILICON WAFER

The properties of the piezoelectric layer are important to construct a highly efficient energy harvester. Since the properties of the piezoelectric layer are influenced by the deposition method and the types of layers and substrate, an investigation of these parameters was executed in these issues.

### 3.1 Film characterization

The dielectric and ferroelectric properties of the PZT thin films were characterized by an impedance analyzer (HP 4192A) and a TF Analyzer 2000 (aixACT, FE-module). For dielectric properties, four different sizes (0.01, 0.04, 0.09, 0.25 mm<sup>2</sup>) of square electrodes were patterned on PZT and the dielectric constants were measured with the dielectric loss. The hysteresis loops of PZT films were characterized for both the small square electrodes and the actual device to evaluate and inspect the piezoelectric performance of each PZT layer after fabrication. With measured remnant polarization and dielectric constant, the piezoelectric coefficient can be roughly estimated by following Equation 3.1.

$$d_{31} = 2Q_{12} \epsilon_0 \epsilon_r P_r \quad (3.1)$$

where  $Q_{12}$  ( $-3.1 \times 10^{-2} \text{ m}^4 \text{C}^{-2}$ ) is the electrostrictive constant,  $\epsilon_0$  is permittivity of vacuum,  $\epsilon_s$  is relative dielectric constant, and  $P_r$  is remnant polarization.

The deposited PZT films were characterized by inspecting their microstructures and crystallographic orientations using a Scanning Electron Microscope (SEM, JEOL) and X-ray Diffraction (XRD)

### 3.2 Thin film deposition

Various deposition methods for ferroelectric thin films have been investigated by many researchers such as chemical solution deposition (CSD), metalorganic chemical vapor deposition (MOCVD), sputter, pulsed laser deposition (PLD), and molecular beam epitaxy (MBE) deposition. Besides the film properties, the scalability to the wafer size is most important when integrating the films on MEMS devices. Their characteristics are summarized in Table 3.1.

Table 3.1 Comparison of deposition methods for ferroelectric materials

	CSD	MOCVD	Sputter	PLD	MBE
Mechanism	Chemical reaction	Chemical reaction	Momentum energy	Thermal energy	Thermal energy
Deposition rate	Low	Moderate	Low	Moderate	High
Scalability	Large	Large	Large	Small	Large

Since our energy harvester requires high efficiency of conversion between strain and electricity, the PZT film was chosen based on the reported materials constant in Table 2.1. To obtain best piezoelectric properties of PZT films, precise control of the  $\text{PbZr}_{1-x}\text{Ti}_x\text{O}_3$  composition is required, which is strongly influenced by the thin film process. Sputtering, PLD, and CSD are commonly used to grow perovskite PZT films with a high piezoelectric coefficient. The sol-gel derived PZT deposition is widely used for thin films, since its films show a high coupling coefficient with precise control of composition and the process is simple and cheap [55-58]. The typical process of sol-gel deposition is shown in Figure 3.1. The sol-gel deposited films are highly affected by precursor chemistry, deposition conditions, and heat treatment conditions. This study experimentally investigates the effect of heat treatment and interaction with conjugated films.

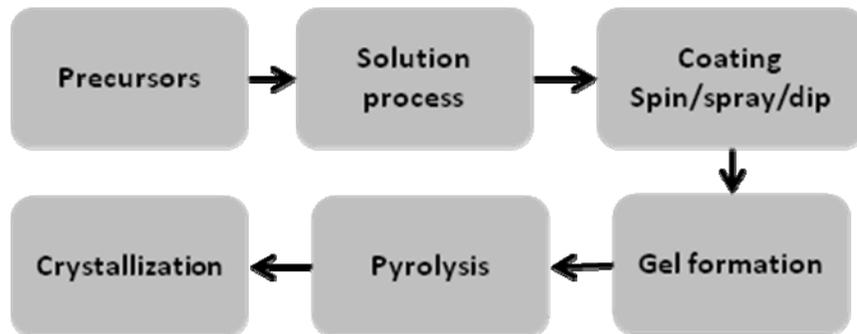


Figure 3.1 Typical sol-gel derived chemical solution deposition (CSD) process.

### 3.3 Chemical Solution Deposition (CSD) of PZT film

The PZT film in this study was deposited by CSD method to obtain a high stoichiometry composition. It is common that the piezoelectric properties of thin film cannot exceed that of

bulk PZT due to limited thickness and internal stress during clamping of the film on substrate. Many researchers have investigated the development of PZT thin film in recent years, and CSD is the most used method for stoichiometry [55-58]. The PZT film study was performed on the composition, substrate effect, thickness, and heat treatment.

We should design the PZT considering the composition of Zr/Ti and the crystallographic orientation as the application of the energy harvester that requires a high piezoelectric coefficient ( $d_{ij}$ ).  $\text{Pb}(\text{Zr}_x\text{Ti}_{1-x})\text{O}_3$  shows maximum piezoelectricity at the rhombohedral-tetragonal phase boundary ( $x=0.48$ ), which is the so-called morphotropic phase boundary (MPB) in Figure 2.3.

In Figure 3.2, the coupling coefficient and piezoelectric constant have maximum values at 52% of zirconium. It is reported that the PZT thin films show the highest piezoelectric coefficient ( $d_{31}$ ) in the 31 mode at the MPB composition of 53% Zirconium and 47% titanium and (111)-textured orientation [59]. The film studies in this paper are performed based on this composition.

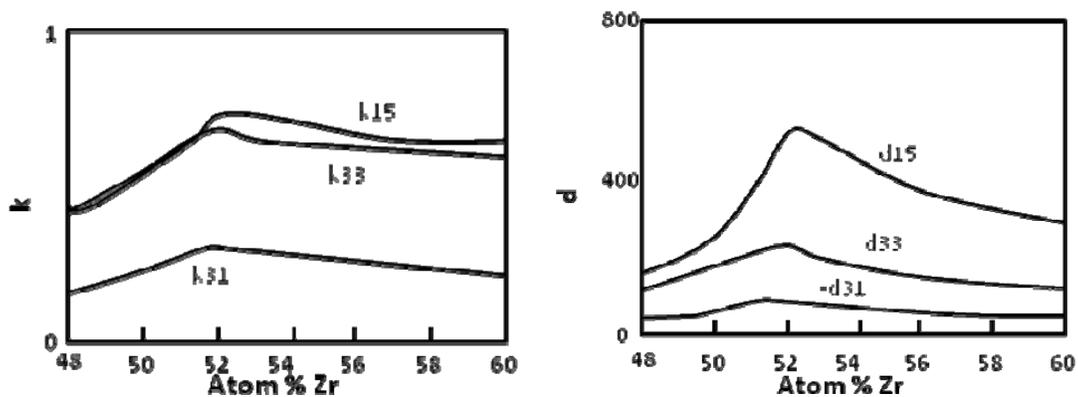


Figure 3.2 The schematic changes of coupling coefficient and piezoelectric constant as mole % of Zr [60]

A PZT (Pb:Zr:Ti=110:52:48) solution of commercial product from Inostek. Inc. was coated on platinized silicon wafers by chemical deposition as 12 layers, which make about 1  $\mu\text{m}$  thickness. The solution was dropped on the entire wafer with filtering probable particles (larger than 0.2  $\mu\text{m}$ ) in solution, and it was coated by spinning at 4000 rpm for 20 second with a high acceleration rate. Each layer was pyrolyzed at 350  $^{\circ}\text{C}$  for 10 minutes on a hot plate or furnace at the ambient condition and annealed at 650  $^{\circ}\text{C}$  for 15 minutes by Rapid Thermal Annealing (RTA) or in a furnace. The PZT film was finally annealed at 650  $^{\circ}\text{C}$  for 30 minutes, and its process is shown in Figure 3.3.

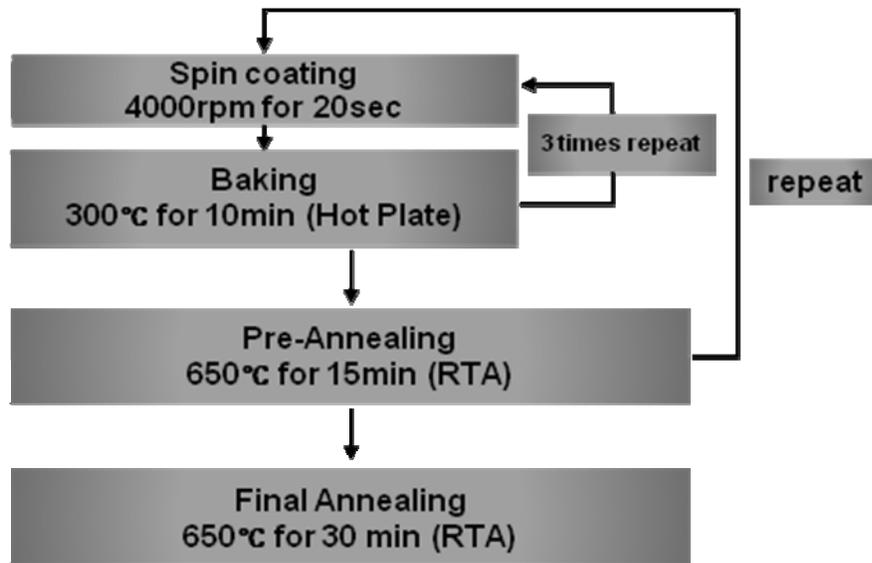


Figure 3.3 Process flow of PZT chemical deposition

A platinum (100nm) layer was used for the bottom electrode due to its stability at high temperature annealing. The bottom Pt was deposited on the amorphous  $\text{SiO}_2$ , which played the

role of electrical barrier. To enhance the adhesion between the Pt and SiO<sub>2</sub> interlayers, Cr, Ta, and Ti were used.

### 3.3.1 Analysis of PZT thin film on Pt substrate

Sol-gel derived PZT film was deposited on (111) textured PT electrodes. The properties of the PZT films that were deposited on the substrate of Pt(120nm)/interlayer/SiO<sub>2</sub>(500nm) were investigated by XRD and SEM for their microstructure, and their electrical properties were also investigated with PE hysteresis loop and dielectric constant. It is a general agreement that (100) and (111) preferred orientation PZT film is observed on the Pt (111) substrate. The Pt was deposited by a sputter system under the conditions of 100W DC power for 360 second, and 5 mTorr. The crystallographic orientation of Pt was firstly inspected as shown in Figure 3.4.

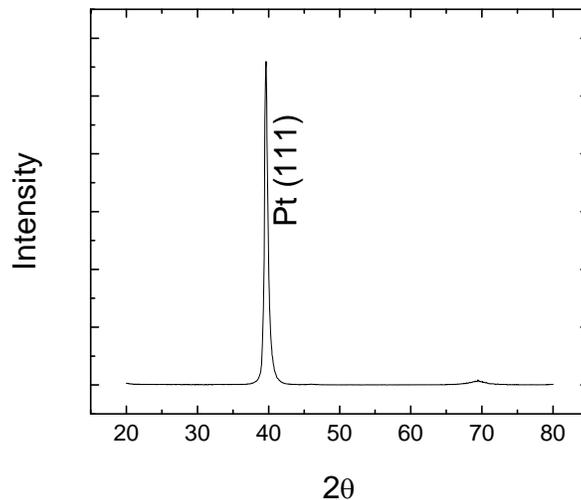


Figure 3.4 XRD data of as deposited Pt/Ti (120/10 nm) by sputtering

When considering minimal interface energy between PZT and Pt, PZT (111) orientation is more favorable on Pt (111) due to lattice matching [61, 62]. Moreover, the intermetallic interface layer between PZT/Pt created by Ti diffusion through Pt energetically enhanced the (111) orientation of the PZT nucleus [63, 64]. The sol-gel process, however, has various factors that affect the film properties and its crystallographic orientation, such as the solution chemistry, substrate orientation and microstructure, and heat treatment [65, 66].

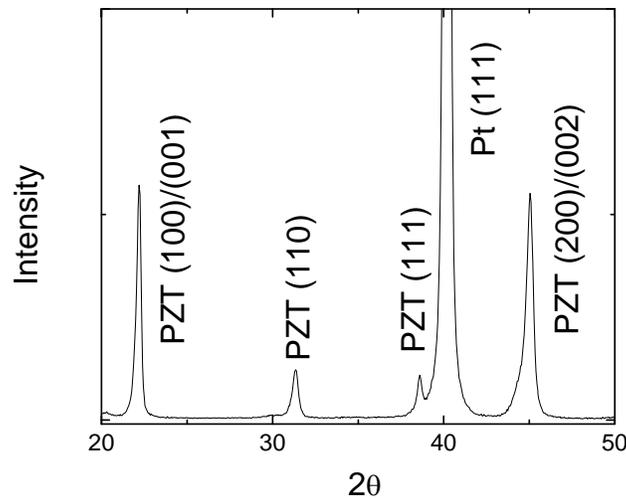


Figure 3.5 PZT/Pt film analysis by XRD

PZT (Zr=48%) whose thickness is 500 nm was deposited on Pt/Ti/SiO<sub>2</sub> substrate with 300 °C pyrolysis on a hot plate and 650 °C annealing in a tube furnace. The obtained orientation of the PZT film is a mixture of (100), (111), and (110), but the (100) orientation is the most preferred in Figure 3.5. The formation of (100) preferred orientation PZT is not fully understood yet. While the (111) texture of PZT is promoted by metastable interfacial layer (Pt<sub>5-7</sub>Pb), the (100) texture is enhanced by the pyrolysis temperature and time, and the texture can be stronger

as the thickness increases due to a lower interfacial energy and a faster growth rate [67-69]. No pyrochlore phase around  $29^\circ$  appeared on the XRD analysis. It is reported that (100) films show higher piezoelectric properties at the 31 mode than the (111) orientation films do [59]. Figure 3.6 shows the microstructure of the PZT film, which has a dense and columnar structure with  $\sim 100$  nm diameter of grain size. The roughness of the PZT film was measured as 2.5 nm by AFM analysis.

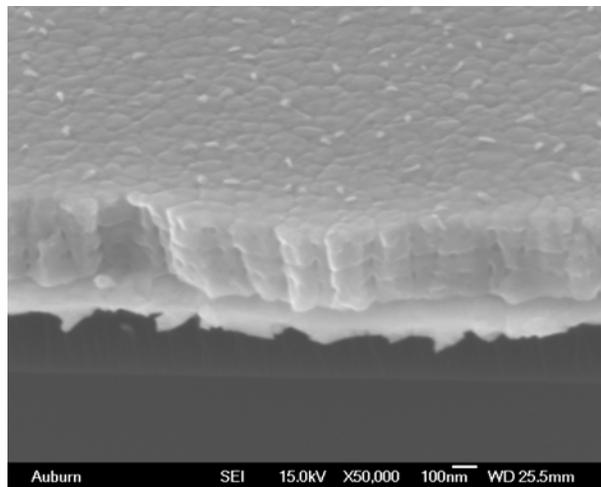


Figure 3.6 SEM image of PZT film

### 3.3.2 Adhesion layers effects on the electrical properties of PZT

The adhesion layers used between  $\text{SiO}_2$  and Pt were studied, and their effects on the electrical properties of PZT layer were investigated. It is known that the adhesion layers diffuse up through the Pt grain boundaries and affect the properties of PZT film during heat treatment [63]. It is assumed that the thin  $\text{TiO}_2$  layer plays the role of seeding layer for PZT

nucleation. First, the thickness of the adhesion layer was investigated. Thickness of 10, 40, and 100 nm Ti films were used for Pt bottom electrodes, and then PZT was deposited with the same conditions. Ti adhesion layers gave sufficient levels of adhesion in the tape test and the ultrasonic test in all thicknesses. The P-E hysteresis loops were measured for each sample, as shown in Figure 3.17. The thinnest Ti film shows the highest piezoelectricity of  $\sim 21.5 \mu\text{C}/\text{cm}^2$  in remanent polarization, as shown in Figure 3.7, while PZT films using thicker Ti adhesion layers show  $\sim 15 \mu\text{C}/\text{cm}^2$  after correcting asymmetry of loop. It seems that the limited source of the Ti layer impeded diffusion to the PZT layer while the thick films provided an excess of Ti on the Pt electrode.

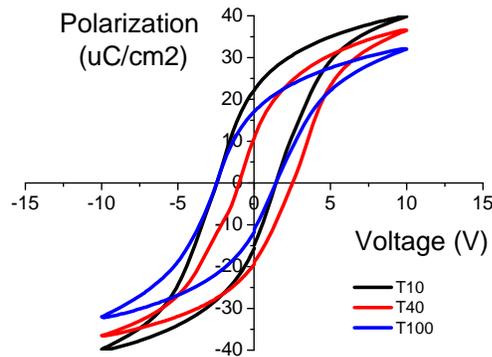


Figure 3.7 PE hysteresis loops of PZT films deposited on different thicknesses of adhesion layers

Various materials (Ti, Cr, and Ta) were deposited with 10 nm of thickness and PZT deposition followed. The polarizations and dielectric constants of the PZT films on different adhesion layers were measured as shown in Figure 3.8 (a) and (b). While PZT films with Cr and Ta adhesion layers show very low remanant polarization of  $6 \mu\text{C}/\text{cm}^2$  and  $11 \mu\text{C}/\text{cm}^2$ , the

film with Ti adhesion layer showed higher values up to  $21.5 \mu\text{C}/\text{cm}^2$ . When comparing the dielectric constants of PZT films on different adhesion layer, the films with Ti and Ta adhesion layers have a high value of about 1100. The piezoelectric constants ( $d$ ) were calculated from remanant polarizations and dielectric constants using equation 3.1, and shown in Table 3.2.

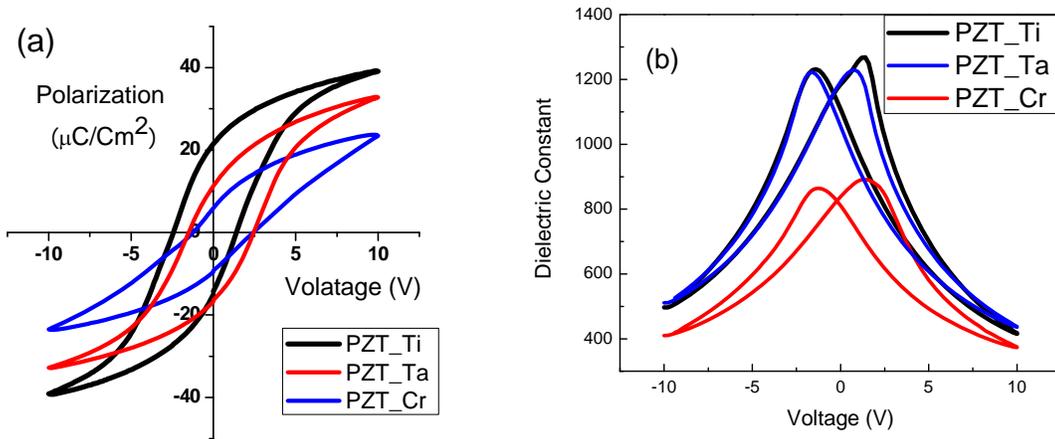


Figure 3.8 Substrate effect on electrical properties of PZT films. Adhesion layers deposited between  $\text{SiO}_2$  and bottom Pt, and their effects described by: (a-b) Ti, Cr, and Ta have different effects on PZT

Table 3.2  $P_r$ ,  $\epsilon_r$ , and  $d_{31}$  of PZT films deposited on different adhesion layers

Material	$P_r$ ( $\mu\text{C}/\text{cm}^2$ )	$\epsilon_r$	$d_{31}$ (pC/N)
Cr	5.95	838	-27
Ta	11.28297	1193	-74
Ti	21.5	1184	-140

PZT films on Ti and Ta adhesion layers had strong (111) orientation, but PZT films on Cr adhesion layers showed random orientations with both (111) and (110) orientations. When it is

assumed that the adhesion materials diffuse up and affect the nucleation of the PZT film, oxidized Cr may restrict the (111) oriented nucleation. In MPB composition, mixture orientations of (111) and (100) have higher dielectric constants than that of single-oriented film due to multiple polarization directions [70]. Figure 3.8 (b) and Figure 3.9 (a) show opposite results. It seems that Cr obstructed the oxide on the ferroelectric PZT film and decreased the dielectric constant [71]. In Figure 3.8 (a), the loop of the PZT/Pt/Cr was shifted because of a non-symmetric electrode configuration due to a second phase of Cr.

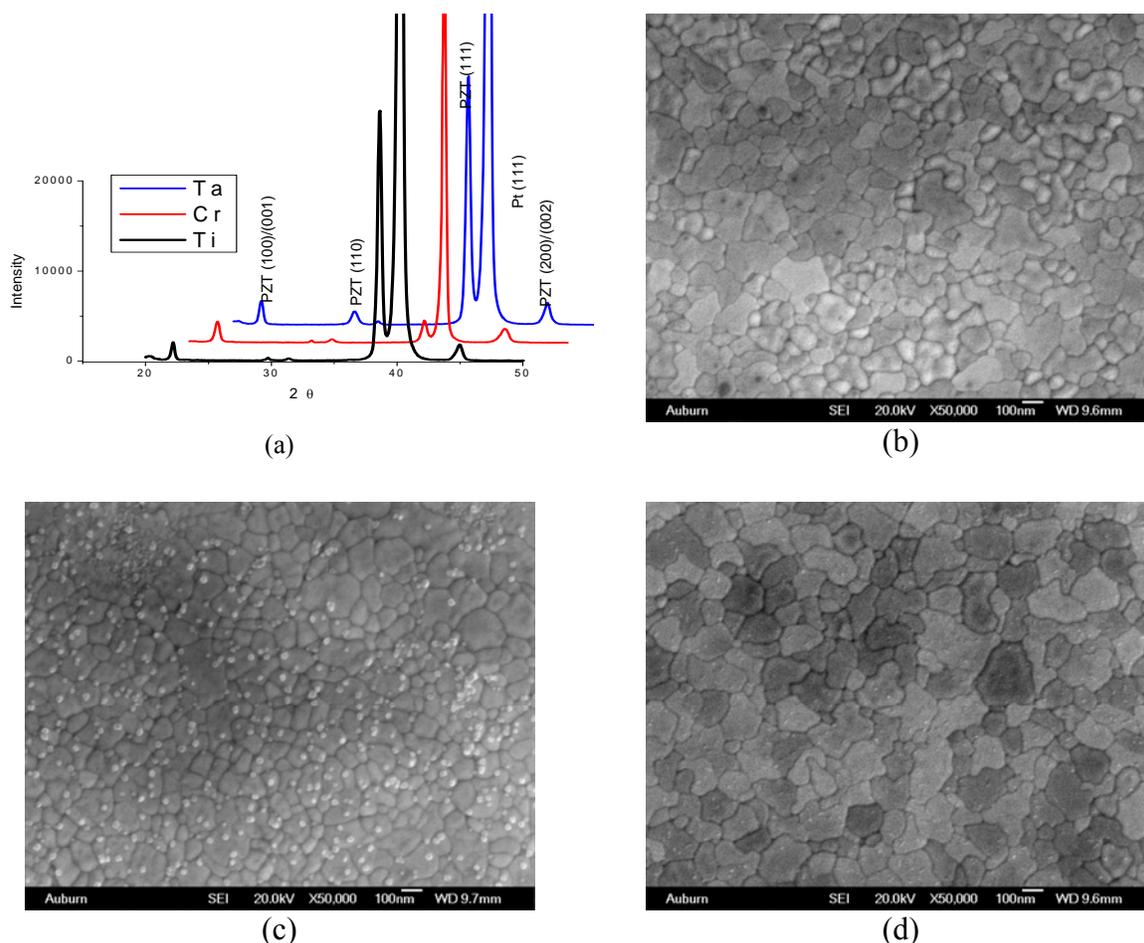


Figure 3.9 XRD and SEM analysis of PZT film on different adhesion layer: (a) XRD analysis, (b) Ti, (c) Cr, and (d) Ta adhesion layers

PZT film on the Ti and Ta adhesion layers had similar grain size, but the film on the Ta has a little more uniform grain size. The (111) preferred orientation of PZT may have a smaller grain size than that of the (100) oriented PZT. The small grain size on Cr seems to be attributed to its homogeneous nucleation and granular microstructure. In Figure 3.10, the cross section views of PZT films on Cr and Ti can be compared in granular and columnar forms, respectively. The hypothesis is that Cr diffuses and creates precipitation on each layer of PZT so that it gives a homogeneous nucleation site. PZT film on the Ti and Ta adhesion layers had a similar grain size, but the film on the Ta has a little more uniform grain size.

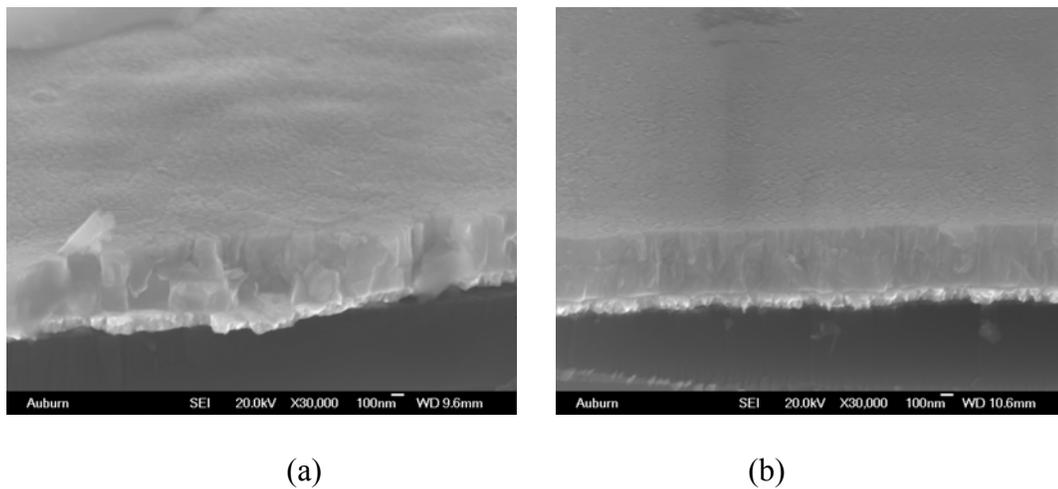


Figure 3.10 Angle view of PZT thin film on Cr and Ti adhesion layers: (a) Cr and (b) Ti

### 3.3.3 Heat treatment effects

The piezoelectric properties of PZT are affected by the heat treatment conditions of pyrolysis and annealing. In CSD deposition, the growth of PZT on a Pt substrate is mainly

influenced by precursor chemistry and substrate conditions [65]. Furthermore, the pyrolysis and annealing temperature are important in the formation of the preferred orientation. During the pyrolysis heat treatment (the typical range of temperature is 200 to 400°C), some reactions occurred, such as formation of volatile organic species (thermolysis), formation of volatile organic molecules by combustion (pyrolysis), dehydration for OH groups, and formation of carbonate ( $M-O-CO_2$ , oxidation). Since the films shrink during pyrolysis with rapid heat treatment, the films have a lot of internal stress, which can induce cracks. The stress can be controlled by keeping the viscoelastic gel on longer during heat treatment by adjusting the reactivity of precursors [72]. After the film reaches its amorphous structure during pyrolysis, crystallization is required by an annealing heat treatment. The nucleation and growth mechanism of the films can be understood by kinetic and thermodynamic driving force. The mobility of atoms, reactivity control by precursor chemistry, and the residual organics and constituent influence the microstructure and orientation of oxide films deposited by CSD during the entire process.

In this section, the pyrolysis heat treatment effect was inspected during CSD deposition. The hot plate is usually used for the pyrolysis process of gel formation of PZT [73]. The tube furnace and hot plate were used to compare the isothermal effect on the pyrolysis process. A total of six layers of PZT (500 nm) were deposited on Pt/Ti, and they were pyrolyzed by a conventional furnace and hot plate. All samples were annealed at 650 °C for 15 minutes after the pyrolysis process. Temperatures of 237, 290, and 330 °C on the hot plate were investigated, which was calibrated on the PZT surface by a thermocoupler. The remanant polarizations of the three temperature points show similar, but low values in Figure 3.11(a). When comparing other films pyrolysed using a conventional furnace and tube furnace, it is about two times less, as

shown in Figure 3.11(b) and (d). The reason seems to be that the hot plate makes a higher temperature gradient than the other methods and the surface of the PZT layer is exposed to ambient air conditions. The dielectric constant of the PZT film pyrolysed using a hot plate showed the smallest value, seen in Figure 3.11(c).

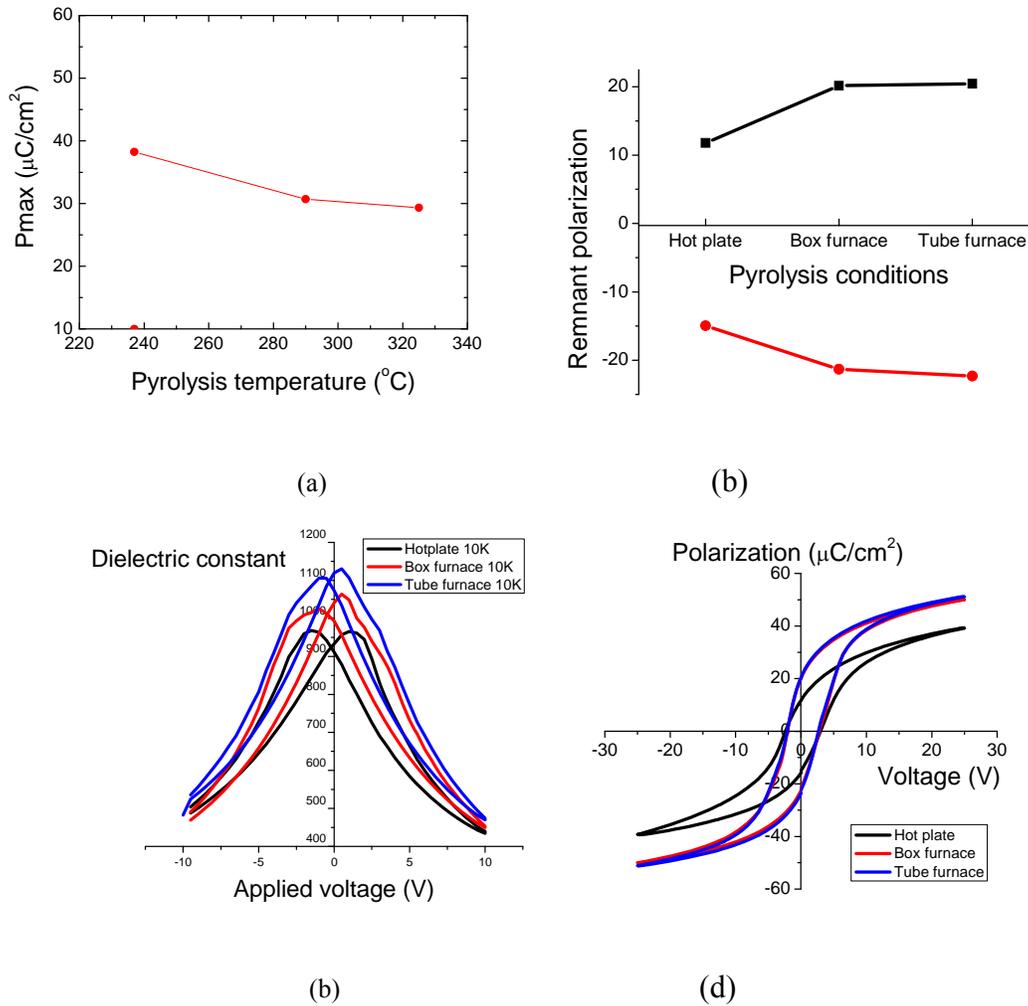


Figure 3.11 Electric properties of PZT films using different heat treatment methods: (a) as temperature on hotplate (b)  $P_r$ , (c) dielectric constant, and (d) PE loop as heat sources

While the strong PZT (100) was shown when the film was pyrolyzed by furnace, the film using a hotplate showed strong (111) and (110) orientations. As mentioned earlier, PZT (100)

has large remanant polarization and dielectric constant.

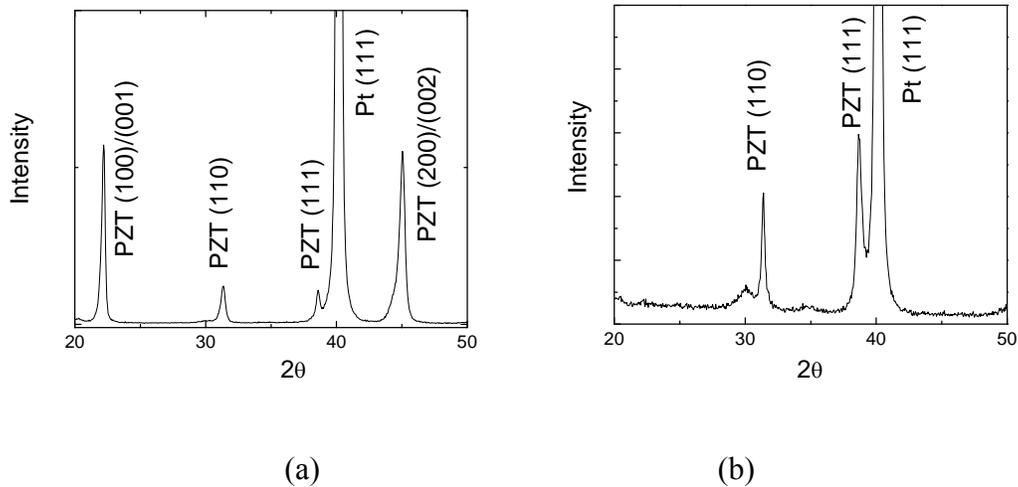


Figure 3.12 XRD analysis of PZT film pyrolysed using (a) furnace and (b) hotplate

### 3.3.4 PZT thickness effects

The piezoelectric properties of PZT film may be restricted due to the constraint on the substrate surface when the film is very thin. If the thickness of the film is large enough, the effect of the substrate can be negligible for the piezoelectric properties. While thick PZT film generates more power from piezoelectric energy harvesters, the thickness of PZT film is limited in film technology in MEMS devices by cost and quality. While many studies are working on increasing the PZT thickness [69, 74-76], in this study the thickness effect of around  $1\mu\text{m}$  was found not to lose the superior piezoelectric properties of PZT due to its small thickness. PZT films that have 500nm, 750nm, and 1000nm thicknesses were prepared by a CSD technique that repeated the 6,

9, and 12 times, respectively. As the thickness increased, the crystallographic orientation was not changed with strong PZT (111) peak, as seen in Figure 3.13.

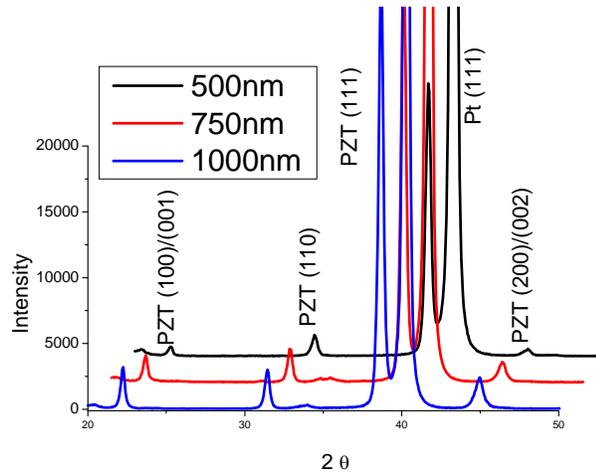


Figure 3.13 XRD analysis of PZT films with different thicknesses

Electrodes of  $0.25 \text{ mm}^2$  were patterned on the PZT film, and the PE hysteresis loop and dielectric constant were measured, as seen in Figure 3.14. Although the dielectric constant decreased a little, the remanant polarization was constant as the thickness decreased. As the thickness increased, the coercive field increased due the large distance between the electrodes. It seems that a further increase of PZT film does not improve its piezoelectric properties but increases the power by a large volume. PZT film of 500 nm still kept reasonable piezoelectric properties without an effect on substrate clamping.

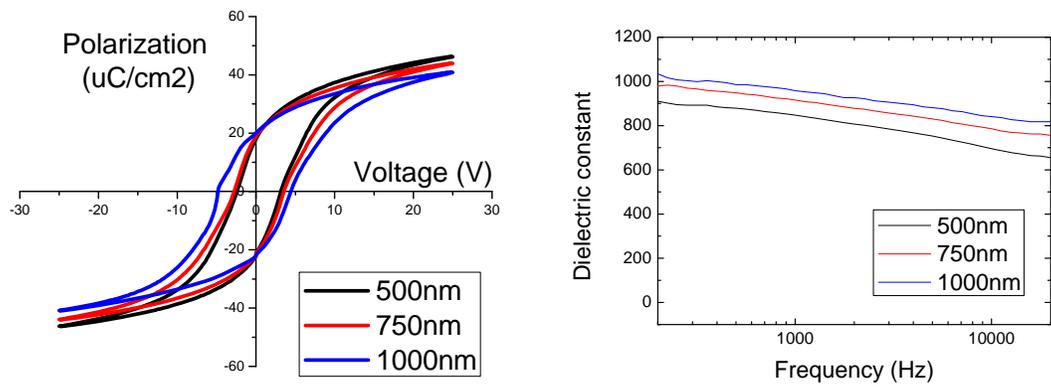


Figure 3.14 PE hysteresis loop and dielectric constant as frequency for the PZT films with different thicknesses

## CHAPTER 4 DESIGN AND FABRICATION OF MEMS ENERGY HARVESTER

### 4.1 Design

The structure of piezoelectric energy harvesters should be designed to create a higher strain on the piezoelectric transducer layer. The representative structures in MEMS are categorized into three configurations: diaphragm, bridge, and cantilever, as shown in Figure 4.1. The cantilever structure was chosen to create a high efficiency energy harvester in this study, since the freestanding structure with one-side clamping shows the highest strain under a seismic force among them.

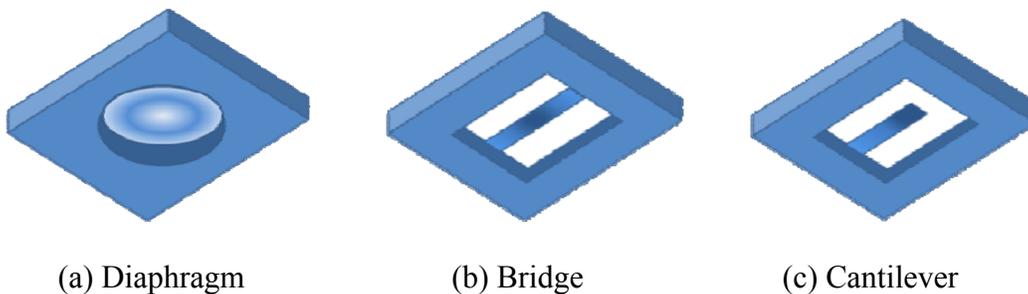


Figure 4.1 Illustration of common configurations of pMEMS structures

In this study, the cantilevers were designed with a silicon proof mass at the end of the tip and a silicon support layer under the beam, as shown in Figure 4.2. The silicon mass

puts a larger force and strain on the beam during seismic accelerating as well as decreasing the resonant frequency of the structure. The proof mass is easily built up by the dry etching process, while integration of the mass by aligning is more complex and costly [77, 78]. Very strong silicon-oxygen covalent bonds create a high mechanical strength between the beam and mass. The silicon support layer gives mechanical strength to the structure and enables the energy harvester to operate at a high accelerating force. The bending of the cantilever due to residual stress between films can be prevented by adding a Si support layer.

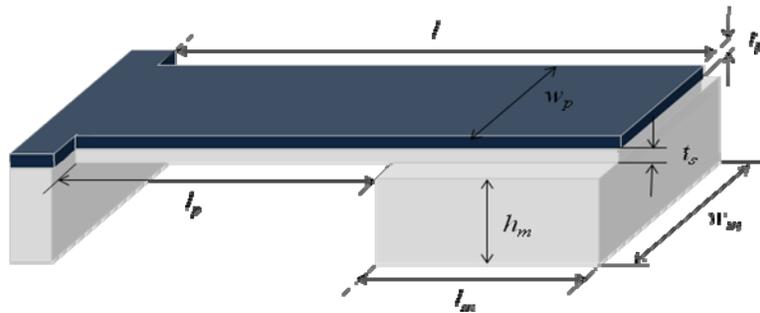


Figure 4.2 Schematic of cantilever energy harvester with proof mass and the notation of dimension

#### 4.1.1 Resonant frequency

The cantilever beam should be designed to achieve maximum output power at the expected resonant frequency of the applications. At the given film structure, the dimensions of the cantilever and mass dominantly decide the resonant frequency of the energy harvester. Since a small deviation from the resonant frequency induces a large reduction of output signals due to narrow resonant peak of MEMS devices, the energy harvester requires precise modeling to determine their dimensions for the expected resonant frequency.

The resonant frequency ( $f_r$ ) can be calculated by Equation 4.1[79]

$$f_r = \frac{\omega}{2\pi} = \frac{1}{2\pi} \sqrt{\frac{K}{m_e}} \quad (4.1)$$

where  $\omega$  is the angular frequency,  $K$  is the spring constant at the end-tip of the cantilever, and  $m_e$  is the effective mass of the cantilever. The resonant frequency of the cantilever without a proof mass can be expressed by Equation 4.2

$$f_n = \frac{v_n^2}{2\pi} \frac{1}{l^2} \sqrt{\frac{EI}{m'}} \quad (4.2)$$

where  $f_n$  is the  $n$ th mode resonant frequency;  $v_n$  is the  $n$ th mode eigenvalue;  $l$  is the length of the cantilever;  $E$  is the modulus of elasticity;  $I$  is the area moment of inertia about the neutral axis, and  $m'$  is the mass per unit length of the cantilever beam.. The equation 4.2 can be express in terms of bending modulus per unit width  $D_p$  by

$$f_n = \frac{v_n^2}{2\pi} \frac{1}{l^2} \sqrt{\frac{D_p}{m}} \quad (4.3)$$

$$m = \rho_p t_p + \rho_s t_s \quad (4.4)$$

where the mass per unit area  $m$  is calculated by the thicknesses and densities,  $\rho_p$  and  $\rho_s$  are the densities of the PZT and substrate, respectively, and the dimensions of the cantilever structure are denoted in Figure 4.2. The bending modulus ( $D_p$ ) is a function of Young's modulus and thickness and is expressed by [80]

$$D_p = \frac{E_p^2 t_p^4 + E_s^2 t_s^4 + 2E_p E_s t_p t_s (2t_p^2 + 2t_s^2 + 3t_p t_s)}{12(E_p t_p + E_s t_s)} \quad (4.5)$$

where  $E_p$  (PZT) and  $E_s$  (Si) are Young's moduli of two materials and their thicknesses,  $t_p$  and  $t_s$ , on the tip of the cantilever.

When the point proof mass is located at the end tip of cantilever and its size is smaller than the length of cantilever, the resonant frequency can be approximately expressed by [80]

$$f_n' = \frac{v_n'^2}{2\pi} \sqrt{\frac{K}{m_e + \Delta m}} \quad (4.6)$$

where  $v_n'^2 = v_n^2 \sqrt{3/0.236}$ , and  $K$  and  $m_e$  are the effective spring constant and effective mass, respectively. The effective mass ( $m_e$ ) is given by  $0.236mw_l$  considering axial velocity in length, but not in width when  $w \ll l$ . The effective spring constant ( $K$ ) can be expressed by

$$K = \frac{3D_p w_p}{l^3} \quad (4.7)$$

The Si proof mass is distributed through the large area rather than at the point at the end tip of the cantilever. When the center of the mass has a concentrated point load, its distance is  $l_m/2$  from the end tip. The effective spring constant at this location can be expressed by [81]

$$K' = K \left( \frac{l}{l - l_m/2} \right)^3 \quad (4.8)$$

The resonant frequency of a unimorph cantilever with a proof mass can be calculated by substituting the effective spring constant ( $K'$ ) for spring constant ( $K$ ) of Equation 4.6. It is expressed by

$$f_n = \frac{v_n^2}{2\pi} \sqrt{\frac{0.236w_p D_p (l - l_m/2)^3}{0.236mw_p l^7 + \Delta m l^3 (l - l_m/2)^3}} \quad (4.9)$$

where  $\Delta m = \rho_m l_m w_m h_m$ .

The low resonant frequency and large displacement of the cantilever beam is determined by the configuration of a long beam or a large mass and/or a softer beam. Hence, the ideal design for the piezoelectric energy harvester demands a high mechanical strength to the beam against the vibration and high density mass for its efficiency. Although the polymer based piezoelectric materials have flexibility, their low coupling coefficient could not satisfy the required output results [28].

The material constants of thin film should be carefully decided to reduce the deviation of resonant frequency due to different values in thin film scale and clamping effect between layers. The Young's modulus of PZT was measured by nano indenter and other values were surveyed, as shown in Table 4.1 [82]. The cantilevers are designed in one direction on a wafer as (110).

Table 4.1 The material constants

$E_p$	$E_{S(110)}$	$\rho_p$	$\rho_s$
160 GPa (m)	169 GPa (c)	7500 kg/m <sup>3</sup> (c)	3100 kg/m <sup>3</sup> (c)

(m): measured value, (c): calculated value

#### 4.1.2 Strain on PZT film & Safety factor

Along with modeling resonant frequency, the mechanical strength of the cantilevered beam was investigated during designing the energy harvester. As most artificial structures are built up with an extra safety factor against predicted maximum stress, MEMS devices also should be designed considering their safety factor. Most of the reported energy harvesters hardly consider their mechanical strength. This section gives the way to estimate the safety factor of cantilevered energy harvesters.

The mechanical strength of the cantilevered structure as the acceleration force was investigated. The mechanical strength of the cantilever can be expressed by a safety factor, and can be derived using Equations (4.11) to (4.15).

$$S = \frac{\sigma_{ys}}{\sigma_{max}} \quad (4.11)$$

$$\sigma_{max} = \frac{M(t_s/2 + t_p)}{I} \quad (4.12)$$

$$I = 2 \left( \frac{w_p t_p^3}{12} + w_p t_p b^2 \right) + \frac{1}{12} \frac{E_s}{E_p} w_p t_s^3 \quad (4.13)$$

$$b = \frac{t_p + t_s}{2} \quad (4.14)$$

$$M = \Delta m g \left( l_p + \frac{l_m}{2} \right) + m g \frac{(l_p + l_m)^2}{2} w_p \quad (4.15)$$

where  $\sigma_{ys}$  is the yield strength or fracture strength of the weakest material,  $\sigma_{max}$  the maximum stress at the bended surface (mostly anchor area),  $M$  the moment of the cantilever with a proof mass, and  $I$  the moment of inertia of the cantilever. The gravity point of the proof mass was considered as a point mass for simplification of analysis.

### 4.1.3 Design consideration for low resonant frequency

When the materials of the cantilever structure are fixed, the resonant frequency is dominantly decided by the configuration of size between the length and the mass as well as the thickness of beam, as shown in previous sections. While the low resonant frequency of the structure is very important for the energy harvester targeting a specific vibration source, its mechanical strength and the stress on the PZT transducer layer should be highly considered for

the energy harvester. The stress is closely related to the output voltage, which can be expressed by Equation 4.16 under open circuit condition.

$$V_{oc} = g_{31}\sigma_{xx}t_p \quad (4.16)$$

where  $g_{31}$  is piezoelectric voltage constant,  $\sigma_{xx}$  is stress in longitudinal direction, and  $t_p$  is thickness of PZT. The stress has a proportional relationship with the output voltage in piezoelectric cantilevered energy harvesters.

#### *Weight of mass effect*

The resonant frequency was plotted as a function of proof mass weight at various cantilever lengths in Figure 4.3 using Equation 4.9. All cantilevers have the same width of 2 mm and thickness of 21  $\mu\text{m}$  consisting Si and PZT layers. The width of mass is the same as that of the beam to prevent waste areas when the beam is narrower than the mass. When the beam and mass have the same width, the resonant frequency is not changed according to the width of the device. The weights of mass range from 1 to 9  $\mu\text{g}$  with the change of cantilever width.

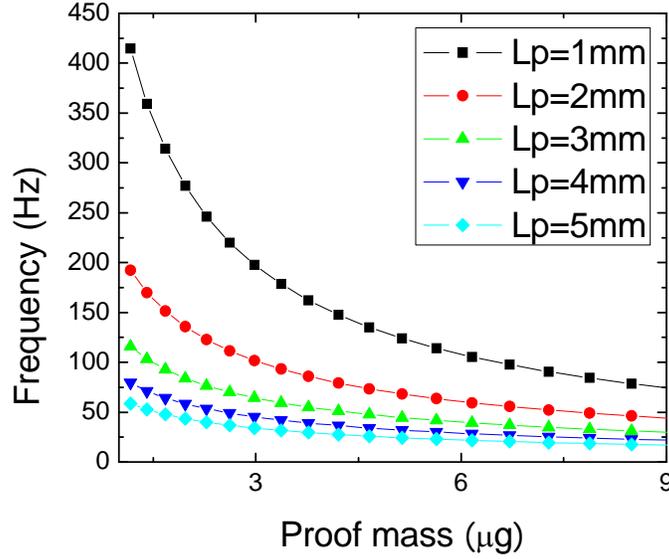


Figure 4.3 Resonant frequency vs. weight of proof mass in various lengths of cantilevers. The widths of the beam and mass are fixed at 2 mm

The change of resonant frequency as the configuration of the beam and mass change can be analyzed considering the size of the cantilevered structure, as seen in Figure 4.3. As the length of the cantilever is reduced, a heavier proof mass is required to obtain a low resonant frequency. While the resonant frequency of the 5 mm long cantilever does not change much by the weight of the Si mass, the 1mm length cantilever requires a much heavier mass. When the widths of beam and mass are fixed as the same, a relatively large mass is required for a smaller beam for the devices working in ambient vibration sources. Since the larger mass ( $\Delta m$ ) increases the displacement of the cantilever ( $M$ ) in Equation 4.15, the stress on the surface of the cantilever increases in Equation 4.12 and the safety factor of the structure decreases in Equation 4.11. When the larger mass obtains a low resonant frequency in the smaller size of the cantilever

structure, we should concentrate on the mechanical strength of the structure rather than the output voltages.

### *Beam thickness effect*

The thickness of the Si under layer was investigated in this section for its effects on the safety factor and the stress on the PZT layer. The low resonant frequency requires a thinner beam, and the resonant frequency of a 1 mm length cantilever is highly affected by the thickness of the cantilever comparing a longer cantilever, as seen in Figure 4.4. For example, 8  $\mu\text{m}$  of thickness is required for the 1 mm length cantilever compared to 35  $\mu\text{m}$  for the 5 mm length cantilever when 100 Hz of resonant frequency is targeted. As the size of the cantilever is smaller with constant mass size, a thinner beam is required to match the low resonant frequency.

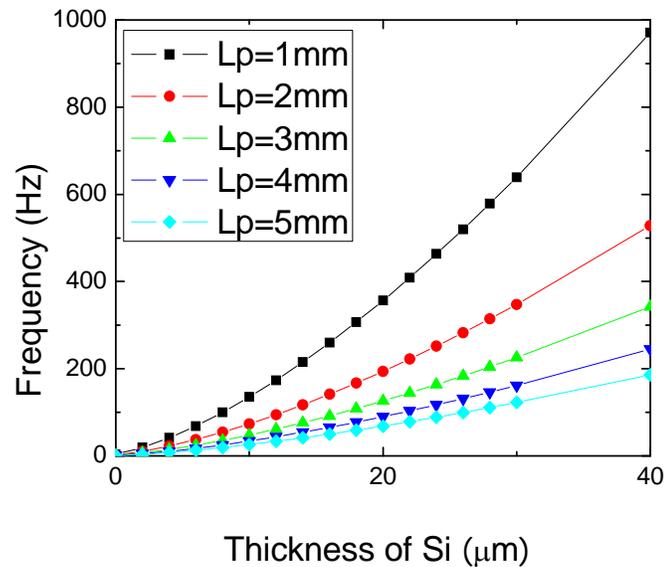


Figure 4.4 Resonant frequency vs. thickness of cantilever in various length of cantilevers whose widths is fixed at 2 mm

A smaller cantilevered structure with the same resonant frequency has a weaker mechanical strength than that of the larger cantilever. The safety factor can increase by increasing the silicon thickness of cantilevered beam, since it has a square relationship with the thickness of the silicon support layer in Equations 4.11 to 4.14 and Figure 4.5. If comparing the 1 mm length cantilever with 5 mm length cantilever, which have same resonant frequency of 100 Hz, the thin structure for the 1 mm length cantilever has a safety factor of more than three times less than that of the 5 mm length cantilever.

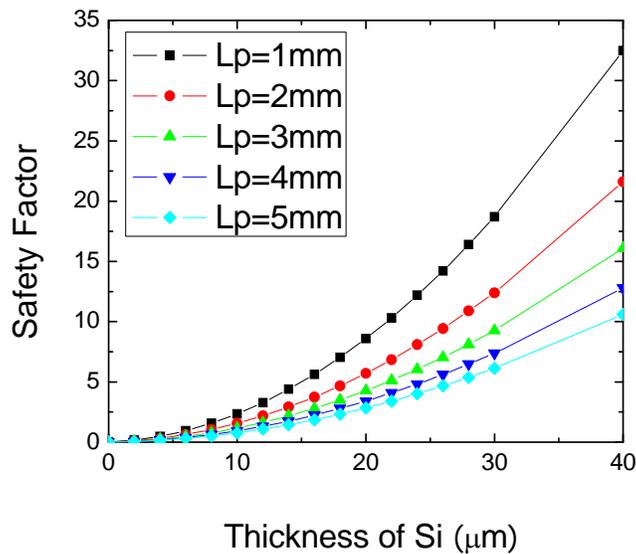


Figure 4.5 Safety factor vs. thickness of Si at different lengths of cantilevers. The width of the beam was fixed at 2 mm and the weight of the mass is constant

If the energy harvester is only designed to obtain a higher mechanical strength by increasing the thickness of the silicon support layer, we simultaneously lose the strain or the stress on PZT layer as shown in Figure 4.6. Beyond the yield point of materials, the stress or

strain dramatically decreases with the thickness of silicon support layer. Since the strain is directly converted to the output voltage by the piezoelectric PZT layer, the power density decreases with an increase in the safety factor at a given acceleration force.

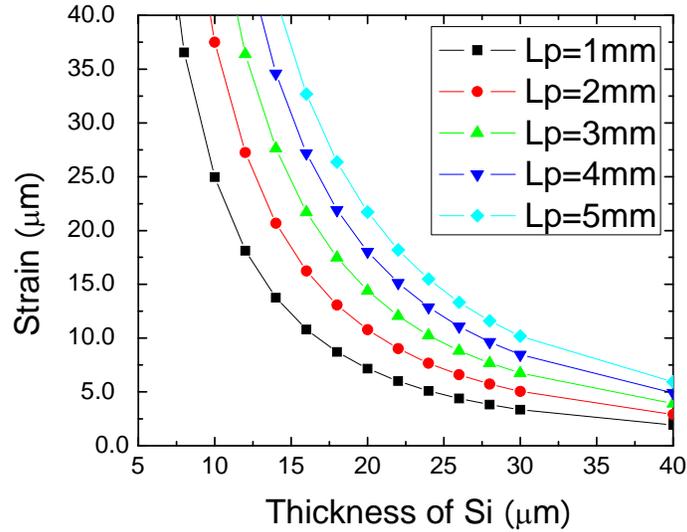


Figure 4.6 Average strain on PZT vs. thickness of Si at different lengths of cantilevers when the width of beam was fixed at 2 mm and the weight of mass is constant

Piezoelectric energy harvesters can be designed considering the required specifications of target environmental conditions using estimation by analytical modeling. The resonant frequency, safety factor, and output power density are estimated to determine the dimensions of the cantilever. The results show that the safety factor shows opposite trends compared with the resonant frequency design and output voltages. The cantilevered energy harvester requires a long beam, a heavy proof mass, and a thinner Si thickness to match the resonant frequencies of the energy harvester and ambient vibration source. In the configuration of the cantilevered structure for low resonant frequency, the safety factor decreased but the output voltages increased. For the smaller dimension, the mechanical strength is highly decreased while the output voltage

converted from the stress increases. The optimization is required to obtain both the mechanical strength and the output voltages with the resonant frequency in the range of ambient vibration source.

#### 4.1.4 Analytical output power

The characteristic behavior of piezoelectric energy harvesting can be described by damping constant and natural frequency. The lumped spring mass system is commonly used to characterize the dynamic behavior of energy harvesting. Figure 4.7 is simple diagram of a cantilever beam with a proof mass [2].

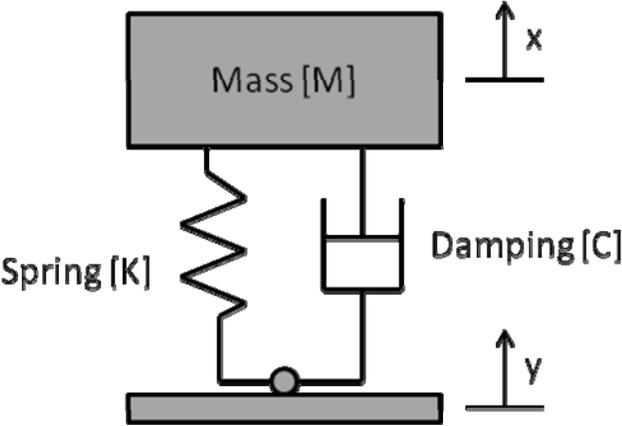


Figure 4.7 Equivalent lumped spring mass system

The governing equation of a lumped spring mass system can be expressed by

$$M\ddot{z} + C\dot{z} + Kz = -M\ddot{y} \tag{4.17}$$

where  $z$  is net displacement of mass ( $x - y$ ). The Laplace transform gives the time domain of the response,  $z(t)$ , by assuming the external base excitation is sinusoidal as  $y = Y \sin(\omega t)$

$$z(t) = \frac{\left(\frac{\omega}{\omega_n}\right)^2}{\sqrt{\left(1 - \left(\frac{\omega}{\omega_n}\right)^2\right)^2 + \left(2\zeta \frac{\omega}{\omega_n}\right)^2}} Y \sin g(\omega t - \phi) \quad (4.18)$$

where  $\omega$  is a mechanical angular frequency,  $\omega_n$  is a angular resonance. The approximate transduced power can be obtained from the product of velocity and force, and its maximum value is under  $\omega = \omega_n$

$$P_{\max} = \frac{mY^2 \omega_n^3}{4\zeta} \quad (4.19)$$

At the given input vibration magnitude and mechanical damping, it is necessary to maximize the mass and minimize the stiffness of the device. Du Toit derived and approximated Equation 2.17, and gives the simplified displacement (2.20) and output power (2.21) normalized by the input force ( $mY$ ) [83] as

$$\left| \frac{z}{(mY)^2} \right| = \frac{1}{K} \frac{\sqrt{1 + (\alpha\Omega)^2}}{\left[1 - (1 + 2\zeta_m \alpha)\Omega^2\right]^2 + \left[(2\zeta_m + (1 + k_e^2)\alpha)\Omega - \alpha\Omega^3\right]^2} \quad (4.20)$$

$$\left| \frac{P_{out}}{(mY)^2} \right| = \frac{\omega_n}{K} \frac{\alpha k_e^2 \Omega}{\left[1 - (1 + 2\zeta_m \alpha)\Omega^2\right]^2 + \left[(2\zeta_m + (1 + k_e^2)\alpha)\Omega - \alpha\Omega^3\right]^2} \quad (4.21)$$

$$\alpha = \omega_n R_l C_p$$

$$k_e^2 = \frac{k_{33}^2}{1 - k_{33}^2}$$

where  $R_l$  is load resistance,  $C_p$  is capacitance of PZT,  $k_{33}$  is the square root of electromechanical coupling coefficient, and  $\Omega = \omega/\omega_n$ .

The output power can be estimated where the piezoelectric coefficient of PZT,  $d$ , is  $3.2 \times 10^{-10}$  m/V, the dielectric constant of PZT,  $\epsilon$ , is  $3.36 \times 10^{-8}$  F/m, the damping ratio of the cantilever,  $\zeta$ , is 0.02, and the coupling coefficient of PZT,  $k$ , is 0.44.

## 4.2 MEMS Fabrication

The following chapter describes the technologies in MEMS fabrication, which led to the successful fabrication of an energy harvester by integrating piezoelectric thin films into a silicon-based microsystem. The schematic process flow with three masks for pMEMS cantilevers is shown in Figure 4.8. A silicon-on-insulator (SOI) wafer was used to give the silicon under layer a precise and uniform thickness.

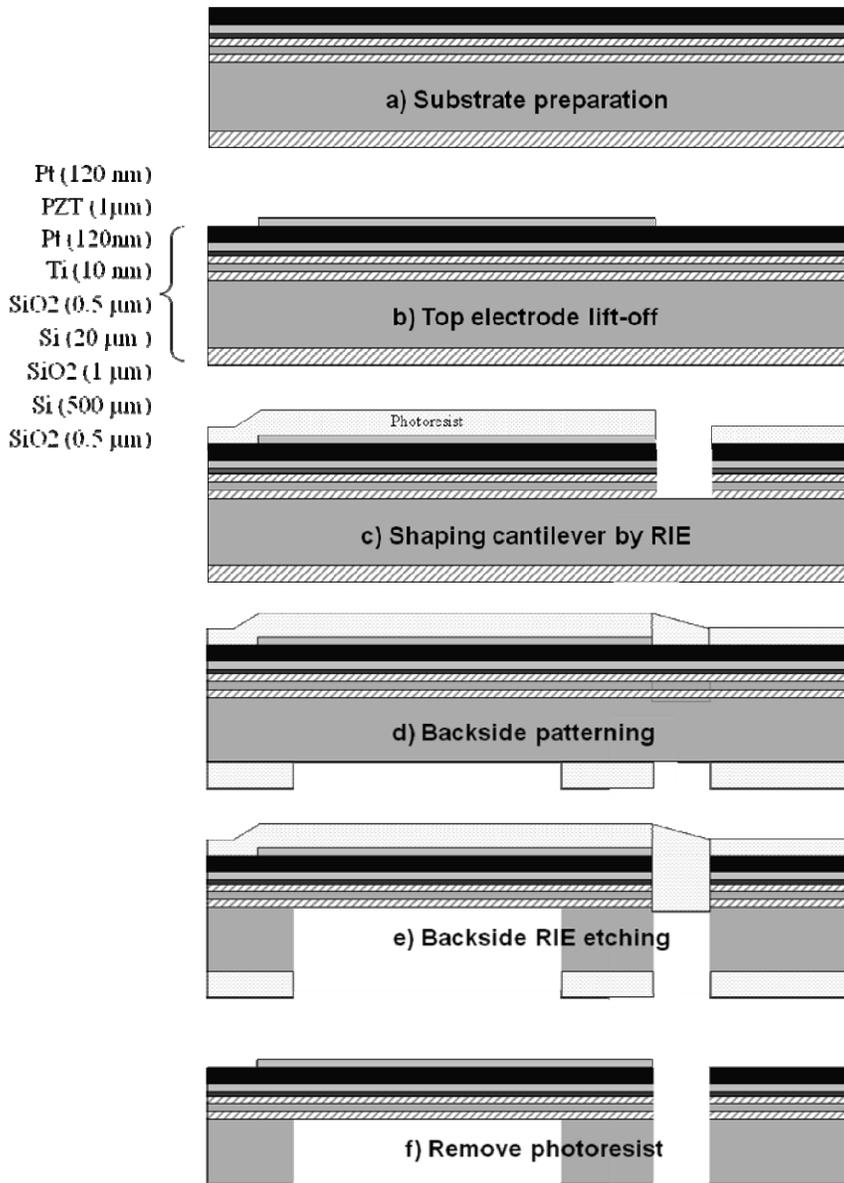


Figure 4.8 Process flow of piezoelectric MEMS energy harvester using SOI wafer.

The SOI wafer has 20 μm Si device layer and 1 μm buried SiO<sub>2</sub> layer. A 500 nm thick SiO<sub>2</sub> was grown on the wafer by thermal CVD, which is the electrical and diffusion buffer layer during the CSD process. Besides the diffusion and electric barrier, these buffer layers (SiO<sub>2</sub> & Si<sub>3</sub>N<sub>4</sub>) can be used to control the residual stress of a multi-layer structure. While other layers are limited in replacing their materials and changing their thicknesses, the buffer layers have

flexibility. It has been reported that SiO<sub>2</sub>, which was deposited by plasma enhanced chemical vapor deposition (PECVD), had higher compressive stress and prevented the bending of the cantilevers by compensating the tensile stress [46]. In this study, the thermal CVD process was chosen for the buffer layer due to equipment limitations of the facility. Sputter-deposited Pt was used as a bottom electrode due to its high electrical conductivity and thermal stability at high temperature. When a multi-layer structure experiences high stress, about 100 MPa, during CSD heat treatment, a delamination between Pt and SiO<sub>2</sub> may occur due to the difference in thermal expansion of constituent layers [84]. A Ti interfacial layer was used to enhance the adhesion between Pt and SiO<sub>2</sub>. After 12 layers of PZT were multicoated to obtain the thickness of about 1 μm, the Pt top electrode was patterned by a lift-off process. The detailed issues of each layer relating to the fabrication of pMEMS structure are summarized in Table 4.2.

Table 4.2 Specification and issues of individual layer in fabricating pMEMS energy harvesters

Layers	Materials (thickness) / dep. Tech.	Functions	Issues	Solution approach
Top electrode	Pt (120 nm) / Sputter	· Electrical contact	· Ohmic contact · Strong adhesion during MEMS process (PR stress)	· Deposition process control
Piezo-electric layer	PZT (1 μm) / Sol-gel	· Transfer strain energy to electrical energy	· Orientation · Stress	· Control substrate orientation with piezoelectric mode
Bottom electrode	Pt (120 nm) / sputter	· Electric contact · Control orientation of piez. Layer	· High stress is loaded during PZT annealing → delamination	· Adhesion layer · Orientation control by sputter parameters

Adhesion layer	Ti, Cr, or Ta (10 nm) / sputter	· additional adhesion between films	· Diffusion through the GB · Degradation of PZT layer	· Limited source → Very thin film
Membrane	SiO <sub>2</sub> (500 nm) / Thermal CVD	· Mechanical support · Buffer layer of diffusion · Etching stop	· Stress after cantilever release	· Stress design
	Si <sub>3</sub> N <sub>4</sub> (300nm) / LPCVD			
	Si (0~20 μm) / CZ	· Mechanical support · Stress compensation	· Uniformity · Thickness control during etching process	· Process (DRIE) engineering · Effective etching stop (SOI wafer)

After piezoelectric thin film is successfully deposited in wafer scale, the cantilever should be built up considering the precise dimension control and damages to the piezoelectric transducer layer. RIE is one of the dry etching methods that can provide cleanness, precise etching control, low chemical damage, an ability for mass production, and fast etching rate with physical and chemical reactions. SF<sub>6</sub> is more innocuous and shows a moderate etching rate on each film of pMEMS, although chlorinated gas has a faster etching rate on PZT due to highly volatility metal chloride. Process parameters are optimized by considering the etching rate, damages, and selectivity to the photoresist mask that serves as a protective layer.

#### 4.2.1 Lithography and mask for DRIE

A lithography process was used for patterning the top electrode and etching mask for the RIE system. The clearly defined vertical profile of the patterned photoresist is required for a precise

pattern of a Pt top electrode after lift off and etching profile at the edge of pattern for the etching mask.

Since a platinum top electrode was deposited by a sputter system, the incident angle of the accelerated atoms during deposition is not normal to the substrate. The angle of the accelerated atoms is created due to the collision through the pathway of atoms from target to substrate. If the cathode is not normal to the substrate in the sputter system, the angular deposition is more serious. As deposited film on the sidewall of photoresist gets thicker, it is hard to be removed by the lift-off process. If the lift-off process is performed by perforce, the top electrode can be torn away or remain as a scrap of sidewall film that induces the electric shortage of a piezoelectric cantilever corresponding to 1  $\mu\text{m}$  thin PZT layer.

A vertical or acute side wall angle of patterned photoresist improves the successful patterning of top electrode without any damage or electric shortage. Their sidewall has been shadowed with less area exposed to the cathode, and the probability of deposition of film on the sidewall is lessened. Negative photoresist (nLoF 5014) was used to obtain the vertical or acute angle. While a positive photoresist easily creates an obtuse angle due to over exposure or development, the negative photoresist can make an acute angle due to its inverse patterning. Figure 4.9 shows the vertical profile of the sidewall using negative photoresist. The delamination at the edge seems to have happened during cracking of the sample to see the cross section view.

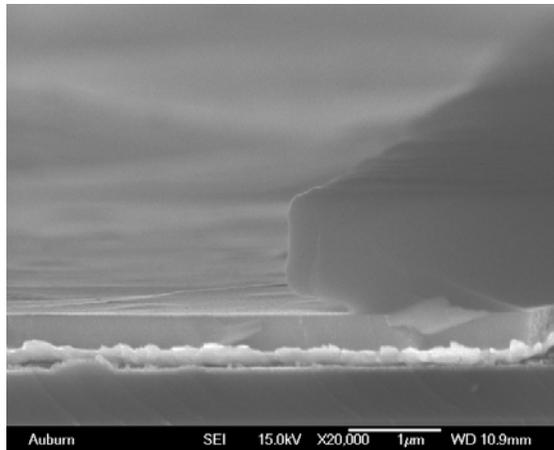


Figure 4.9 Edge of negative photoresist patterning.

Various polymer and metal film masks are used in the DRIE system for patterning. Metal masks have a high resistivity to the DRIE etching mechanisms, which have both chemical reactions and physical ion bombardment of ionized gas. In spite of the advantages of metal masks, the metal mask cause damage to PZT by creating internal stress due to the deposition of the metal mask. After removing the Al mask by an acid solution of phosphoric acid and acetic acid (PAE) etchant, damage on the top electrode was evident, as shown in Figure 4. 10. The top electrode was completely peeled off, and the bottom Pt was seen at the edge of cantilever where the PZT was peeled off. Conductive Pt was indicated by a brighter image than that of PZT layer in the SEM. The cracked Pt shows that internal stress existed, and the damages at the edge of cantilever seem to be caused from the internal stress or penetrating of etchant between PZT and bottom Pt.

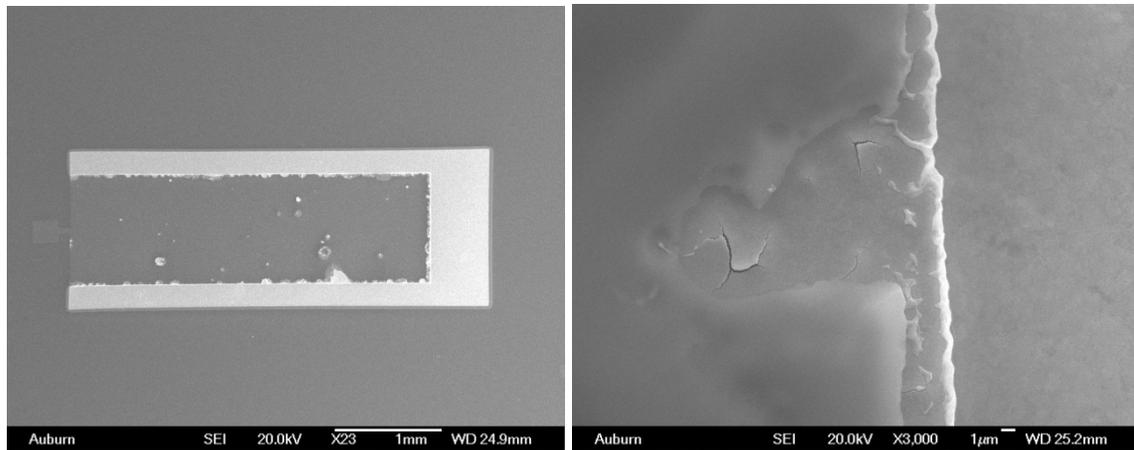


Figure 4.10 Damages on edge of Pt/PZT film after removing Al metal mask

The process parameters of the photoresist masks for lithography and RIE etching mask are summarized in Table 4.3. AZ4620 enables us to make thick masks for RIE etching from 5  $\mu\text{m}$  to 50  $\mu\text{m}$  depending on spin speed. When a low spin speed was used for thicker film, a sort of rehydration time was required for an hour to remove the bubbles in the masks. A spin speed of 4000 rpm makes 5  $\mu\text{m}$  thickness of AZ4620, and a good quality photoresist mask was obtained without any rehydration steps. Since the thicker mask results in larger stress on surface, a proper thickness of 10  $\mu\text{m}$  was chosen for etching the front side to prevent the Pt top electrode from peeling off. Two rounds of patterning were required to etch the thin films, including a 20  $\mu\text{m}$  silicon device layer. Since a total of 15  $\mu\text{m}$  was enough to etch the whole silicon wafer, three layers of AZ4620 were used for the back side etching mask.

Table 4.3 Parameters of lithography processes

Resist	Thickness per layer (μm)	Spin speed (rpm)	Prebake (°C/min)	Exposure time (sec)	Post bake (°C/ min)	Developing (min/developer)
nLOF 2031	2.5	3500	110/1	4.5	110/1	0.8 / AZ300
AZ 4620	5	4000	105/3	30	N/A	1 / A400K(1:3)

#### 4.2.2 Silicon bulk micromachining (Advanced Silicon Etcher)

Silicon bulk micromachining has been widely used for micro sensors and actuators. Etching silicon is categorized as the medium of etchant by wet and dry etching. Silicon etching from the wet phase mostly uses the mixture (HNA) of hydrofluoric acid (HF), nitric acid (HNO<sub>3</sub>) and acetic acid (CH<sub>3</sub>COOH), Tetramethyl-ammonium hydroxide (TMAH), or Alkali-OH (KOH) chemicals. While HNA shows an isotropic etching profile in three dimensions, others have anisotropic etching profiles depending on the crystallographic orientation. The dry etching for silicon is usually conducted using a vapor phase etchant or plasma with etchant gas. XeF<sub>2</sub> is mostly used for the vapor phase etchant of silicon, and its reaction is  $2XeF_2 + Si \rightarrow 2Xe(g) + SiF_4(g)$ . Plasma etching using reactive gas is the favorable method in micromachining the device due to fast etching rates and available process variables to control the etching profile. This study used plasma etching for bulk micromachining due to its excellent shaping ability, and detailed information is given in the next section.

When a cantilevered energy harvester is micromachined, its successful techniques need a uniform etching profile and precise thickness control on the silicon based structure to obtain the targeted resonant frequency. In this section, the basic information of plasma etching is introduced and the process is optimized based on its process mechanisms.

Figure 4.11 shows the schematic illustration of Inductive Coupled Plasma Deep Reactive Ion Etcher (ICP DRIE, STS). The inductive coil generates plasma by an electromagnetic field (13.5GHz), which ionizes the reactive gasses and densifies them by the magnetic field even at low pressure. The RF platen power is applied to the wafer to accelerate the ions to the negatively charged wafer. The positively charged ions are contributed to two etching mechanisms of chemical reaction and physical impact. The targeted material has chemical reactions with a reactive gas or gasses and is etched out by producing a volatile gaseous phase. The ion bombardments simultaneously knock out and etch out the target materials. During the RIE process, the physical impact remains the heat on the wafer. Although the generated heat is sunk by helium cooling, it should be also controlled by platen power not to burn the polymer mask. The burned photoresist mask is removed by saluting or O<sub>2</sub> plasma etching.

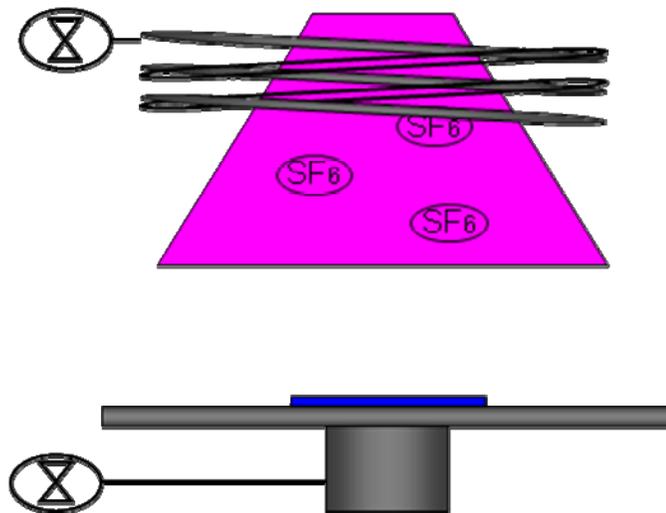


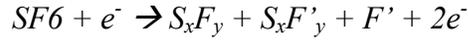
Figure 4.11 The illustration of DRIE

The process condition was systematically investigated by examining the variables and their effects as summarized in Table 4.4.

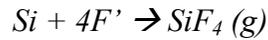
Table 4.4 The process variables and their effects on etching

Parameters	Effectiveness
Coil power	Plasma density
Platen power	Acceleration of ions
Gas flow and Chamber pressure	Plasma density Turbulence
Gas type	Chemical reaction
Etching and passivation cycle	Etching rate and surface profile

SF<sub>6</sub> and O<sub>2</sub> gasses were used in the etching process. The glow discharge in plasma generates free radicals of S<sub>x</sub>F'<sub>y</sub> and O':



while the free oxygen radicals create the passivation layer of an oxide film, and the fluorine is desorbed after the reaction with Si as a gas of SiF:



To obtain a vertical etching profile, silicon etching is performed by repeating passivation using C<sub>4</sub>F<sub>8</sub> for 7 seconds and etching using SF<sub>6</sub> and O<sub>2</sub> for 13 seconds as shown in Figure 4.12. After etching, the passivation layer is covered on all surfaces and walls. This etching effectively removes the surface due to the normal direction of accelerated ions, but not on the wall. This cyclic process makes the vertical angle of the side wall, and the ratio between etching and passivation determines the profile of the sidewall in microscopic and macroscopic forms. For example, the short cycle induces the smooth surface in the microscopic form and the very vertical sidewall in macroscopic form. The short cycle also results in a slow etching rate. The optimal cycle is required to balance the profile of the sidewall and etching rate.

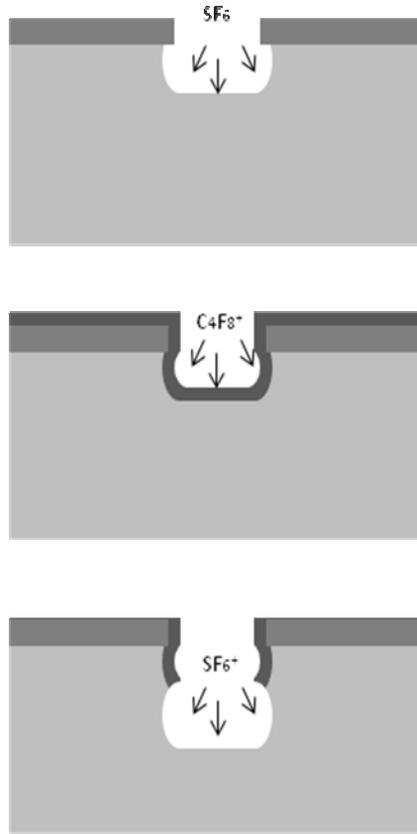


Figure 4.12 Cyclic etching and passivation of ASE process

Since the uniform etching profile mostly induces the slow etching rate, the optimized condition was investigated considering a moderate etching rate and uniformity. Plasma density, platen power of acceleration, chamber pressure, cycle of etching, and passivation were considered, and the recipe was built in Table 4.5.

Table 4.5 ASE recipe

Layers	Process engineering	Etching conditions
Si	Etching rate and profile are inversely affected by power, pressure, cycle of etching and passivation and gas flow rate	Etching: 600W(c), 12W(p), 130 sccm(SF <sub>6</sub> ), 13 sccm(O <sub>2</sub> ), 13 sec Passivation : 600W(c), 0W(p), C <sub>4</sub> F <sub>8</sub> 85 sccm, 7 sec, 37:1(s), 17 mTorr, 230 V bias

(c): Coil power, (p): Platen power, (s) selectivity to AZ 5214 photoresist, DRIE has helium cooling system

After etching the whole Si wafer's thickness, the proof mass was made with an undercut, and the sidewall angle was 85° as shown in Figure 4.13.

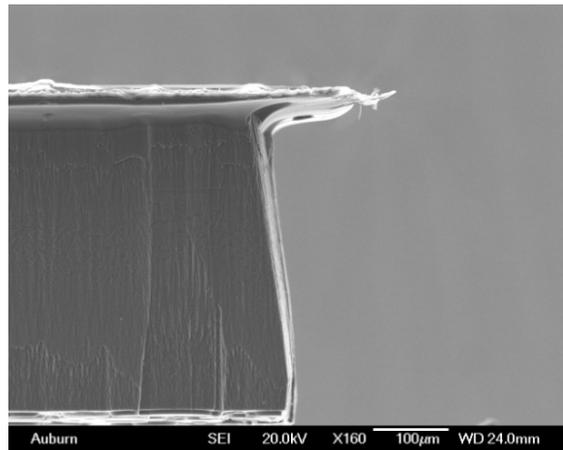


Figure 4.13 SEM image of side wall angle in silicon etching

Controlling the uniform etching profile and rate is required for the bottom of the trench. After etching 500 µm of thickness of wafer, the convex profile on the bottom appeared at the bottom trench whose window has 2 mm of width. It is reported that this convex profile disappeared at the narrow window size or changed to a concave profile [85]. But the reason for

this phenomenon is not well understood yet. It is suggested that the plasma density in the deep trench is not uniform so that the profile shows a difference as the window size varies. The chamber pressure was regulated to control the turbulence by the gas flow rate. The uniform depth profile was improved from 5% to 3% during the etching of the thickness ( $\sim 500 \mu\text{m}$ ) of the silicon wafer when using lower pressure conditions at the same gas flow. The uniformity was near a  $15 \mu\text{m}$  thickness variation and was created during etching out  $500 \mu\text{m}$  of silicon wafer by reducing the chamber pressure from 36 mTorr to 17 mTorr at the same gas flow, as shown in Figure 5.3. The uniformity was obtained by measuring the thickness of the cantilevers by an SEM image at the anchor, middle, and end tip positions.

The different sizes of the window made different etching rates during silicon etching, and this is called the lag effect [86, 87]. Although the widely opened areas have been etched out and the cantilever has been released, the narrow window beside the proof mass still remained, as shown in Figure 4.14. These results from etching rates that are different from the window sizes. The excess etching time to release the cantilever under the lag effect resulted in thinner cantilever beams whose window size is larger than that expected. Since the lag effect can be hardly solved by changing the etching conditions, it should be considered during designing the masks. After giving the same space as that between the window and the cantilever head, the lag disappeared.

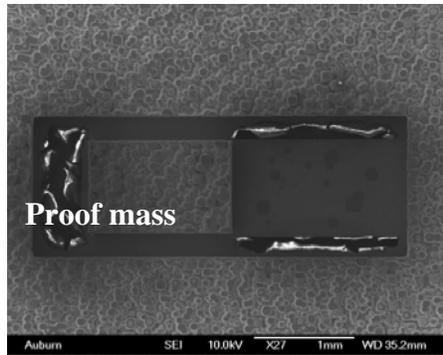


Figure 4.14 Different etching rate as window size. The side of mass was not released, while other area start to etch out

### 4.2.3 Surface micromachining (Advanced Oxide Etcher)

For etching PZT by plasma, various gasses and their mixture are widely used. The most favorable gasses are  $\text{CHF}_3$ ,  $\text{CF}_4$ , and  $\text{SF}_6$  combining with Ar or  $\text{C}_2\text{F}_6$ , and their optimal parameters as reported in the literature are shown in Table 4.6.

Table 4.6 The reported recipes of PZT plasma etching

Gas	Pressure (sccm, mTorr)	Power, bias	Mask	Etching rate (Film/Mask) (nm/min)	Reference
$\text{CHF}_3$	35, 15	60W, -	PR	8 / 4	[75]
$\text{CF}_4$	-, 7.5	1000W, 200V	PR	165/140	[76]
$\text{SF}_6$	-, 6	150W, 390V	Ni	300 / 8	[14]
Ar+O <sub>2</sub> (5:1)	-, 6	300V	PR	2.6	[77]
CF <sub>4</sub> +Ar (5:1)	-, 3	250V	PR	4	[77]
$\text{SF}_6$	20, 5	200W, 510V	Ni	120 / 4	[78]

Etching conditions for thin films using an Advanced Oxide Etcher (AOE) were

investigated using SF<sub>6</sub> gas by considering the etching rate, etching profile, and selectivity against photoresist mask. A literature survey showed that SF<sub>6</sub> gas can show a moderate etching rate without leaving remaining residues on the surface after a reaction. When the reactive gas is chosen, the process parameters should be optimized not to make damages on the films due to the physical energy of ion bombardments with considering etching rate.

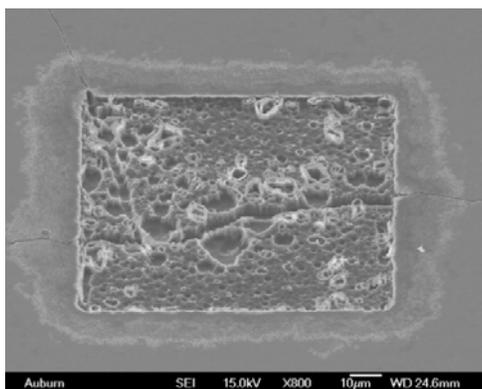


Figure 4.15 Cracks were created on PZT due to strong ion bombardment

Energetic ion bombardment induced from high platen power would result in burning photoresist that is hard to remove or decreasing the selectivity of the photoresist mask. Since the charges concentrate on the edge of the photoresist mask, the mask shows a faster etching rate at the outside of the pattern. The excess etching rate using a high platen power etched out photoresist mask from opened window and damaged the PZT films around the square window, as shown in Figure 4.15. Moreover, energetic ion bombardments can directly damage the PZT film [ ]. After etching the PZT using a high platen power of 300W, cracks were found on the PZT film and they continued to the area where the PZT film was already etched out in Figure 4.15. Since these cracks were found to be distributed around the other square etching windows of the same sample, it seems that the cracks were initiated from the etched region and were caused by

RIE etching. When the platen power was reduced to 200W and the chamber pressure was increased to provide sufficient reactive gas for etching, the etching rate increased 2.5 times but the cracks were still found with the fast etching rate. The power, chamber pressure, and gas flow were optimized with 70 nm/min of moderate etching rate and without damage on the films.

The process parameters for AOE were investigated and optimized for the moderate etching rate and to prevent damage on the PZT films as shown in Table 4.7.

Table 4.7 Investigation of process parameters of AOE system for etching PZT thin film

Coil power	Platen power	Pressure (mTorr)	Gas flow (sccm)	Bias (V)	Etching rate (nm/min)	Comments
800	300	5	20	250	40	-Moderate etching rate -Crack on PZT film
600	200	30	50	270	100	-Fast etching rate -Low selectivity -Little crack
600	150	20	30	230	70	- Moderate etching rate - No damage
600	100	5	20	90	10	- Slow etching rate

Photoresist is selected as the mask for the DRIE process due to its clean patterning and least amount of created stress as compared to the metallic films that can induce device failure during fabrication due to their high internal stress. SiO<sub>2</sub> and thin metal layers (Pt/Ti) were also effectively etched out by the same recipe for PZT etching, as seen in Table 4.8. The high

momentum of ion bombardment resulted in an acceptable etching rate for the very thin Pt/Ti/SiO<sub>2</sub> films.

PZT/Pt/Ti/SiO<sub>2</sub> layer could be vertically etched after 25 min of etching time, and its cross section view is shown in Figure 4.16. The detailed recipe for AOE is shown in Table 4.8.

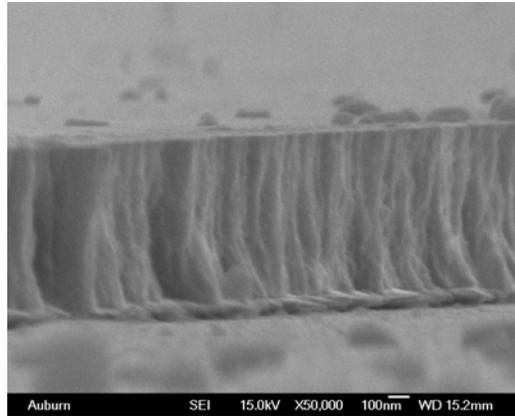


Figure 4.16 SEM image of etched PZT film surface in cross section view

Table 4.8 AOE recipe

Layers	Growth	Thickness	Selectivity	Etching rate	Etching conditions
SiO <sub>2</sub>	Wet oxidation	500 nm	1:1(s)	90 nm/min	
Ti interlayer	Magnetron sputtering	10 nm			600W(c), 150W(p), 30sccm(Ss)
Pt electrode	Magnetron sputtering	120 nm	1:30(s)	23 nm/min	
PZT 52/48	Sol-gel deposition	1 μm	1:10(s)	70 nm/min	

(c): Coil power, (p): Platen power, (s) selectivity to AZ 5214 photoresist, DRIE has helium cooling system

#### 4.2.4 Designing masks based on optimized pMEMS process

For successful fabrication of piezoelectric-based MEMS devices, two processes should be optimized. First, the integration of PZT on the Si wafer is a key issue of piezoelectric-based sensors, actuators, and energy harvesters. Although PZT has very high piezoelectric properties in bulk scale, the degradation of its properties might occur in film scale. The other is that the design and fabrication process were optimized for the successful energy harvester. A non-uniform etching profile during micromachining is a major issue in fabricating pMEMS energy harvesters. The micromachining might result in the degradation of piezoelectric properties, the shift of resonant frequency from the targeted one, and a decrease in the mechanical strength due to stress concentration on the weak part during resonating. This section shows how the designing masks have been developed considering the optimized film deposition and micromachining. The masks were designed by a LASI software tool.

The first mask in Figure 4.17 was designed by following concepts and resulted in the following:

1. The rectangular cantilever and bridge structures were included in this mask. The bridge structure withstood a very high acceleration, but the output signals were very low. The detail evaluation of the bridge structure do not apply in this study.
2. The main objectives of the first mask were to optimize the fabrication process and verify the resonant frequency. After etching out the whole Si layer, the Pt/PZT/Pt/Ti/SiO<sub>2</sub> multi layer was seriously bent up due to internal stress as predicted. The bending curvature was so large over 90 degrees that the cantilever structure almost failed without the silicon

substrate. The devices were also easily broken down during carriage.

3. Besides compensating for the internal stress by applying thin oxide films and/or controlling their stress by a growth method, the remaining Si under layer was chosen due to easy fabrication and because it supplies high mechanical strength. The bending configurations of the cantilever structure as Si thickness will be introduced in Chapter 5. The thickness of the Si layer could be controlled by adjusting the etching parameters of a DRIE Si etching system.
4. The conventional lag effect was found during releasing cantilever by Si etching. The different window size induced the non-uniform etching rate. Since the lag effect required additional etching time to complete releasing of the cantilevers, it was hard to control the accurate Si thickness only by etching time.

After examining the feasibility of the fabrication process using the first mask, the second mask was designed considering the Si membrane.

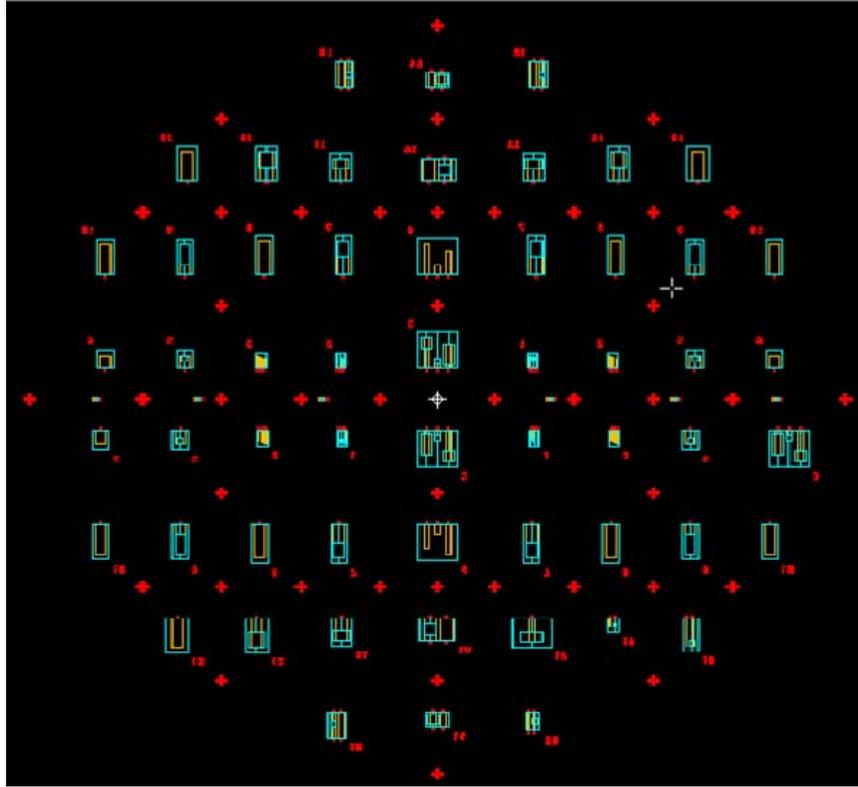


Figure 4.17 1<sup>st</sup> generation photolithographic mask for prototype fabrication

The second mask is shown in Figure 4.18, and it was designed considering the following issues and resulted in:

1. The Si under layer whose goal thickness was 20  $\mu\text{m}$ .
2. The gap between the cantilever and the outside window was unified to prevent the lag effect of the DRIE system, and the cantilever could be released without lag around the side and head of proof mass. The thickness of the Si under layer supports the PZT active layer and can be controlled without the lag.
3. The electrode size was enlarged to enable hand-wiring, and the bottom electrode was added for clear contact.

4. The top electrode size was reduced comparing PZT layer to prevent possible electric shortage between the top and bottom Pt electrode after fabricating. The electric shortage could be prevented by reducing the size of the top electrode by 15  $\mu\text{m}$ .
5. The prototype of the cantilever array was tested. The non-uniformity of the device dimensions in wafer scale might exist from the center to edge or right to left due to the special density of plasma and turbulence of reactive gases. The discrepancy of resonant frequency of each cantilever due to non-uniform etching rate will be introduced in Chapter 5.
6. The non-uniform etching profiles on the side wall of the proof mass and the surface of Si under layer were found by SEM analysis. Such a non-uniform etching profiles induced the shift of resonant frequency from calculated values. The process was optimized to improve etching profiles, but the RIE process has a limitation to maintain the sufficient etching rate. The etching top layer was required for the uniform profile to obtain a goal resonant frequency.
7. The cantilevers have resonant frequencies under 100 Hz. The dimensions of the cantilever were varied to examine the width, length, and weight of the proof mass. The detailed dimensions are denoted in Table 4.9 and Table 4.10. The notations in the Tables are based on Figure 4.2.

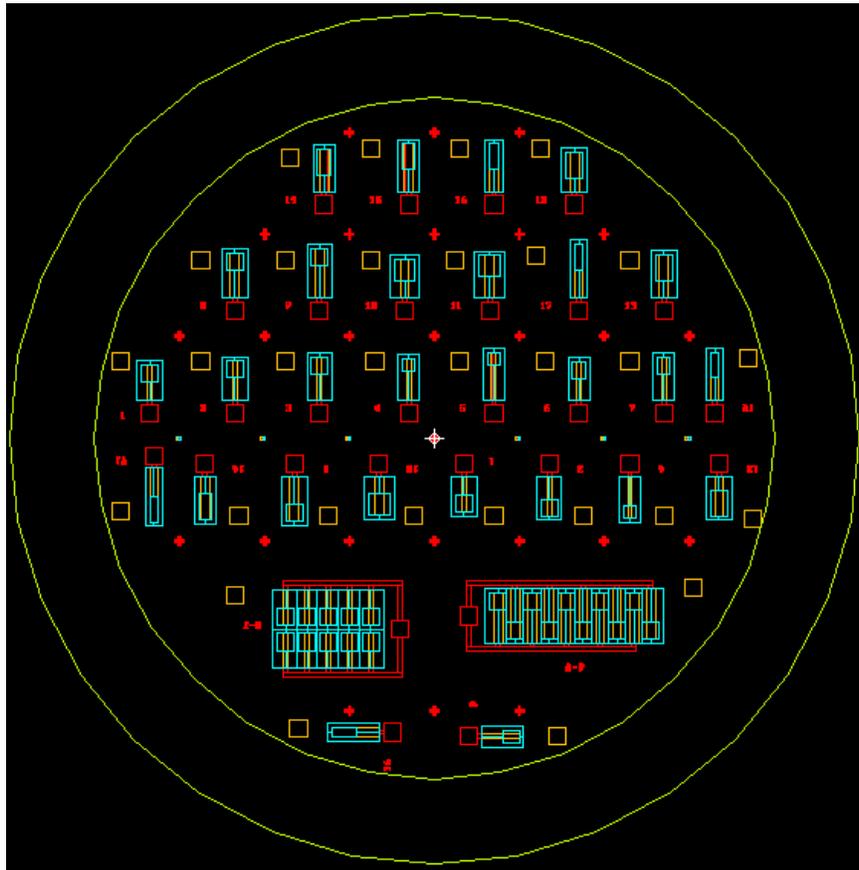


Figure 4.18 2<sup>nd</sup> generation photolithographic mask for prototype fabrication

Table 4.9 Dimension and resonant frequency of single cantilever in 2<sup>nd</sup> mask

$W_p$	$L_p$	$L$	$W_m$	$H_m$	$L_m$	$Fr$	#
500	2100	4100	2000	500	2000	99	1
	2600	4600				83	2
	3100	5100				71	3
	3400	4900	1500	500	1500	101	4
	4100	5600				82	5
	2500	4500	1500	500	2000	99	6
	3100	5100				82	7
1000	3200	5200	2000	500	2000	98	8
	3800	5800				83	9
	2000	4500	2500	500	2500	97	10
	2500	5000				83	11
	1600	4600	2000	500	3000	96	12
	2100	5100				82	13
	2000	5000	1500	500	3000	97	14
	2600	5600				82	15
	2600	5600	1000	500	3000	100	16
	3300	6300				83	17

Table 4.10 Dimension and resonant frequency of cantilever array in 2<sup>nd</sup> mask

$W_p$	$L_p$	$L$	$W_m$	$H_m$	$L_m$	$Fr$	#
500	2100	4100	2000	500	2000	99	1
1000	2500	5000	2500	500	2500	83	11

The third mask is shown in Figure 4.19, and it was designed and resulted as follows:

1. The mask was designed to test various shapes of cantilever structures such as circularly filleted notch and trapezoid.
2. A SOI wafer with 20  $\mu\text{m}$  Si device layer was used for a uniform Si under layer, and to verify the modeling of the resonant frequency.
3. The design radically located the devices from the center of the wafer to investigate the uniform etching rate in the wafer scale. The highly dense location of the cantilever was enabled when using the dicing line of the silicon etching mask, and a high product yield might be possible. The failure during the dicing device also might also be prevented

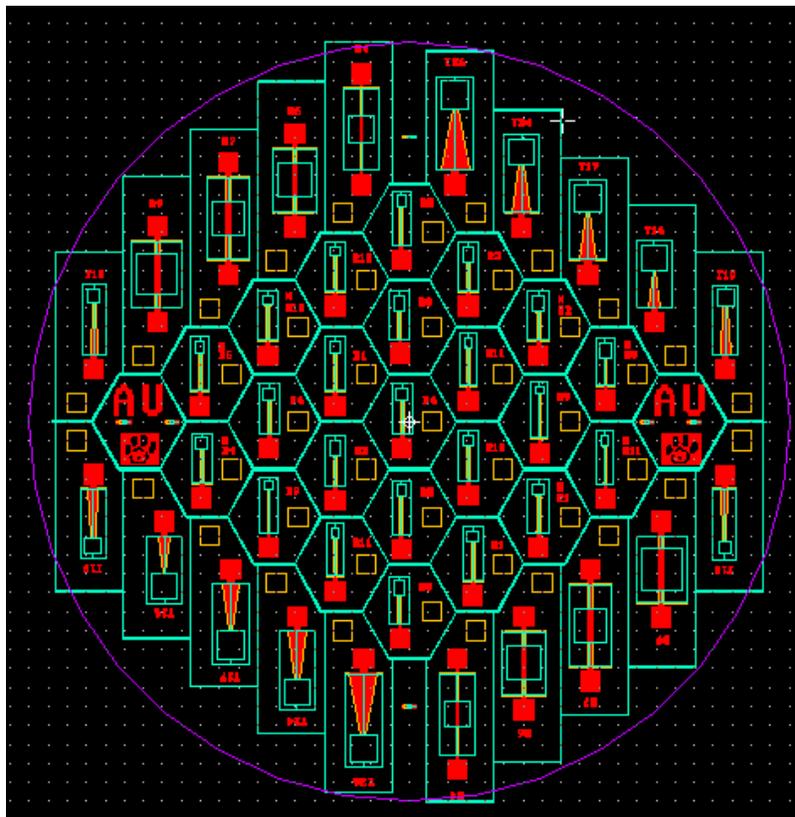


Figure 4.19 3<sup>rd</sup> generation photolithographic mask for prototype fabrication

The cantilever array was tested in parallel connection for high current yield. The cantilever array was located in radical for uniform etching. These prototypes were relatively large scale to observe the feasibility of the cantilever array. The fabricated prototypes are pictured in Figure 4.20, and the mask is shown in Figure 4.21. The SiO<sub>2</sub> layer can be seen on the backside of the cantilever, and the reflected color shows the non-uniform thickness of the SiO<sub>2</sub> after ASE etching. The different between the maximum and minimum thickness is very small, under 300 nm, when inspecting the color chart of SiO<sub>2</sub>.

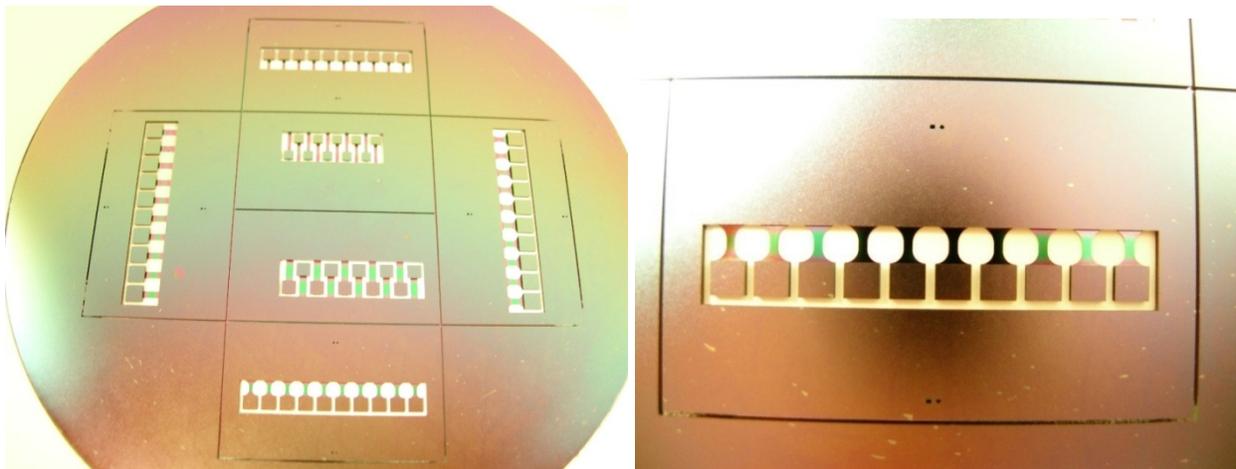


Figure 4.20 prototype cantilever array

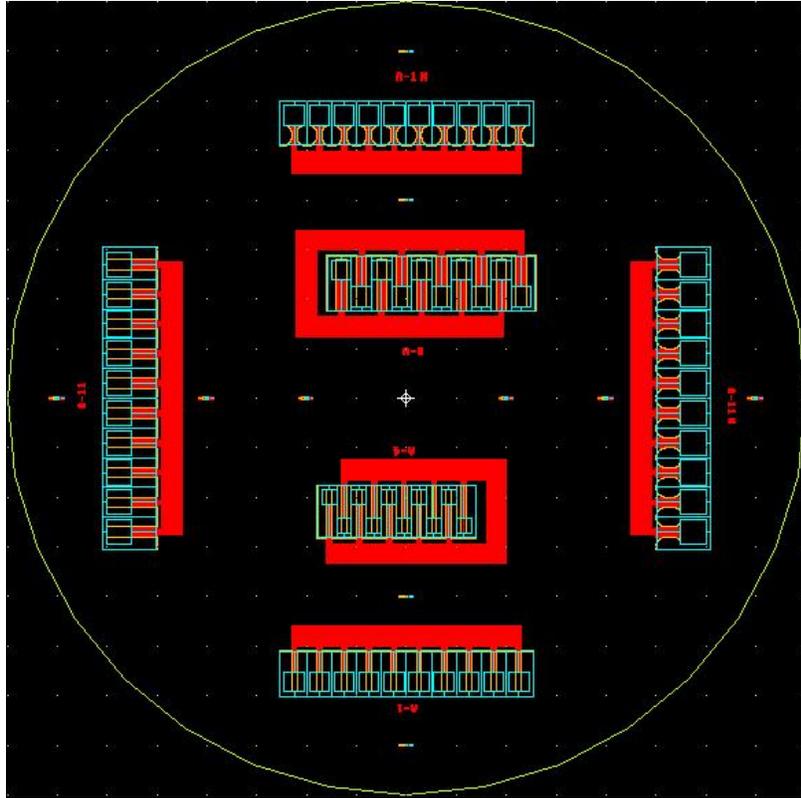


Figure 4.21 photolithographic mask for cantilever array prototype

The fourth mask (Figure 4.21) was designed to study the modeling of resonant frequency, interdigitated electrode, and serially connected array, mechanical strength, and geometry stress on the cantilever. The detailed dimensions as a function of device are denoted in Table 4.11.

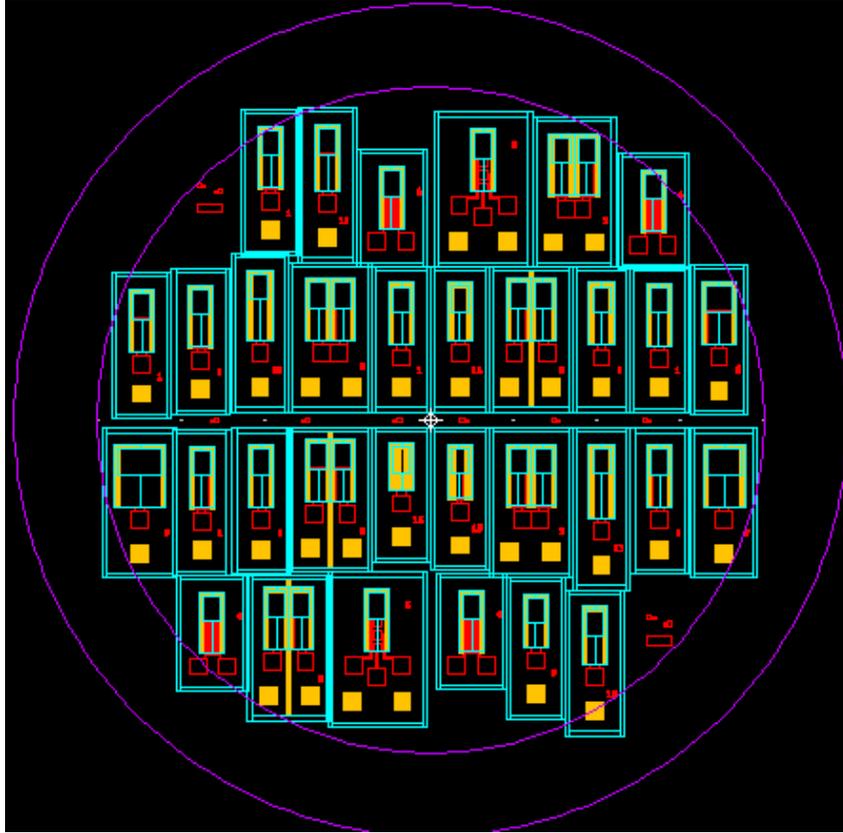


Figure 4.22 4<sup>th</sup> generation photolithographic mask for prototype fabrication

Table 4.11 Dimension of devices on 4<sup>th</sup> mask

<i>Comments</i>	<i>Cal fr (Hz)</i>	<i>Safety factor</i>	<i>Lp (m)</i>	<i>Wp (m)</i>	<i>Wm (m)</i>	<i>Lm(m)</i>
Array (S & P)	125.58	7.90	4.00E-03	2.00E-03	2.00E-03	3.00E-03
d <sub>33</sub>	125.58	7.90	4.00E-03	2.00E-03	2.00E-03	3.00E-03
Geometry stress test	125.58	7.90	4.00E-03	2.00E-03	2.00E-03	3.00E-03
Width check	125.58	7.90	4.00E-03	2.00E-03	2.00E-03	3.00E-03
	125.58	7.90	4.00E-03	3.00E-03	3.00E-03	3.00E-03
	125.58	7.90	4.00E-03	4.00E-03	4.00E-03	3.00E-03
	125.58	7.90	4.00E-03	5.00E-03	5.00E-03	3.00E-03

Frequency check	169.98	9.72	3.00E-03	2.00E-03	2.00E-03	3.00E-03
	145.01	8.72	3.50E-03	2.00E-03	2.00E-03	3.00E-03
	125.58	7.90	4.00E-03	2.00E-03	2.00E-03	3.00E-03
	110.12	7.22	4.50E-03	2.00E-03	2.00E-03	3.00E-03
	97.57	6.64	5.00E-03	2.00E-03	2.00E-03	3.00E-03
	87.23	6.14	5.50E-03	2.00E-03	2.00E-03	3.00E-03
Length/Width Ratio	125.58	7.90	4.00E-03	2.00E-03	2.00E-03	3.00E-03
	125.45	6.61	3.52E-03	1.50E-03	2.00E-03	3.00E-03
	125.86	5.37	3.10E-03	1.00E-03	2.00E-03	3.00E-03
	125.59	3.42	2.17E-03	5.00E-04	2.00E-03	3.00E-03

The fifth mask in Figure 4.23 was designed as follows:

1. To investigate the size effect, poling effect, aging and fatigue effect, and geometry stress test as cantilever shapes.
2. The trapezoid might alter the stress distribution and increase the output power. The trapezoidal cantilever in the third mask showed no improvement of output power. The stress distribution will be examined by three different electrodes located on the near anchor, middle of cantilever, and near proof mass. The output signals from three electrodes will be used to analyze stress distribution. Signals from the rectangular cantilever with the same electrode configuration will be compared and analyzed.

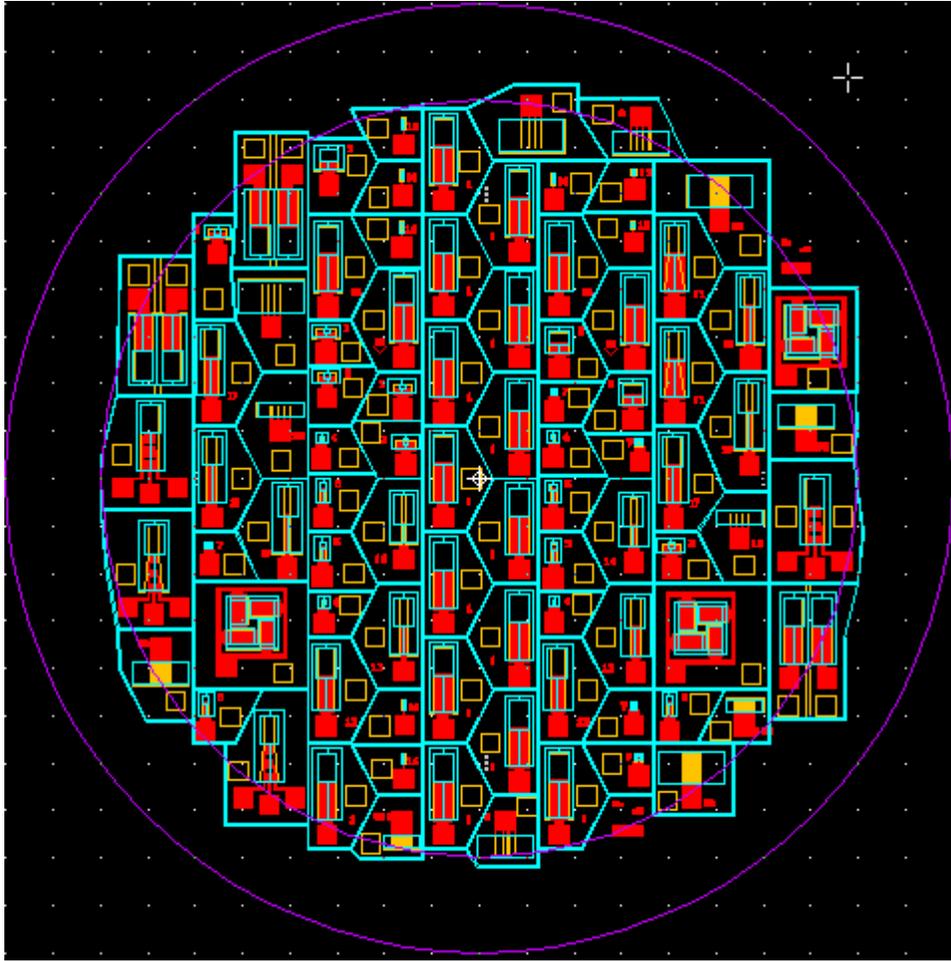


Figure 4.23 5<sup>th</sup> generation photolithographic mask for prototype fabrication

## CHAPTER 5 DEVICE CHARACTERIZATION AND ANALYSIS

### 5.1 Experimental setup for characterization

The output signals from the energy harvester mounted on an electromagnetic shaker (Labworks In., ET-132-203) were measured by an oscilloscope (Tektronix, TDS3014B). The sinusoidal vibration of the shaker was controlled by an amplified source from an external function generator (Agilent 33220A). The generated vibration was simultaneously monitored by an accelerometer (PCB piezoelectronics Inc., 352C65) whose signal was monitored by the oscilloscope. The output voltage from the device was measured across the resistor, which could be regulated to find the optimal resistance load. The peak-to-peak output power was used to calculate the power, and the recorded output wave was used for the AC-DC rectifier. The schematic experimental setup is shown in Figure 5.1.

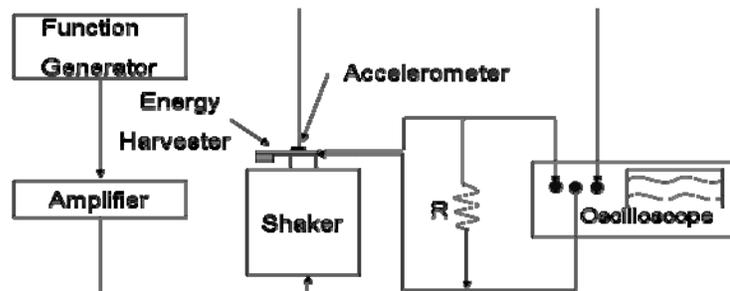


Figure 5.1 Schematic of experimental setup for energy harvester

## 5.2 Performance of single cantilever

The characterizations of the energy harvester are categorized to investigate PZT film properties, resonant frequency, and output signal during resonance conditions. The results of film properties have already been discussed in Chapter 4. The PZT films had similar polarization values ( $P_r$  and  $P_{max}$ ) before and after been fabrication process. A minor distortion of been hysteresis loop occurred, and it seemed to be caused from the stress of the released cantilevers after completing the fabrication and/or defects due to the diffusion of substrate materials. The resonant frequency was measured by an impedance analyzer and the maximum output voltage was tracked via an oscilloscope. The power was calculated from the output voltage by Ohm's law. To rectify the AC generated from the energy harvester to DC for storage, the  $V_{rms}$  value was used for calculations. In this chapter, the resonant frequency of the devices and output signal will be analyzed and discussed.

### 5.2.1 Device on silicon wafer

The internal stress makes the cantilever bend upward due to the net tensile stress between the Pt/PZT/Pt/SiO<sub>2</sub> layers. Although the net stress can be controlled by the proper material selection of a buffer layer (SiO<sub>2</sub> or SiN<sub>x</sub>) with a different deposition method, a thicker Si layer was chosen to prevent bending as well as to enhance the mechanical strength of the device. Without a Si support layer, the Pt/PZT/Pt/SiO<sub>2</sub> was seriously bent and failed, as shown in Figure 5.2 (c). The internal stress was experimentally examined as the thickness of the Si under layer. At 100  $\mu\text{m}$  of Si thickness, the cantilever showed no bending due to stiff support in Figure

5.2 (a). The bending of the cantilever was not found up to 20  $\mu\text{m}$  of Si thickness, but it started to bend upward due to tensile residual stress from 10  $\mu\text{m}$  of Si thickness.

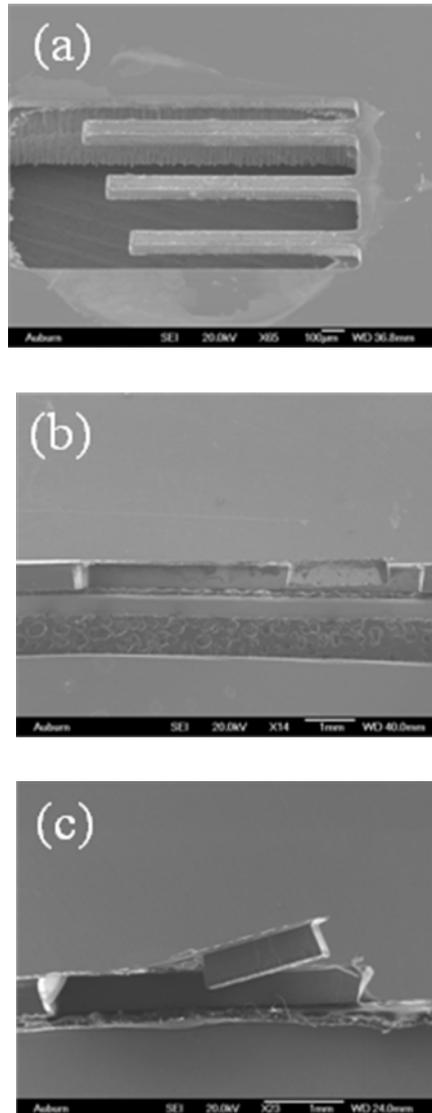
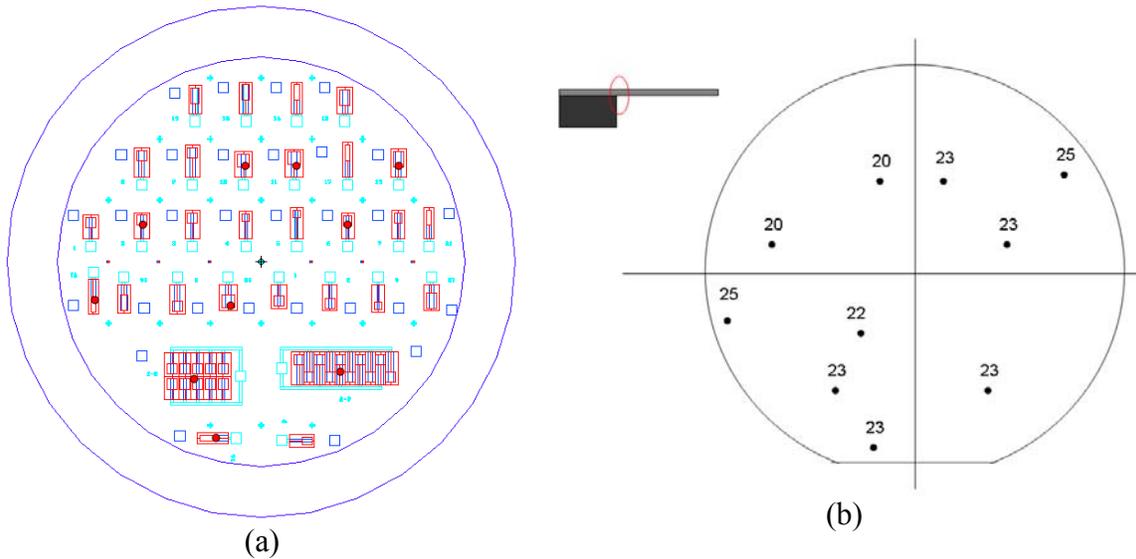


Figure 5.2 Cantilevers have different average thicknesses on Si under layer as (a) 100  $\mu\text{m}$ , (b) 20  $\mu\text{m}$ , and (c) 10  $\mu\text{m}$  (cantilever was bend due to internal stress).

The resonant frequency is sensitive to dimensions, especially the thickness. The first attempt was controlling the Si thickness by a DRIE etching process. The current recipe etches the

silicon with about 1  $\mu\text{m}$  per 1 cycle (20sec) of etching rate. A 20  $\mu\text{m}$  Si under layer was expected by the DRIE etching process. The process time-based thickness control was reasonably capable when the silicon support layer remained at an average of 27  $\mu\text{m}$  of silicon thickness during 550  $\mu\text{m}$  etching. The problem was very poor uniformity of the cantilever, as shown in Figure 5.2. The Si layer showed a convex profile with a relative thinner region on the near anchor and proof mass. It is not clearly understood why the convex profile occurs at the wide window size of the pattern, while the narrow window shows a concave profile. It seems that the bias in the trench is locally different [88]. It was also suggested that abnormal gas conditions in the deep trench with a wide window made non-uniform etching configuration geometrically [89].



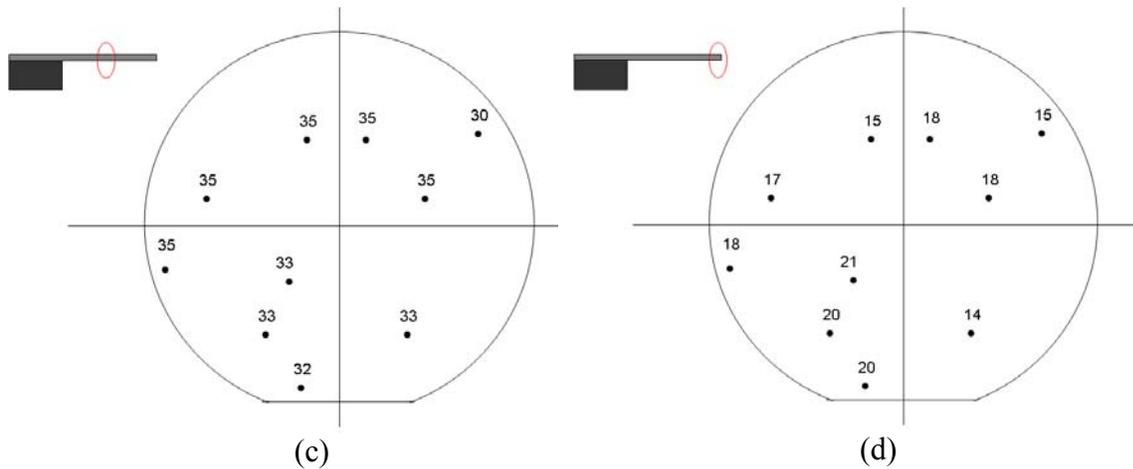


Figure 5.3 Geometric graph of cantilever's thickness: (a) Sampled cantilevers and their location, (b) Thickness at the joint on proof mass, (c) Thickness at the middle of cantilever, (d) Thickness at the joint on anchor

Based on a literature survey and experiments, the uniformity can be controlled to set a short etching and passivation cycle, low power, and high chamber pressure [88-90]. But their adjustments by optimizing process conditions lead to a very slow etching rate. The uniformity was investigated in wafer scale from the center to the edge. Three positions, near anchor, middle, and end tip, of the cantilevers were measured for their thickness. The uniformity of the cantilever beam varies according to the window size, but does not depend on the location in the wafer scale. On the contrary, the etching rate varies according to the location in the wafer. The center of the wafer generally shows a higher etching rate compared to that of the edge. Since the calculated resonant frequency assumes a uniform beam thickness, the measured values show a large deviation. The measured resonant frequencies were in a range between resonant frequencies calculated using maximum and minimum thickness. Since the piezoelectric energy harvester needs precise frequency matching with the target application, an unexpected shift of resonant frequency caused from non-uniform thickness led to functional failure.

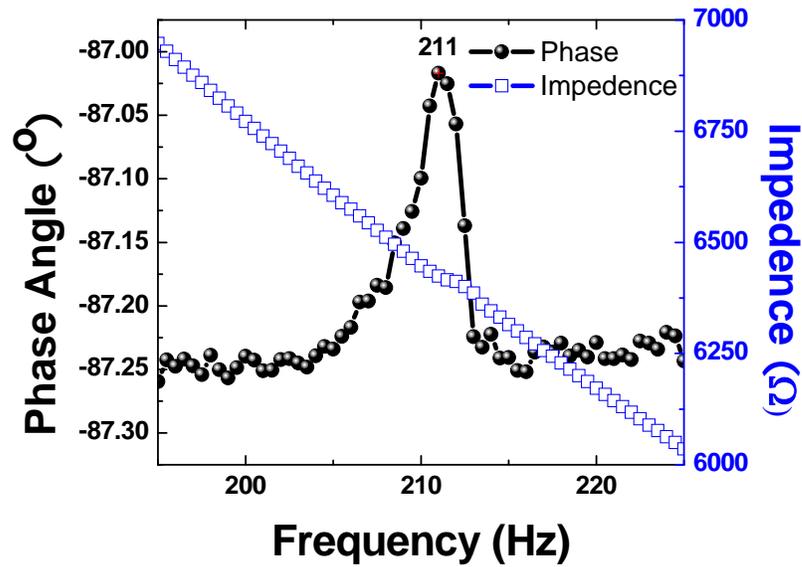


Figure 5.4 The resonant frequency of cantilever fabricated on Si wafer

A cantilever that has 5mm(L) x 1mm(W) of beam dimensions and 2.5mm(L) x 2.5mm(W) of mass dimensions has 162 Hz of calculated resonant frequency. The measured resonant frequency was 211 Hz as shown in Figure 5.4. The cantilever has 15  $\mu\text{m}$ , 35  $\mu\text{m}$ , and 20  $\mu\text{m}$  of thickness near the anchor, middle, and end tip. A 211 Hz resonant frequency estimates 24  $\mu\text{m}$  of uniform silicon thickness. A 1  $\mu\text{m}$  thickness deviation makes a shift of resonant frequency of about 10 Hz as shown in Figure 5.5.

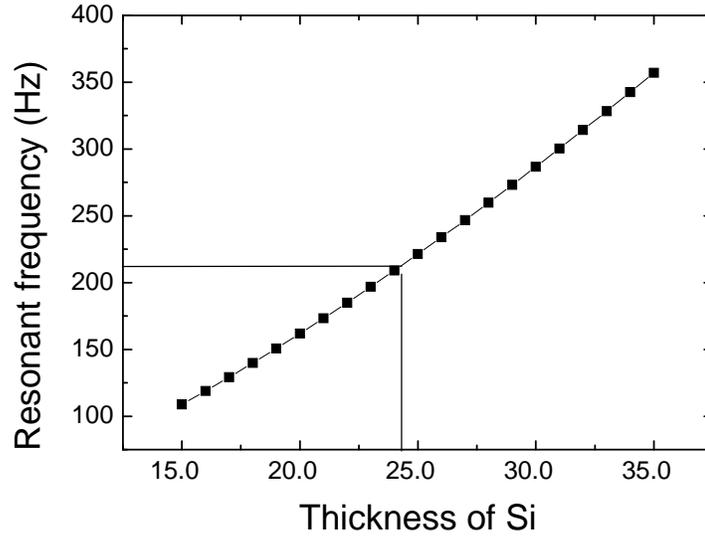
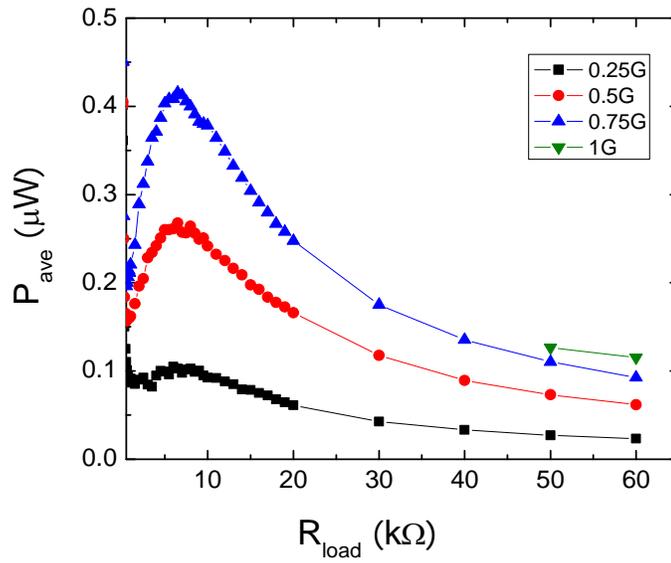
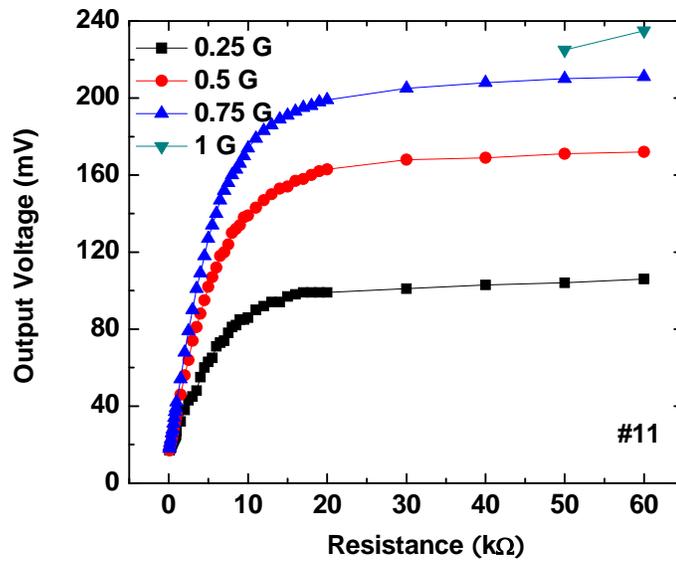


Figure 5.5 Resonant frequency is very sensitive to the thickness of the beam

Figure 5.6 shows the output peak voltage and calculated output power by Ohm's law at different resistance loads. When the maximum power is attained at 6.5 kΩ of optimal resistance load, the device had 147 mV of peak voltage and 0.42 μW under 0.75 g acceleration force. The power density is 130 μW/cm<sup>3</sup> that is very low when comparing it with the other prototypes shown in Table 2.5. The peak voltage is too small for the rectifier circuit to diminish the current. The device was broken during resonating under 1 g of acceleration force. All characterized values are shown in Table 5.1.



(a)



(b)

Figure 5.6 (a) Output peak voltage, and (b) power from single cantilever on Si

Table 5.1 Characteristic of energy harvester on Si wafer

Resonant frequency (Hz)		Optimal resistance (k $\Omega$ )	Optimal peak voltage (mV)	Optimal peak power ( $\mu$ W)	Safety factor	Volume (mm <sup>3</sup> )
Cal.	Mea.					
162	211	6.5	147	0.42	0.75	3.225

The device, fabricated on a silicon wafer, induced low power and mechanical strength, and a large deviation of resonant frequency. A hypothesis will be that the insufficient performance was caused by non-uniform silicon layers, as shown in Figure 5.7. During resonating, the stress was concentrated on the thinner part near the anchor. Since the amount of polarization is limited over large strain due to limitations of switching the domain, the low output power seems to be contributed from the restricted strain and area. While a resonating cantilever under a mechanical load induces nonlinear strain on the surface of the cantilever structure, the wide distribution of strain on the cantilever beam is preferred for energy harvesters. The low mechanical strength is also caused from the concentrated stress, and it made failure on the red circled zone, which generally has the highest strain. The frequency tuning by dimension control seems hard with silicon wafers, even if optimized etching conditions are used.



Figure 5.7 Concave etching profile after silicon etching

### 5.2.2 Device on Silicon-On-Insulator (SOI) wafer

To obtain a uniform profile of the silicon under layer, an SOI wafer was used. The SOI wafer had 20  $\mu\text{m}$  of silicon device layer and 1  $\mu\text{m}$  of buried oxide ( $\text{SiO}_2$ ). The thickness of the silicon device layer was chosen to prevent bending of the cantilever based on the previous experiments. The uniformity of the device layer was less than 0.1  $\mu\text{m}$ , which may induce about 0.5 Hz shift from the estimation considering 0.05  $\mu\text{m}$  of average value. The device has dimensions of 4 mm x 2 mm x 0.021 mm for the beam and 3 mm x 2 mm x 0.5 mm for the proof mass, and its total volume, including beam and mass, are shown in Table 5.2.

The uniformity of the silicon under layer is estimated at less than 500 nm based on the reflectance index of  $\text{SiO}_2$ . The buried  $\text{SiO}_2$  can be removed by wet etching when the front side of the wafer is passivated by photoresist after releasing the cantilever from the backside. It was not removed in this study to prevent creating an initial notch, which may result in early failure of the structure during resonating. The buried oxide thin layer was not considered on modeling of resonant frequency.

The resonant frequency was measured by tracking the maximum voltage during sweeping

the frequency in 0.1 Hz intervals. The difference from the calculated value was just 1%. To verify the analytical modeling of the resonant frequency, four other cantilevers in Figure 5.8 (a), which have different lengths, were sampled, and their resonant frequencies were measured as shown in Figure 5.8 (b). The modeling of the resonant frequency has good agreement with the experimental results at low frequency ranges.

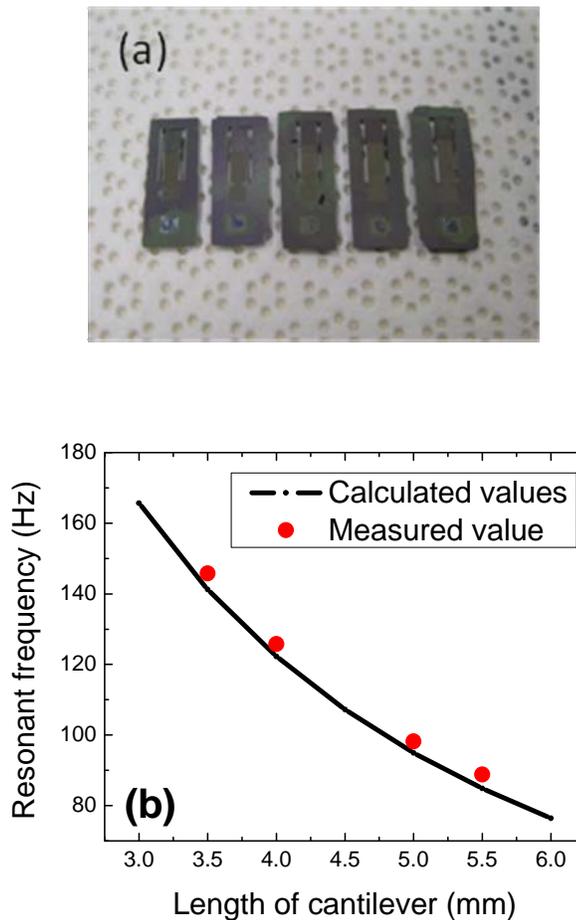


Figure 5.8 Comparison between resonant frequencies of estimated and measured values

The output peak voltage and power were plotted in Figure 5.9. When PEH was poled at 120 °C under 20V for 10 minute, the maximum power was obtained as 6.72  $\mu$ W at the 11 k $\Omega$  of

optimal resistance load and 0.5 g of acceleration. This energy harvester fabricated on SOI wafer increased the power density as  $2144 \mu\text{W}/\text{cm}^3$ , which is 15 times higher than that of the harvester fabricated on the Si wafer. The output voltage also increased and generated 0.77 V at optimal conditions, which may be over the threshold voltage of circuits. The characterized values are summarized in Table 5.2.

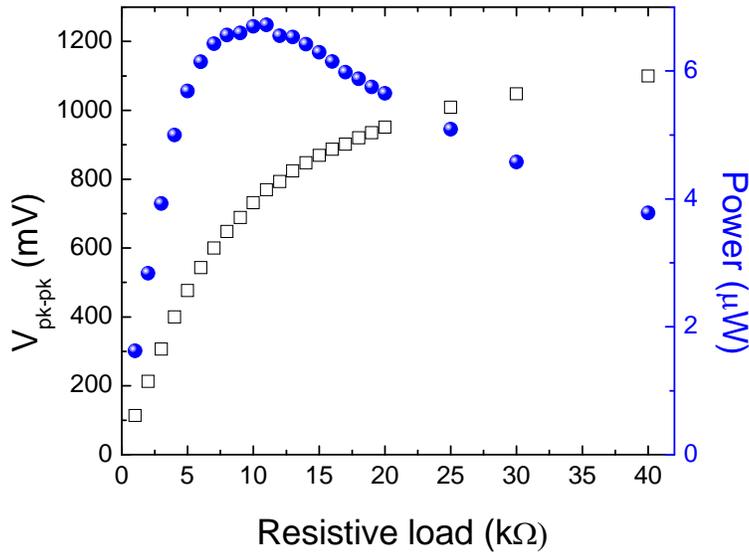


Figure 5.9 Output peak voltage and power of device fabricated on SOI wafer

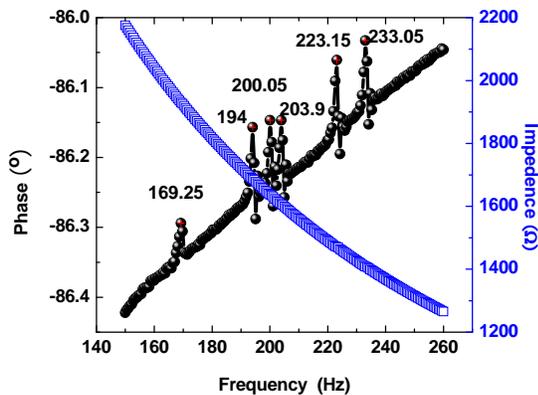
Table 5.2 Characteristics of energy harvester fabricated on SOI wafer

Resonant frequency (Hz)		Optimal resistance (kΩ)	Optimal peak voltage (mV)	Optimal peak power (μW)	Safety factor	Volume (mm <sup>3</sup> )
Cal.	Mea.					
138.61	140.3	11	769	6.72	8.1	3.294

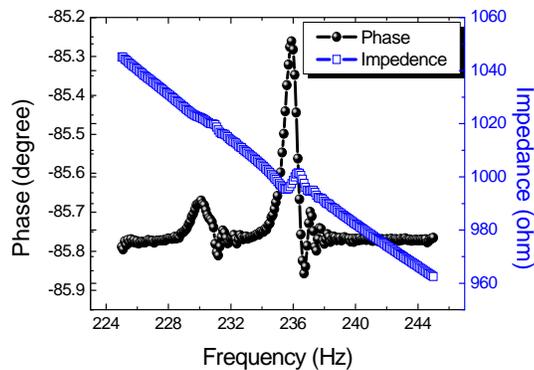
### 5.3 Performance of cantilever array

#### 5.3.1 Device on Si and SOI wafer

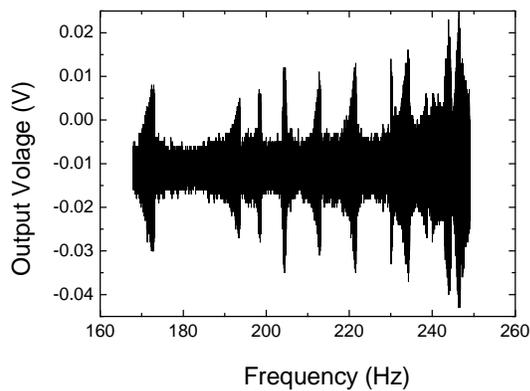
The cantilever array consisted of 10 cantilevers fabricated to increase output power via increasing voltage or current. The resonant frequencies of cantilever arrays were evaluated as shown in Figure 5.10. The array energy harvesters were separately fabricated on Si and SOI wafers. Several peaks of resonant frequencies of each cantilever in the array were shown in the 80Hz range from 169Hz to 233Hz, as seen in Figure 5.10(a). The discrete resonant frequency caused from a thickness variation in Si layer induced by unsatisfied etching uniformity in the wafer scale and window size in the ASE system. This wide range was reduced to 5Hz in range by using an SOI wafer. A varied SiO<sub>2</sub> layer works as an etching stop with about 100 times slower etching rate than that of the Si in the ASE system. The thickness variation of the cantilever structure on the SOI wafer is less than 1 micron, which is 15 times less than that of the Si wafer. Four cantilevers have the exact same resonant frequency. Raw voltage signals were collected in their frequency ranges as shown in Figure 5.10 (c) and (d).



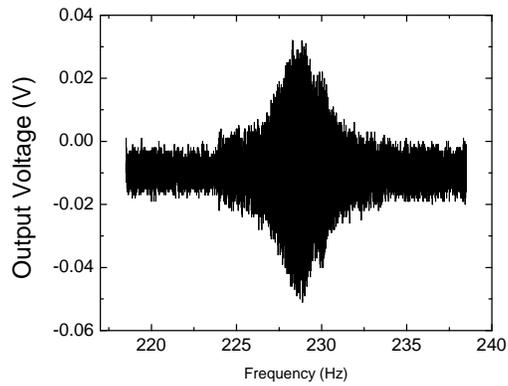
(a)



(b)



(c)



(d)

Figure 5.10 Resonance frequencies of cantilever array fabricated on (a) Si wafer, and (b) SOI wafer, and output voltage from device in sweep mode for (c) Si wafer, and (d) SOI wafer. The acceleration is 1 g

### 5.3.2 High power device

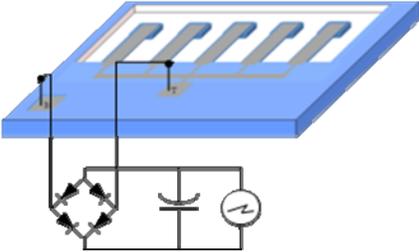
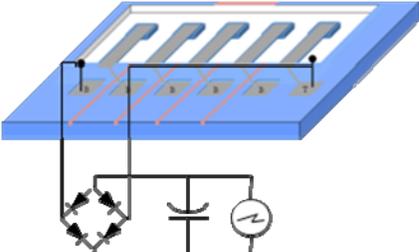
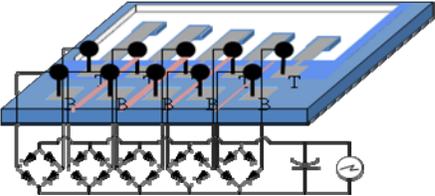
The micro single cantilever still does not have sufficient power for a power converter. In a bulk device, the bimorph with double sided PZT or multi layers are bonded and stacked to increase output voltage and power. This scheme of multi layers is hard to realize in MEMS devices, because stacking thin film PZT layers is limited by internal stress in thin film technology. In this section, the arrayed cantilever is proposed for a higher power source. Each cantilever from the array should be individually collected and rectified to multiply the signal due to being out of phase in AC. The individual circuitries may result in power losses as the numbers.

If the signals from each cantilever are in phase, the space and cost of individual power management circuitries can be saved. The design and fabrication of array cantilevers using SOI wafers enabled us to reduce the number of rectifier circuitries and led to the development of

current energy harvesting technology.

In this study, the output from the energy harvester is regulated to increase voltage or current depending on the type of connections among cantilevers in MEMS scale. Table 5.3 shows the schematic illustrations of array energy harvesters and their management circuitries as a type of connection. The first two types do not need individual AC-DC rectifiers when their resonant frequencies are finely tuned through fabrication techniques.

Table 5.3 Cantilever array and their power management for high power device

Diagram	Connection	Requirement
	<p>Parallel &amp; Unified rectifier</p>	<p>Signals in phase Uniform etching process</p>
	<p>Serial &amp; Unified rectifier</p>	<p>Signals in phase Uniform etching process</p>
	<p>Serial &amp; Individual rectifier</p>	<p>Fine frequency tuning from design and fabrication</p>

### 5.3.2.1 Parallel connection

Cantilevers with the same resonant frequency were connected in parallel for high current generation, as seen in Figure 5.11. The cantilevers were resonated at the same frequency when the same dimensions of each cantilever were built up through the optimized fabrication process and using SOI wafers. In-phase AC signals from each cantilever require just unified AC-DC converters without squandered individual rectifiers and electric loss from the interference of signals. Since the difference between the resonant frequencies of each cantilever attenuate the gain from the array configuration, their tolerances were also inspected.

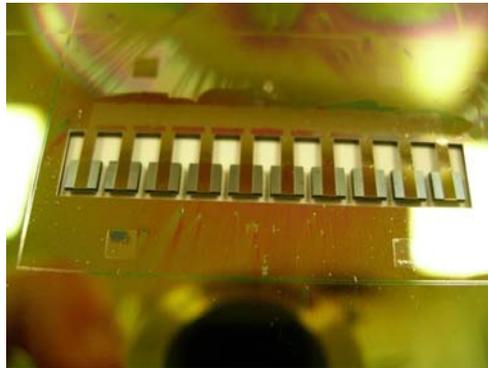


Figure 5.11 Prototypes of a pMEMS cantilever array

The array energy harvester was fabricated by the same process conditions on a SOI wafer. The resonant frequency of each cantilever was in the range of 228.6 Hz to 230.2 Hz as shown in Figure 5.12.

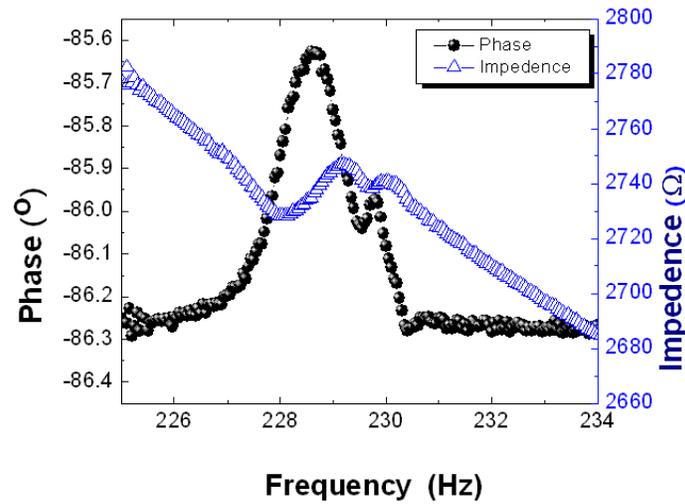


Figure 5.12 The resonant frequency of the cantilever array fabricated on a SOI wafer

The distribution of resonant frequency was caused from the thickness variation on buried SiO<sub>2</sub> layers about a few hundred nanometers due to geometrical non-uniform Si etching in the wafer scale. Three (C1, C2, and C3) out of 10 cantilevers were chosen to investigate the shift tolerance of their resonant frequencies. Their maximum peak-to-peak output voltages ( $V_{pk-pk}$ ) (subscript will be skipped from now on) were located at 229 Hz (C1), 229.1 Hz (C2), and 230.2 Hz (C3) as shown in Figure 5.13(a) when the frequency was swept from 226 Hz to 232 Hz under 1g acceleration and 5.3 kΩ of optimal resistive load of C1+C2+C3. The  $V_{C1+C2+C3}$  had a similar shape and peak to that of the phase angle measured by an impedance analyzer. The output voltage ( $V_{C3}$ ) from C3, which has a 1 Hz shift from C1 and C2 in resonant frequency, did not promote the net voltage ( $V_{C1+C2+C3}$ ), but made it decrease to around 229.5 Hz. The phase shift of the output signal for each cantilever from that of their parallel connection (C1+C2+C3) was investigated and plotted in Figure 5.13(b). C1 and C2 show about -10°, and -18° of phase shift at 229 Hz, while C3 had 48° of phase shift at that frequency. Since C3 had a low output voltage at

the frequency, it contributed very little to the increase in the net voltage in spite of its large phase difference. In Figure 5.10(a), the  $V_{C1+C2+C3}$  was 172 mV, which is 32 % less than 254 mV of the total  $V_{total}$  ( $=V_{C1}+V_{C2}+V_{C3}$ ) due to the phase difference. If seeing the second peak at 230.1 Hz, the output voltage was dominant due to its much larger output voltage and lower phase difference comparing that of C1 and C2. Since this discrepancy of resonant frequency was caused from the thickness variation of buried SiO<sub>2</sub>, it might be overcome by removing the layer by wet etching without further etching of the Si under layer.

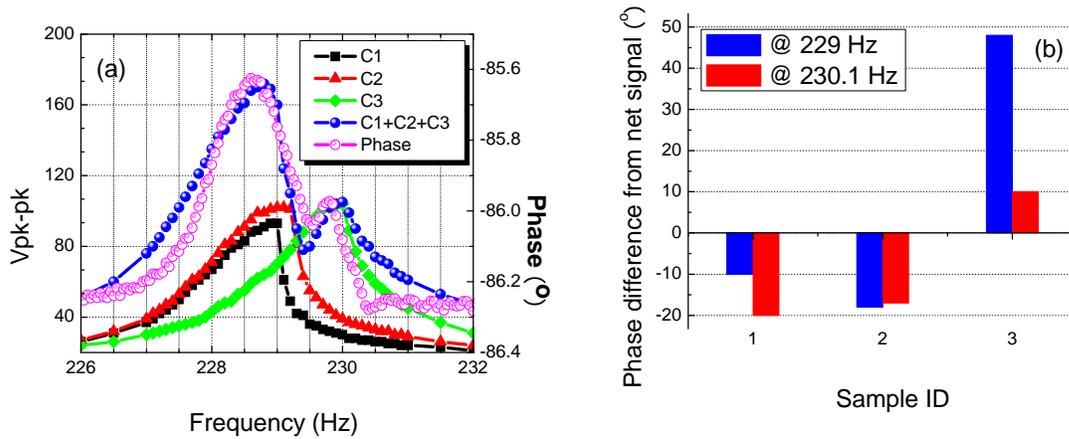


Figure 5.13 (a) The output voltage vs. frequency, and (b) the phase shift of C1, C2, and C3 from the phase of their parallel connection (C1+C2+C3)

Four cantilevers (A1, A2, A3, and A4) were chosen, and they had very close resonant frequencies in the range of from 228.9 Hz to 229.3 Hz. The evaluation of the device was performed at 1g acceleration and 229 Hz for each cantilever and their parallel connection. The peak voltages ( $V_{pk}=V_{pk-pk}/2$ ) and calculated average power ( $P_{ave}=V_{rms}^2/R_{load}$ ,  $V_{rms}=V_{pk}/\sqrt{2}$ ) were respectively plotted in Figure 5.14(a) & (b).

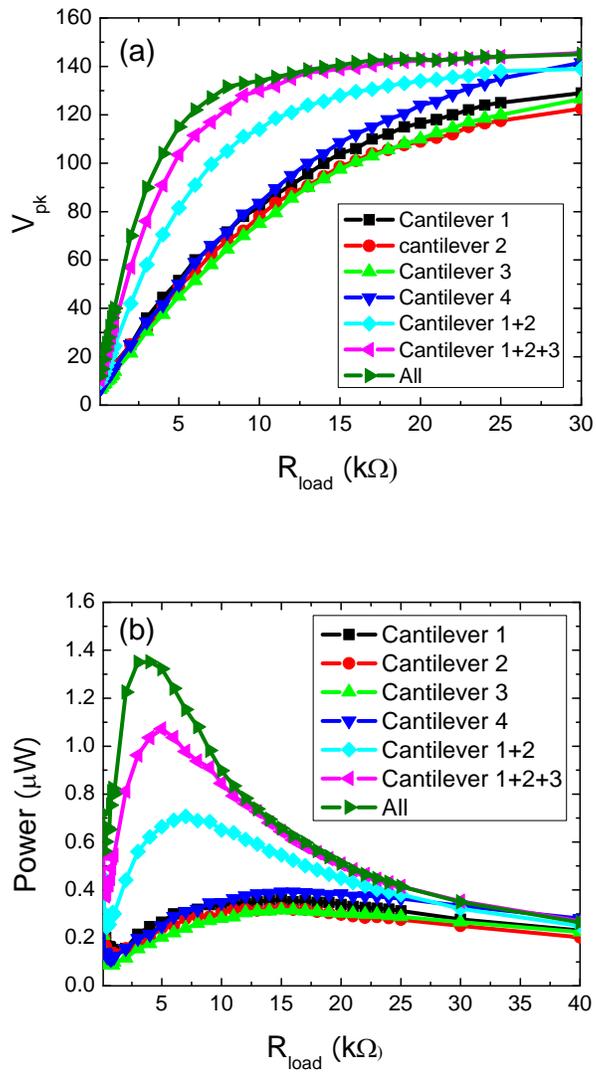


Figure 5.14 The signals from four cantilevers and their resonant frequency at 229.1 Hz; (a) Peak-to-peak voltage, and (b) output power with resistive load

The generated voltage can be expressed by Equation (5.1) using piezoelectric constant ( $d_{31}$ ) and applied force ( $F$ )

$$E = \frac{V}{l} = \frac{d_{31}}{\epsilon_0 \epsilon} \cdot \frac{F}{A} \quad (5.1)$$

where  $E$  is the electric field,  $V$  is the voltage,  $l$  is thickness of PZT,  $\epsilon_0$  is the permittivity of free space,  $\epsilon$  is dielectric constant of PZT thin film, and  $A$  is the area of top electrode. Equation (6) shows the capacitance of the cantilever array ( $C_N$ ), when the array consists of  $N$  numbers of cantilevers in parallel connections.

$$C_N = N \epsilon_0 \epsilon \frac{A}{l} \quad (5.2)$$

$$I = C \frac{dV}{dt} \quad (5.3)$$

The generated current ( $I$ ) can be obtained by substituting Equation (5.1) and Equation (5.2) into Equation (5.3)

$$I = N d_{31} \frac{dF}{dt} \quad (5.4)$$

Table 5.4 Resonant frequency, optimal resistance, generated current, and maximum output power from the array device with the number of cantilevers

	A1	A2	A3	A4	A1+A2	A1+A2+A3	A1+A2+A3+A4
Resonant frequency (Hz)	228.9	229	229.1	229.3	229	229	229
Measured R (kΩ)	5	5	5	5	7	5	4
Current (μA)	4.8	4.8	4.8	4.8	10	14.64	18.39
Max P <sub>ave</sub> @ 229Hz, 1g (μW)	0.36053	0.32341	0.31687	0.39241	0.70716	1.07123	1.352

While the output voltages under the optimal resistance load for maximum power were similar at the different resistive loads, the currents linearly increased by the number of cantilevers due to the decrease of the optimal resistive load from 15 k $\Omega$  to 4k $\Omega$ . The current was calculated using  $V_{rms}$  by Ohm's law and is plotted in Figure 5.15. The device with four cantilevers generated 18.4  $\mu$ A and 1.352  $\mu$ W, which is about 3.8 times higher than that of the single cantilever in Table 5.4. The linear relationship between the number of cantilevers and output results have shown that the device has lost a very small amount of electricity where the resonant frequencies of the cantilevers are distributed at 0.4 Hz.

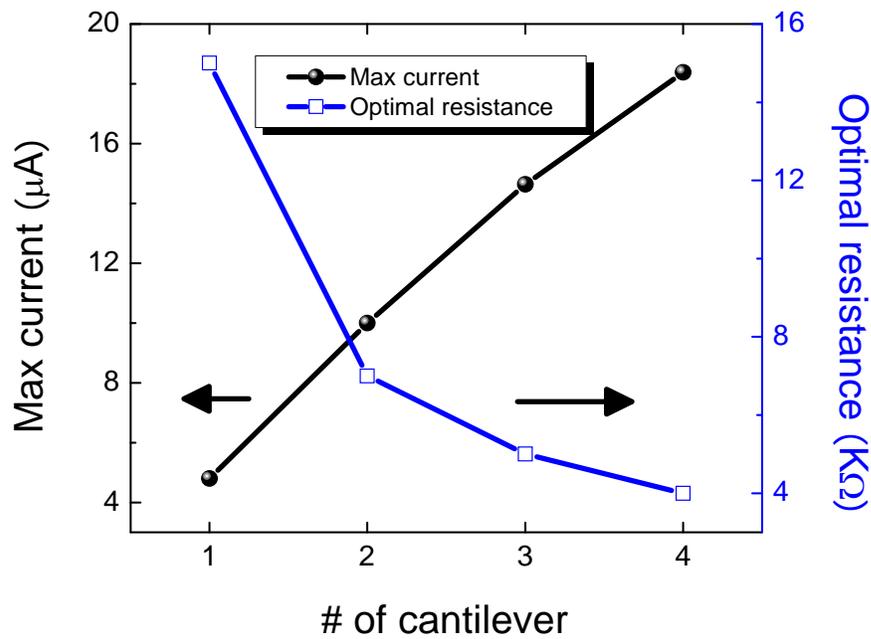


Figure 5.15 The generated current at the optimal resistance load as number of cantilevers in parallel connection

The output voltages were recorded with the same trigger at the optimized resistor load of 4 k $\Omega$  for each cantilever and their parallel connections. The output voltages were converted to current by Ohm's law and plotted in Figure 5.16. The phase difference between the signals from each cantilever was less than 10°. All signals were negatively shifted as much as 5  $\mu$ A, and it seems that the shift was induced due to the asymmetric structure of the unimorph cantilever structure. The unimorph structure only has the transducer layer on one side of cantilever frame. The neutral axis of cantilever is located at the middle of the Si under layer.

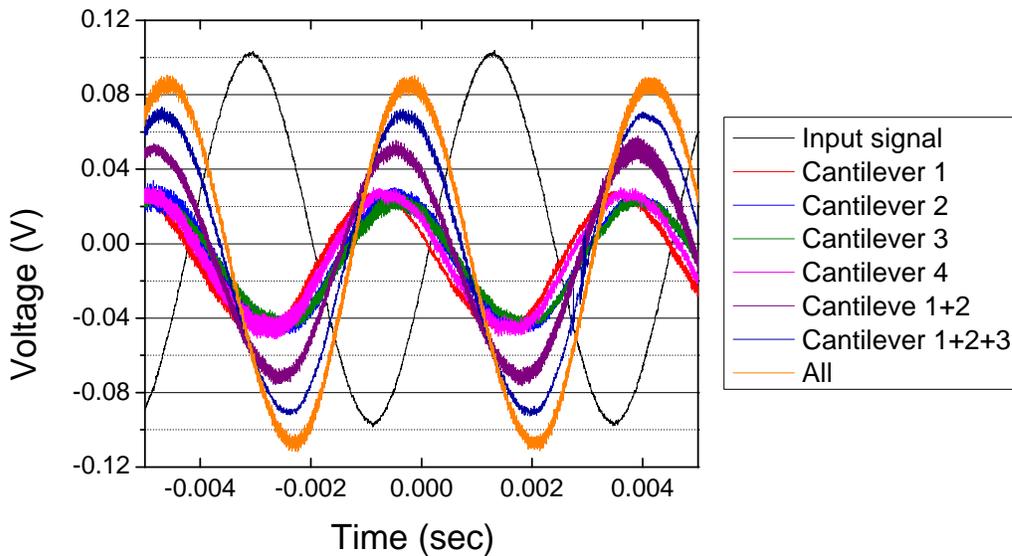


Figure 5.16 AC output of four cantilevers and their parallel connections

The results show that multi cantilever sources resonate in phase and amplify the output signal without excess electric loss and space expenses by individual circuitries. The array design can overcome the narrow bandwidth, as well as the insufficient output power of the vibration-based energy harvester.

### 5.3.2.2 Serial connection

Two cantilevers were wired in serial connection to increase the output voltage and power. Each cantilever was isolated, and its bottom electrodes were opened by dry etching. An evaluation of the device was performed under 0.25g acceleration and 129.8 Hz, and their output voltage and power was plotted in Figure 5.16.

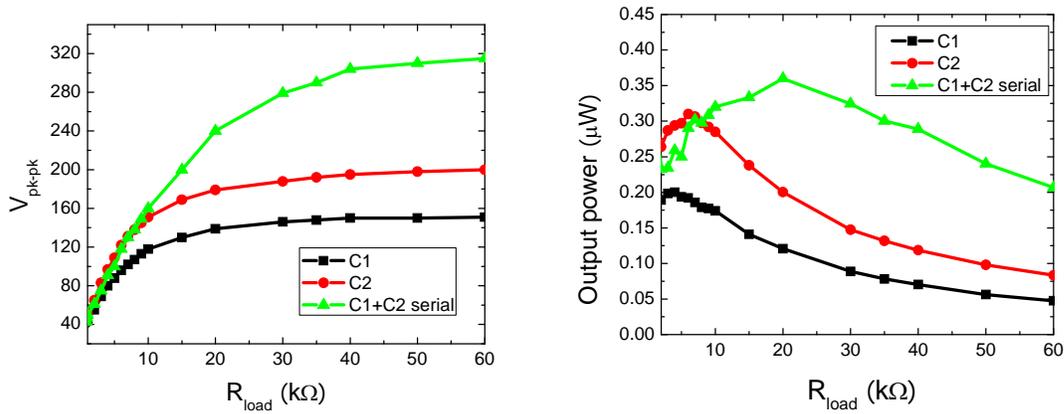


Figure 5.17 Output peak voltage and power as resistance load from serial connected array

Both cantilevers have the same resonant frequency. The serial array generated 240 mV and 0.35  $\mu$ W, which are respectively 88% and 70% from the sum of the individual cantilevers, while the currents were similar at the different optimal resistive loads.

The array device consists of two cantilevers with serial connections between the cantilevers, and its output voltage was measured as shown in Figure 5.17. Each cantilever was directly connected without an AC-DC rectifier, and the serial connection increases the output

voltage if the signal is in phase as shown in Figure 5.18.

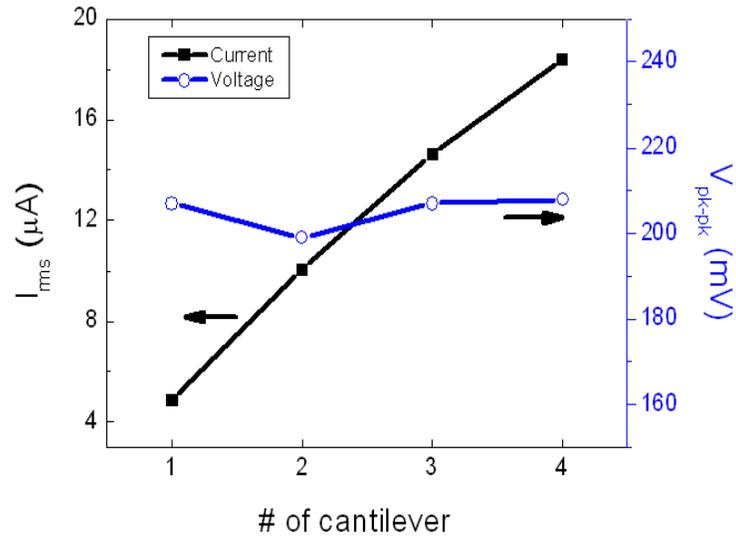


Figure 5.18 The generated current at the optimal resistance load as number of cantilever in serial connection

The array cantilever can increase the voltage or current as a type of connection depending on the requirements of the application and the specification of power management circuits.

#### 5.4 Uniform stress contribution

Much attention has been focused on developing the design of piezoelectric films to distribute the strain on the films and to increase the power density. The rectangular cantilever

beam is commonly chosen due to its high mechanical strain during environmental vibration and simple design for microfabrication, but most of stress is driven to near the anchor of the cantilever beam. This non-uniform distribution weakens the strength of the structure and reduces the conversion efficiency, since the piezoelectric charge is not always proportional to the strain induced by the stress. Baker et al. has introduced a trapezoidal-shaped cantilever for higher efficiency in bulk scale, which increased the power output as much as 30% as compared to rectangular cantilevers [91]. Although considerable research has been devoted to simulating the strain distribution according to the shape of the cantilever, rather less attention has been paid to analyzing it experimentally in MEMS devices. This study is designed to analyze the strain distribution according to the cantilever geometry and verify it experimentally using the local output power from different positions. Both trapezoidal- and rectangular-shaped cantilevers were applied on this experiment. This study gives preliminary and experimental results for improving the power density by the strain distribution on the surface of the MEMS energy harvesting device.

The geometry change in the cantilever shape may distribute the strain on the piezoelectric film and improve the power output due to nonlinear piezoelectric properties, even in the MEMS scale. The generated powers were separately collected from different positions, and they had good agreement with the strain estimation from finite element analysis.

To improve the stress uniformity, two structure configurations were proposed as shown in Figure 5.19. One has a different thickness, and the other has a different width in the longitudinal direction of the cantilever. Since the former one requires a different fabrication method in the MEMS process, the second one was chosen for this experiment. In this section,

the distribution of stress according to the cantilever shape will be inspected using a trapezoid-shaped cantilever.

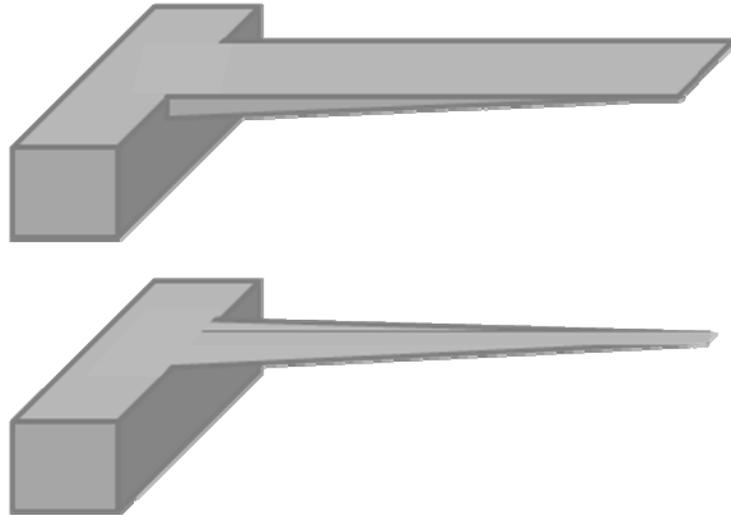


Figure 5.19 The structure for uniform stress distribution

The size of both cantilevers was  $4 \times 2$  (L x W) for the beam and  $3 \times 2 \times 0.5$  (L x W x H) for the mass in unit of mm. The trapezoidal cantilever had the reduced width at the mass of 1 mm. The platinum electrodes were separately patterned on each cantilever near the anchor, middle, and the mass, as seen in Figure 5.20. Since the direct wire bonding might affect to the vibration motion of the device or failure of wiring, the thin film wires were also patterned on the PZT surface from the electrodes to the contact pads. The wiring on the contact pad was constructed with a hand-made polymer packaging for the top electrode and silverpaste for the bottom contact.

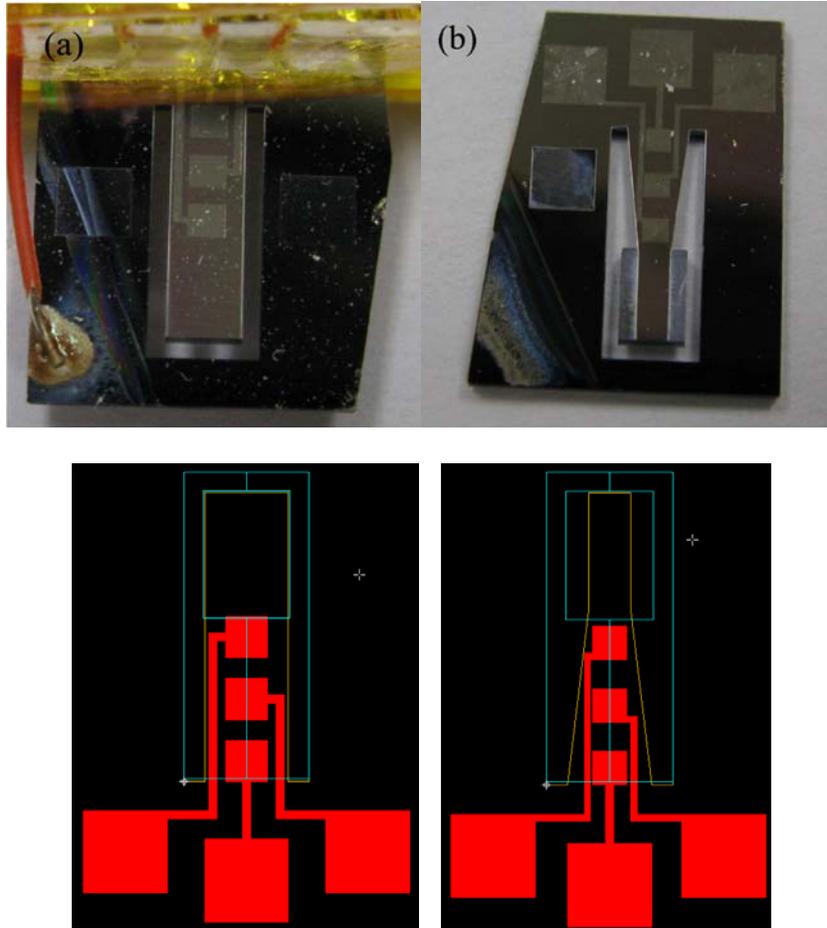


Figure 5.20 The configuration of electrode for geometric stress concentration: (a) rectangular, (b) trapezoidal. [98]

ABAQUS finite element modeling software was used to estimate the strain on the PZT film. The modeling was performed using the real size of the device with loading static force as much as  $9.8 \text{ ms}^{-2}$  of acceleration. The material properties of PZT film were as follows: Young's modulus was 69 GPa which was measured by nano indentation, Poisson's ratio was 0.3, and the density was  $2330 \text{ kgm}^{-3}$ . The material properties of Si were as follows: Young's modulus was 160 GPa in (110), Poisson's ratio was 0.27, and the density was  $7500 \text{ kgm}^{-3}$ . The modeling considered  $\text{SiO}_2$  and its properties were as follows: Young's modulus was 70 Gpa, Poisson's ratio was 0.165, and the density was  $2200 \text{ kgm}^{-3}$ . The local strains from each electrode were

obtained by averaging the strains of the mesh elements. The meshes were distributed on both cantilevers, as shown in Figure 5.21.

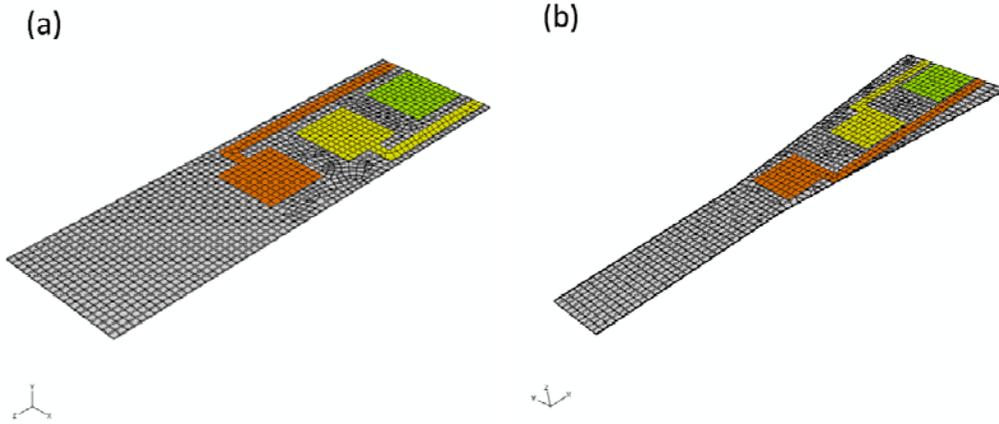


Figure 5.21 The configuration of electrode and mesh elements on (a) rectangular and (b) trapezoidal cantilever [98]

The output powers were obtained from different locations for both cantilevers to investigate the effect of strain distribution as a cantilever shape. The normalized power for each electrode was plotted at a given resistive load under  $2.45 \text{ ms}^{-2}$  of acceleration, as seen in Figure 5.22 (a) and (b). To exclude the generated powers from the wire electrodes, the total power was normalized into the real size of electrode area of  $1 \text{ mm}^2$ . There were two limitations on this normalization. Since the generated charge on piezoelectric material is not proportional to the amount of strain, its nonlinearity should be considered on this normalization. The nonlinear piezoelectric coefficient, however, was neglected in this study. The other limitation is that the output powers were not exactly come from the square electrode pad as modeling. The resistive loads of  $20 \pm 1 \text{ k}\Omega$  were shifted, because the small differences between electrodes were caused by

the different sizes of electrodes, including the area of the wire before normalizing. The maximum powers of each electrode were nevertheless used to compare them with the average strains from FEM analysis, as shown in Figure 5.22 (c) and (d). The electrodes near the anchor showed the highest normalized power in both cantilevers, while the power near the mass has the smallest values. Now, it is important how much power was generated at the middle electrodes as compared with that of the others. The powers from three electrodes showed a nonlinear relationship, while the powers at the middle electrodes for rectangular and trapezoidal cantilevers had smaller and larger values, respectively. If assuming that the materials in the cantilever structure are linear, elastic, and isotropic with average stress applied along the 1-1 direction, the output power is expressed as follows [12]:

$$P = \frac{v^2}{R_L} = \frac{1}{R_L} \left\{ g_{31} E \varepsilon \left( \frac{x}{L} \right) L \right\}^2 \quad (5.5)$$

where  $x$  is the location from the clamped end,  $L$  is the length of beam,  $g$  is the piezoelectric voltage constant,  $E$  is the Young's modulus, and  $\varepsilon$  is the strain. In Figure 5.22(c), the relationship between strain and power has been agreed with Equation (5.5). On the contrary, the trapezoidal cantilever has a negative parabolic relationship. The reason is suggested that the strain distribution has spread out into the region of the middle electrode in the trapezoidal cantilever structure. The nonlinear piezoelectric coefficient ( $g_{31}$ ) might increase the total output power through the entire PZT thin film. Unfortunately, the quantitative analysis of this result cannot be accompanied at this time.

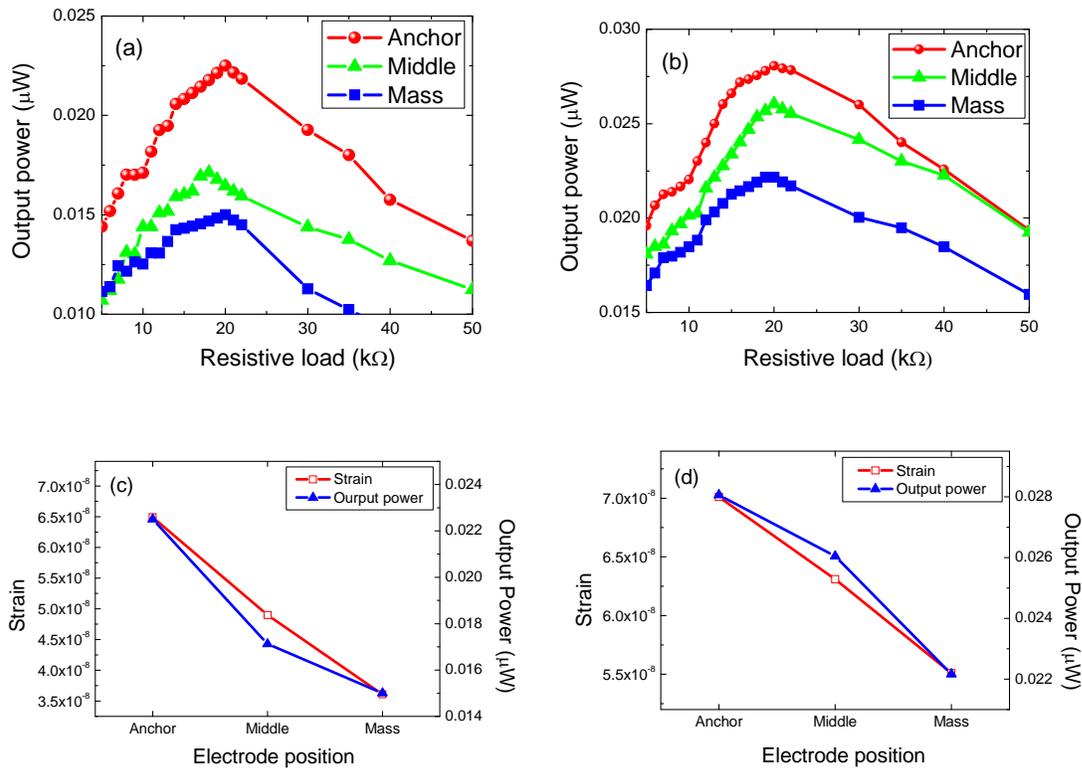


Figure 5.22 The output powers of different electrodes for (a) the rectangular and (b) the trapezoidal cantilever across the resistive load. Their maximum powers were compared with the FEM modeling strains for (c) the rectangular and (d) the trapezoidal cantilever [98]

The strain profiles along the length of cantilever for both cantilevers were plotted in Figure 5.23. Three electrodes were located in range of 0~0.2, 0.4~0.6, and 0.8~1.0. Since the strain profiles were obtained from the center of the cantilever, the amount of strains are a little different from what's reported in Figure 5.20 (c) and (d). While the rectangular cantilever shows a linear decrease, the trapezoidal cantilever has distributed the strain along the length of the cantilever.

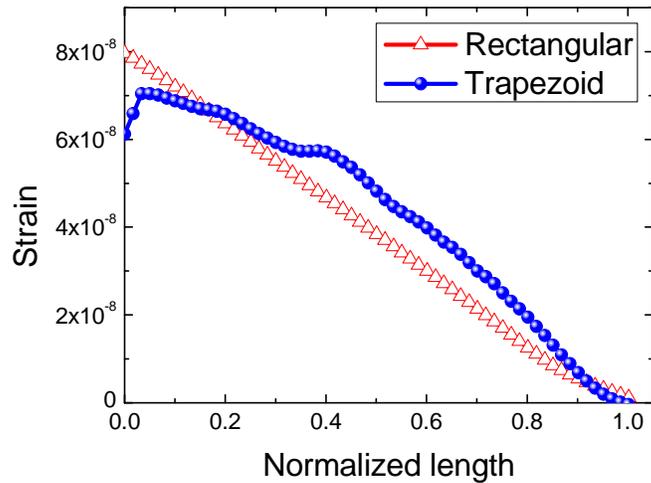


Figure 5.23 The strain profiles from anchor to mass for the rectangular and the trapezoidal-shaped cantilever [98]

When all electrodes were connected together, the output powers were measured for both cantilevers to compare the total power. as seen in Figure 5.24. The maximum powers were obtained at the same resistive load of 7 k $\Omega$  for both cantilevers. since the electrode area was increased, the optimal resistive load was decreased compared to the small individual electrodes. The output power from the trapezoidal cantilever was increased to 39% comparing with that of the rectangular cantilever.

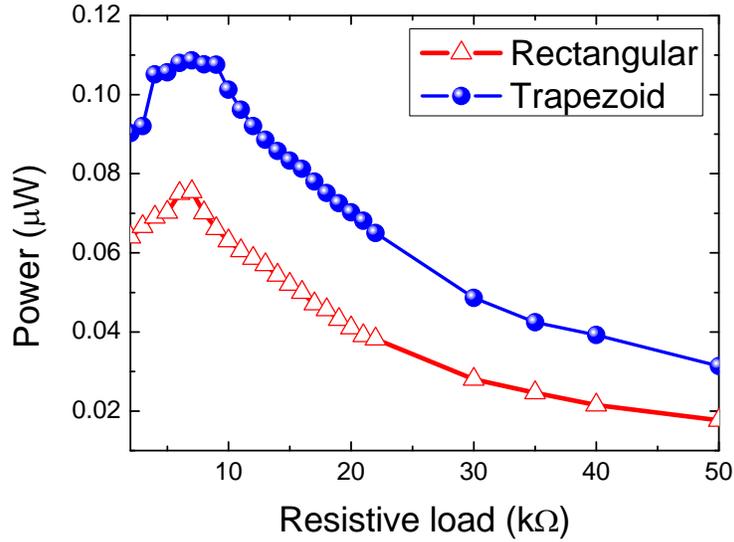


Figure 5.24 The generated total powers from rectangular and trapezoid [98]

The maximum power per unit area ( $\mu\text{W}/\text{cm}^2$ ) was compared between the rectangular and trapezoidal cantilevers. Two cantilevers have the same dimension of proof mass and length, but different widths due to their shapes. The unit area was used to compare the power density affected by the shape of the cantilever. Although the trapezoid has a smaller area than that of the rectangle, the power density using the unit area shows the efficiency of the cantilever depending on the shape. The trapezoidal shaped cantilever might have improved stress distribution, the stress still concentrated near the anchor. This is why the width near anchor has the same dimension between the rectangular and trapezoidal cantilevers. The output maximum power density ( $\mu\text{W}/\text{cm}^2$ ) from the cantilevers is compared in Figure 5.25.

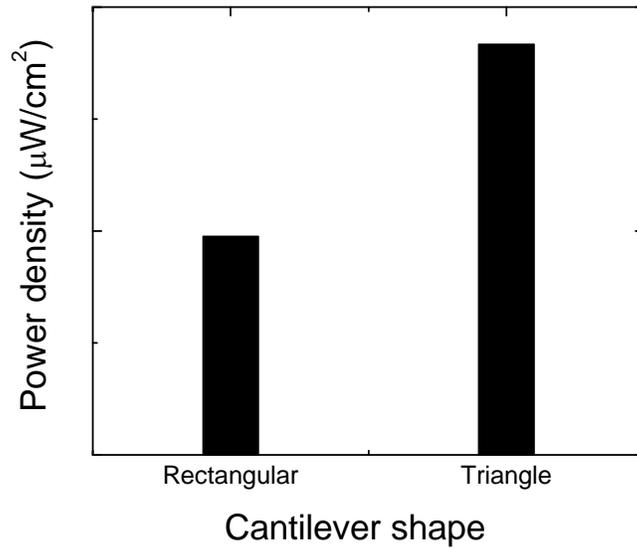


Figure 5.25 The output maximum power as cantilever shape

## 5.5 Comparison between 31 mode and 33 mode

The 31 mode and 33 mode piezoelectric energy harvesters were fabricated using the same dimensions and same quality PZT film on Pt film. When the strain is loaded in a longitudinal direction of the cantilever, the top and bottom electrodes are normal to the strain, and interdigitated electrode (IDE) collected the discharge directly to the strain. The schematic illustrations of the two modes are shown in Figure 5.26. A simple comparison of output power for the two different modes does not have much meaning because they have their own piezoelectric characteristics for energy harvesting applications. Furthermore, output power from the 33 mode strongly depends on the configuration of IDE.

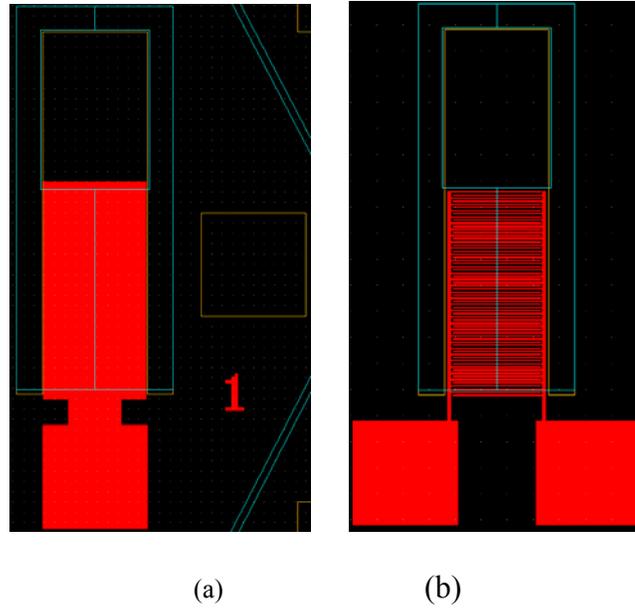


Figure 5.26 Schematic illustration of 31 mode (a) and 33 mode (b) cantilevered piezoelectric energy harvester

The 31 mode conversion usually generates a high current with a low impedance due to the large electrode area. On the contrary, a relatively small output voltage might be the limitation of this mode. The output voltage is decided by the thickness of the piezoelectric film between the top and bottom electrodes, since the output voltage ( $V_{oc}$ ) under open circuit is defined by Equation 5.6.

$$V_{oc} = \sigma_{xx} g_{31} H \quad (5.6)$$

where  $\sigma_{xx}$  is stress on the length of cantilever,  $g_{31}$  is the piezoelectric constant of 31 mode, and  $H$  is the thickness between the top and bottom electrodes.

The output of the 33 mode conversion using IDE is more complex because the electrode area and the distance between the two electrodes varies by dimensions of width and gap in IDE. Figure 5.27 is a simple illustration of IDE and shows dimensional notations.

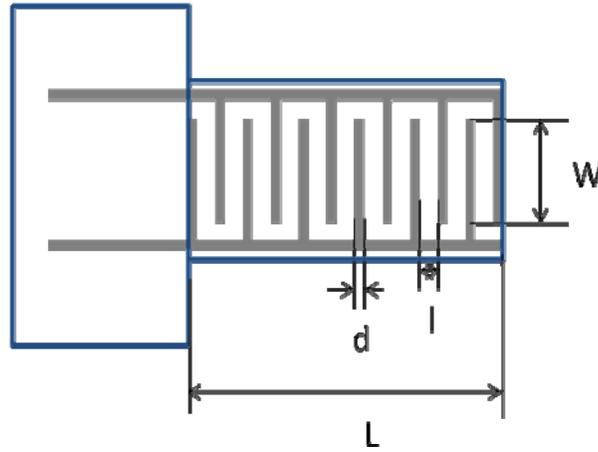


Figure 5.27 Schematic of IDE

The output voltage from IDE is decided by the distance ( $l$ ) between the finger electrodes, and the output voltage ( $V_{oc}$ ) under open circuit is defined by Equation 5.7.

$$V_{oc} = \sigma_{xx} g_{33} l \quad (5.7)$$

where  $g_{33}$  is the piezoelectric constant of the 33 mode. The expected voltage level can be obtained in the 33 mode conversion by adjusting the distance ( $l$ ). In the mean time, the 33 mode conversion generates a low current with high impedance. As the distance ( $l$ ) increases to obtain the high voltage level, the number of fingers is reduced in the limited length ( $L$ ) of the cantilever. When the electrode area ( $A$ ) of IDE is defined by Equation 5.8, the impedance is the same as the optimal resistive load ( $R_{opt}$ ) where the maximum power is obtained increase as shown in Equation 5.10.

$$A = Wl(N - 1) \quad (5.8)$$

$$C = \epsilon_r \epsilon_0 \frac{A}{l} \quad (5.9)$$

$$R_{opt} = \frac{1}{2\pi \cdot f \cdot C} \quad (5.10)$$

where  $t$  is the thickness of piezoelectric film,  $N$  is the number of IDE finger,  $C$  is the capacitance,  $\epsilon_0$  is the permittivity of vacuum,  $\epsilon_r$  is dielectric constant, and  $f$  is the resonant frequency.

When poling PZT using IDE, the entire film cannot be poled in a longitudinal direction due to the area under the electrode area. The unpoled area under the electrodes is called dead zone as shown in Figure 5.28.

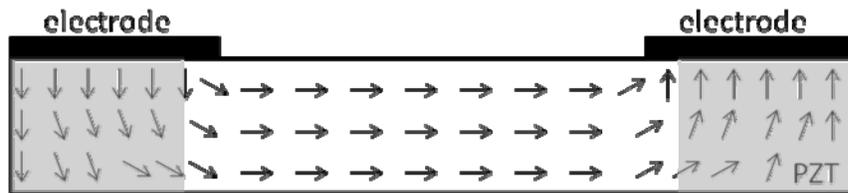


Figure 5.28 Poling directions on IDE with show dead zone as a grey color

The dead zone can be reduced by deciding the width of the electrode considering the thickness of PZT, such as 0.8 ratio of width to thickness [99]. While the large thickness of PZT is limited in film technology, the width of electrode is hardly reduced in few microns due to lithography process. Although the 33 mode piezoelectric constant may be larger than the 31 mode, the output powers from the two different modes are hardly compared due to the limitation of IDE.

### 5.5.1 Poling effect on 31 mode and 33 mode

The dipoles of as-deposited PZT thin film has random orientation and the film show no

piezoelectric character due to zero net-polarization. When the film has poled in proper condition, it shows the maximum piezoelectric properties which mostly increase the efficiency of PEH. The bulk PZT, which has no internal field and no stress, has symmetry piezoelectric response. In the contrary to the bulk PZT, PZT thin film has asymmetry character due to its residual stress, defective region created during the deposition, or internal field. The poling condition should be adjusted considering the asymmetry condition of the films. In this chapter, the poling conditions were optimized and the output performance of PEH simultaneously investigated as the poling conditions.

#### *Poling effect on d31 mode*

The P-E hysteresis loop was given in Figure 5.29 for the PZT film on PEH which had dimension in 4100 x 500  $\mu\text{m}$ . The loop had saturated polarization in 200 kV/cm and gave about 20  $\mu\text{C}/\text{cm}^2$  of the remnant polarization and about 32 kV/cm of coercive field. The loop was shifted -7 kV/cm along the field axis, and the shift was preserved after several tests. It was assumed that the shift is caused from the internal field. The internal field seems to be created during the etching by RIE, since the shifted field had same direction with the bias during plasma etching. The internal field was probably induced by energetic bombardment or ion peening during RIE. The other assumption is that the tensile residual stress shift the loop [100, 101].

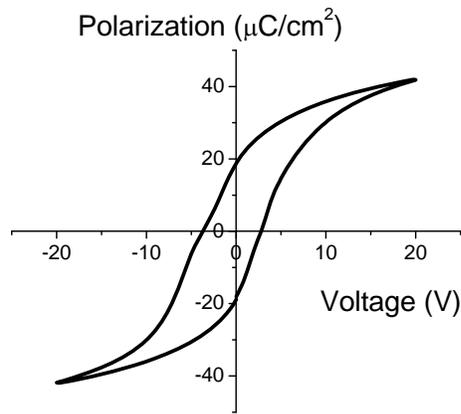


Figure 5.29 Polarization hysteresis loop for the PZT film on PEH

The effect of poling direction under existed internal field was investigated. Figure 5.30 gives the output power versus the resistive load under 0.25 g acceleration force as poling conditions at room temperature. The negative poling field which is in a direction antiparallel to the internal field gained small output powers, while the film poled in the direction parallel with positive field on the top show higher efficiency of PEH. It is believed that the change of the output power caused from the internal field. It is reported that the piezoelectric constant decrease by the poling in a direction antiparallel to the internal field [102, 103] varies by the poling direction, and the results has good agreement with them.

A slight increase of the output power in positive poling field was shown as the poling time from 1 second to 15 min. In the contrary, the negative poling field in short time had much smaller output power since it is insufficient due to existed opposite internal field.

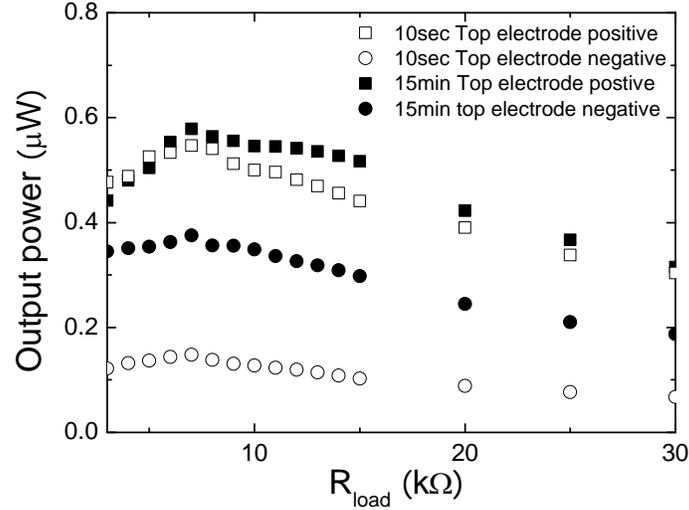


Figure 5.30 Output power as the poling conditions

The poling with DC bias field was performed in either direction parallel and antiparalle to the internal field. During the poling, the PEH was heated to increase to obtain the full poling state. Positive and negative 20 V were applied at 120 °C for 10 min, and the hysteresis loop was measured in Figure 5.31 (a). The P-E loop was shifted  $-4\text{kV/cm}$  before DC bias, and  $V_{c\text{-shift}}$  ( $(|V_{+c}|+|V_{-c}|)/2$ ) increase or decrease as negative or positive biase was respectively applied as seen in figure 5.31 (b). The output voltage and power was plotted in figure 5.32 (a) and (b). When the poling at 120 °C with the positive bias was perfomed, the output power increase about 30 % comparing room temperature poling. The poling direction still affects to the piezoelectric response at high temperature.

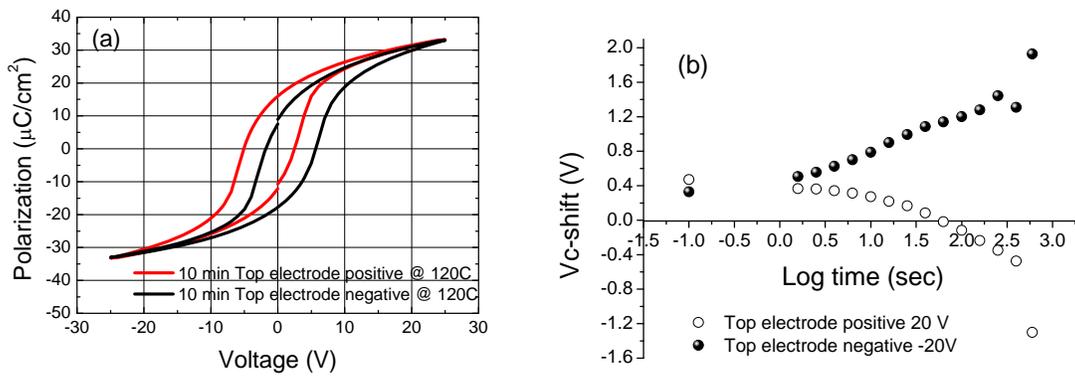


Figure 5.31 (a) P-E loop after DC biased poling at high temperature (b) the shift of coercive field as direction of poling

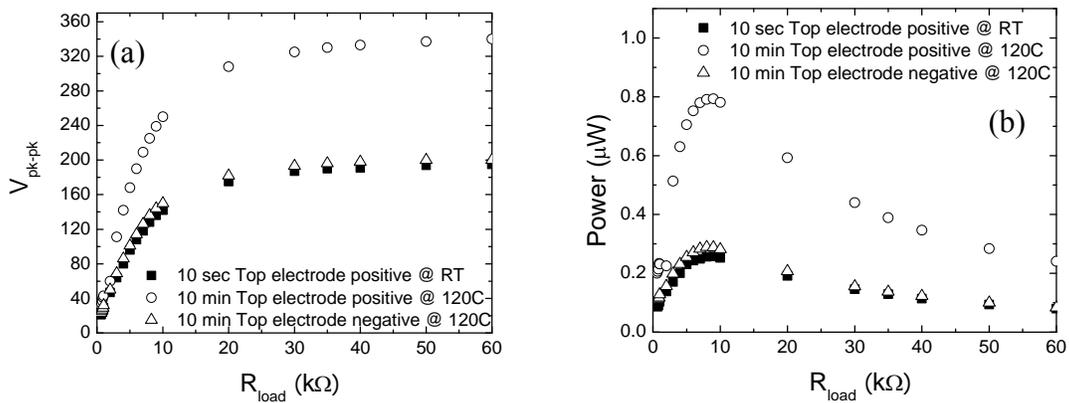


Figure 5.32 The performance of PEH as poling conditions: (a) output voltage and (b) output power

### Poling effect $d_{33}$ mode

A proper poling condition is required to obtain the highest conversion efficiency from the PZT-based piezoelectric energy harvester. The  $d_{33}$  mode using IDE has difficulties in poling because it requires high voltage and its high electric field might pass through interface with

substrate or surface of PZT film. The PZT film (4000 x 2000  $\mu\text{m}$ ) which has IDE (20  $\mu\text{m}$  electrode width and 40  $\mu\text{m}$  gap) was sampled, and its hysteresis loop was obtained up to 200 kV/cm in a silicon oil bath as shown in Figure 5.33 (a). The electrode area was defined by Equation 5.8. The slant loop was obtained with saturated remant polarization at 100 kV/cm and linear increase of maximum polarization as electric field. A possible reason of the slanted loop is the existed non-ferroelectric regions between electrodes or space charge which restrict the domain motions. When inspecting the crystallization of the PZT film on  $\text{ZrO}_2$  in Figure 5.33 (b), the PZT film was well-oriented without pyrochlore phase. The different microstructure along the longitudinal axis probably induced the space charge, but its deeper study should be followed. it is well oriented to and The wide gap between electrode in IDE

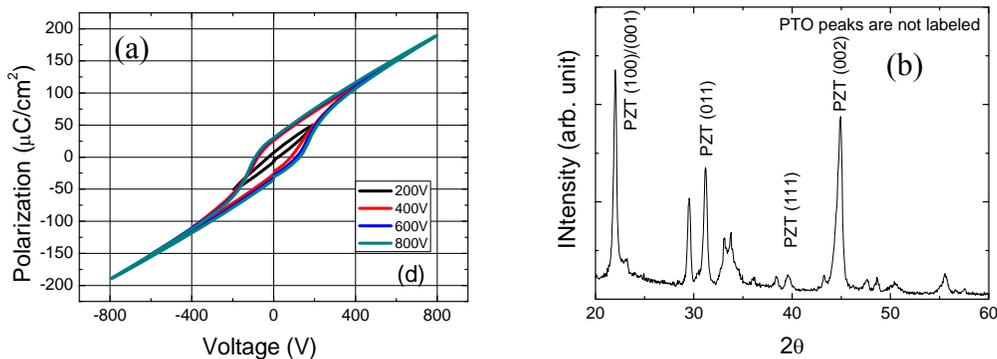


Figure 5.33 The characterization of the PZT film in d33 mode PEH: (a) hysteresis loop & (b) XRD analysis

The output voltages as the poling conditions were plotted in Figure 5.34 (a). A longer poling time for 20 min was inspected to obtain a full poling state. When poling electric field was increase from 25 to 100 kV/cm at room temperature, the output voltage at a open circuit saturated from 62.5 kV/cm. The PZT film seems to be damaged under 100 kV/cm DC bias

poling for 20 min. In figure 5.34 (a), the output voltages are same under two poling conditions between the short time 200 kV/cm AC poling at room temperature and the long time 62.5 kV/cm DC poling at 100 °C. Further increase of DC bias poling could not be performed due to a breakdown.

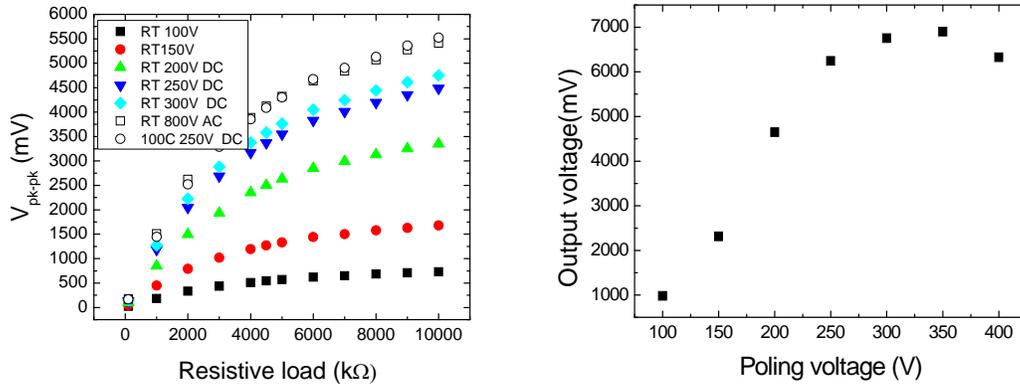


Figure 5.34 (a) output voltage as different poling conditions & (b) output voltage at open-circuit as poling field

### 5.5.2 Outputs in 31 mode and 33 mode

The output voltage and power were measured and plotted for two different modes in Figure 5.29 (a) & (b) under same poling condition of 200 kV/cm electric field. Under 0.25 g of acceleration condition, the 31 mode had 127.2 Hz of resonant frequency and the 33 mode has a little lower frequency at 118.1 Hz. The difference induced by the configuration of top electrode. The IDE has a 20  $\mu\text{m}$  width and 40  $\mu\text{m}$  distance between the finger electrodes, so a total of 67 fingers make up the length of the cantilever. Output voltage from  $d_{33}$  mode is 15 times larger than that from  $d_{31}$  mode, not 40 times as distance increase between the electrodes. It seems that high concentration of grain boundaries across the finger electrodes in IDT makes more pinning

of the domain walls and reduce the piezoelectric response. Max power from {3-1} mode is 2 times higher than that from {3-3} mode due to its 28 times larger current with small optimal resistive load. Both modes were sufficiently poled in 200 kV/cm of electric field where the film exhibits reasonable ferroelectric properties. The remnant polarizations of 31 and 33 modes respectively were 20 and 32  $\mu\text{C}/\text{cm}^2$ , and the coercive electric field was  $\sim 31$  kV/cm for both films.

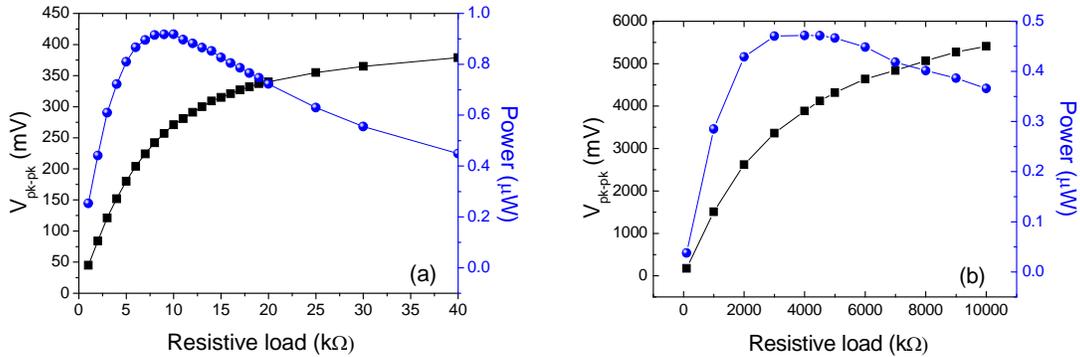


Figure 5.35 The comparison between 31 and 33 mode conversion based on cantilever displacement: (a) output from 31 mode, (b) output from 33 mode

Table 5.5 Comparison between  $d_{31}$  and  $d_{33}$  mode

	$d_{31}$ mode	$d_{33}$ mode
Distance	1 $\mu\text{m}$	40 $\mu\text{m}$
Dimensions	7x2mm (4x2 mm cantilever, 3x2 mm mass)	
R.F. (Hz)	127.2	118.1 Hz
Poling conditions	200 KV/cm AC @ RT	
Optimal R (kΩ)	10	4000
Output Voltage (mV)	271	3885

Max. Power ( $\mu\text{W}$ )	0.918	0.472
Current ( $\mu\text{A}$ )	9.6	0.34
Open Circuit Voltage (mV)	~400	7320

When 31 and 33 modes PEHs, poled at the same condition, work under the same mechanical conditions with the same dimension, the output power of 33 mode is seriously restricted by its low current due to small area and large distance. The 33 mode has advantage on the high voltage generation which is preferred to store charge and rectify AC to DC. The increase, however, was not as much as the longer distance compared with that of thickness in 31 mode. While the 31 mode PEH could increase the output voltage in sufficient level by the poling condition such as high temperature and DC electric field, the 33 mode with 40  $\mu\text{m}$  electrode's gap could not as seen in chapter 5.5.1. In current results of this study, the 31 mode PEH has higher output power more than the 33 mode and its output voltage is enough high for the rectifier and charge storage. The 33 mode PEH is simultaneously under investigation, since the IDE design with shorter electrode's distance might increase the current and power.

## 5.6 Reliability

Most MEMS devices have issues with their reliability. The energy harvester targets more than 10-year lifetime, but the piezoelectric MEMS cantilever also has been faced with the mechanical failure of the cantilever as well as degradation of piezoelectric properties. The shift in resonant frequency caused from any deformation of its structure loses the target frequency and results in a decrease of output power from the energy harvester. The cyclic loads on the PZT

cantilever may induce the deterioration of structure on both structure and PZT. The created cracks on PZT film reduce the piezoelectric response, and the cracks on the Si substrate may result in the instantaneous failure of the structure due to its brittle structure of single crystalline. The possible mechanical failure and degradation of device are listed on Table 5.5.

Table 5.5 Reliability issued on Si-based PZT cantilever

Structure failure	Piezoelectric PZT
Fracture or fatigue	Depoling due to
- Overload fracture	- Mechanical load
- Fatigue fracture	- Temperature
Delamination	Leakage current
	- microcracks

The failure of structure might be caused from fracture or fatigue under overload of high acceleration or cyclic load, respectively. The delamination between the PZT active layer (Pt/PZT/Pt) and the substrate or generated cracks might induce a shift of resonant frequency or failure of Ohmic contact between the PZT and electrode, which results in decreasing output power due to the resonant frequency. The cyclic load or high temperature might induce the depoling of PZT, or generated microcracks could result the in unrecoverable degradation of PZT. In this section, the reliability of energy harvester will be investigated in structure damage and degradation of PZT film by mechanical and environmental impacts.

### 5.6.1 Structural failure by fatigue

The PZT cantilever was designed considering the mechanical strength of PZT compared

to the maximum stress on PZT, and the safety factor ( $\sigma_{\text{yield}}/\sigma_{\text{max}}$  at 1g acceleration) was already estimated in Chapter 4. The safety factor was calculated using the average stress on the surface of structure, and the device should stand up to several gravity forces (G). The structural failure, however, occurred before the estimated point during experiments. Although errors in estimation exist, early failure of the structure can be caused by other reasons. The microstructural investigation of failure in the cantilevered structure was performed. Since the notch test for fatigue for the MEMS cantilever is technically restricted, the fractured surface was investigated by SEM in Figure 5.30.

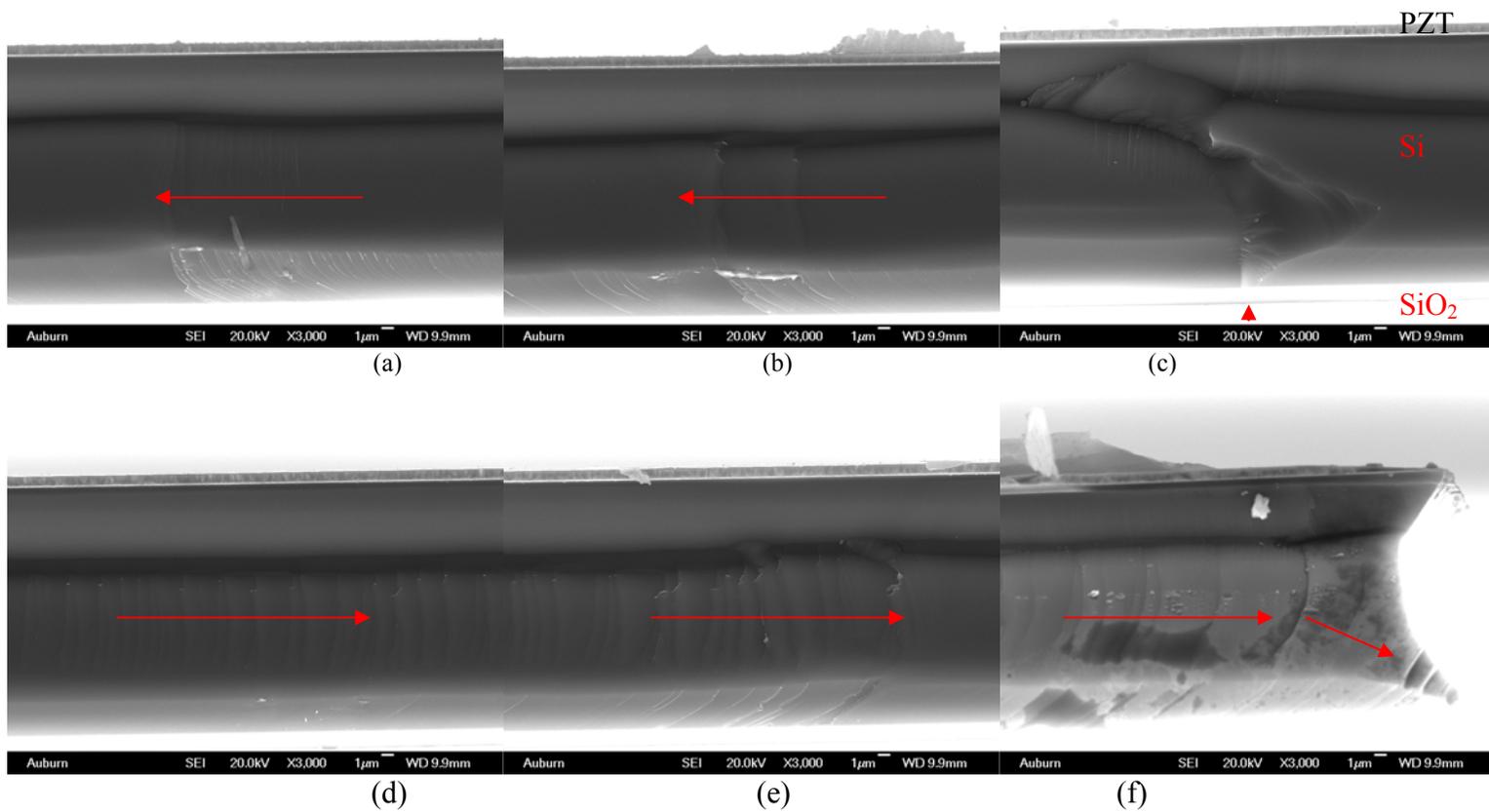


Figure 5.36 SEM image of fractured surface of cantilever

The cantilevered structure failed due to the cyclic load of high acceleration. It was resonated for 5 minutes each under 0.75, 1, 1.25, 1.5, 1.75, and 2 g acceleration conditions. The failure occurred near the anchor, and the wide range of the fractured surface was inspected. The single crystal of Si has almost four times higher fracture toughness than that of PZT thin film. While the Si has below  $1.94 \text{ MPa}\sqrt{\text{m}}$  of toughness, the PZT thin film has below  $0.5 \text{ MPa}\sqrt{\text{m}}$  [92-94]. The fatigue propagation should be nucleated and grown from a PZT surface that has low toughness if they are under same amount of cyclic load, but the crack was initiated from the Si surface and generated failure quickly. When inspecting the fractured surface, it was found that the origin of fracture was located some place in the middle as seen in Figure 5.30(c). It seems that the crack started from the base side across the  $\text{SiO}_2$  layer. From this point, the water marks were found on the entire cantilever and they emanate toward the outside in a regular pattern. Figure 5.30 (a) & (b) show the left side from the origin of crack and Figure 5.30 (d)~(f) show the right side. The etching surface by RIE is not perfectly smooth (based side in Figure 5.30), and it seems that the rough surface created the initial crack. They have played role of initial crack, and the Si failed prior to the PZT film in spite of its higher toughness. The substitute of substrate material is restricted in the MEMS process, but the polymer might be one of the possible materials using the techniques of polymer MEMS. These questions remain to be answered in future work based on this study.

## **5.7 Degradation of PZT**

### **5.7.1 Degradation during fabrication**

The degradation in piezoelectric properties of PZT thin film during the microfabrication

processes was investigated. The PZT thin film might be degraded in a plasma or hydrogen containing process during etching or passivation of microfabrication processes. The strong plasma accumulates the positive ions on the surface of PZT, which decrease the polarization [95]. The damage during sputtering or RIE can result in a decrease of polarization due to the second phase. The hydrogen atoms lead to the hydroxyl ions, which prevent the switching of Ti ion in piezoelectric response [96, 97]. The piezoelectric properties of PZT in each fabrication steps were investigated, as shown in Figure 5.31.

No degradation was found after plasma etching in the aspect of remanent polarization and dielectric constant. Since the PZT film is covered by Pt and photoresist during the etching process, the film is effectively protected from damage through the surface and even through the side.

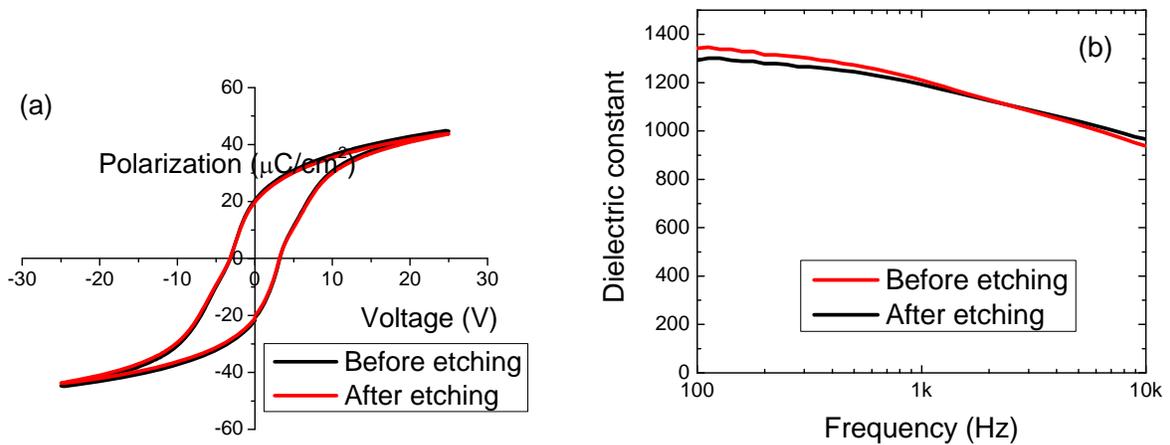


Figure 5.37 The piezoelectric properties of PZT after plasma etching: (a) PE loop and (b) dielectric constant

### 5.7.2 Reliability during cyclic load

PZT film has degradation of its piezoelectric properties as electric or mechanical cyclic load. Adding to the mechanical failure, the cyclic load might induce the mechanical or electrical degradation to the PZT film such as cracks or depoling. Since the energy harvester only encounters the mechanical load, its performance was inspected for up to three days in a continuous cyclic load. Under 0.5 G acceleration and open circuit conditions, the output voltage from the device was monitored with the resonant frequency, as shown in Figure 5.32. When the temperature on the PZT film was monitored, no heating up was found during the entire cyclic load.

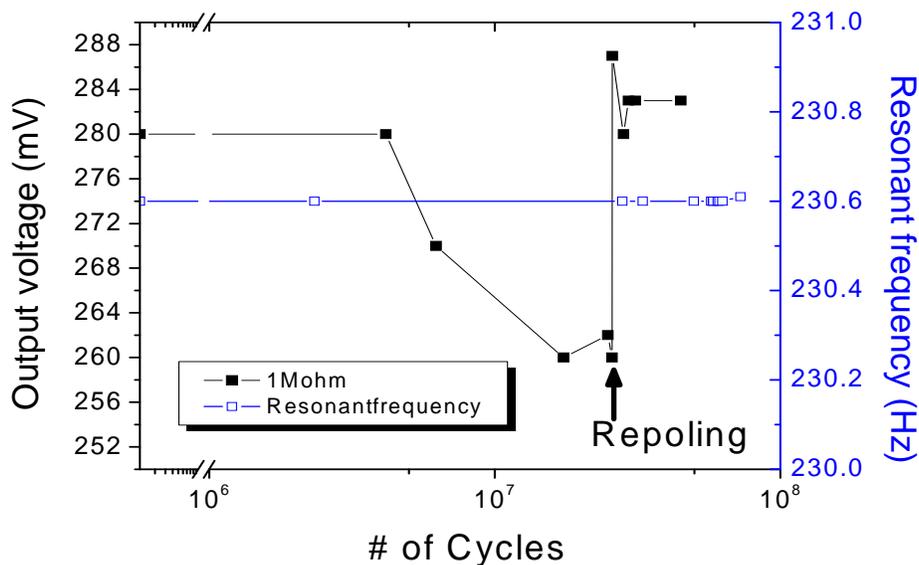


Figure 5.38 Output voltage and resonant frequency as number of cyclic load

After  $5 \times 10^6$  cycles, the output voltage started to decrease from 280 mV without any shift of resonant frequency. Since the resonant frequency is constant, the change of output voltage is not caused from out of maximum output at the natural frequency of the device but because of the degradation of piezoelectric conversion. The decrease was slightly saturated at 260 mV around  $2 \times 10^7$  cycles. At  $2.3 \times 10^7$  cycles, the PZT was poled again with the same conditions. After the second poling, the output voltage recovered slightly higher than that of first poling. The reason that the output voltage fluctuated is not explained, but the output voltage was saturated at 282 mV. The resonant frequency was still constant. The results show that the PZT had depolarization during the resonating, not generating microstructural defects.

### **5.7.3 Reliability under high temperature**

The performance of the energy harvester was investigated at the high temperature, since it can be exposed to a harsh environment of temperature. The high temperature affects the structural stiffness and material properties of PZT. The Young's modulus, density of materials, and damping ratio have different values at high temperatures, and it is expected that the natural frequency of the device or change in the magnitude of displacement under same acceleration condition. The expansion of the dimensions might be neglected in the influence to the shift of the natural frequency. The dielectric and piezoelectric constants tend to increase and reach their maximum at Curie temperature, since a low energy barrier makes it easy to polarize the materials at a high temperature. The conversion efficiency of PZT film itself should be affected by temperature.

The temperature was set from 25 to 140 °C in a box furnace, and each temperature was stabilized for more than 1 hour. The close circuit output voltages on optimal resistance load were measured depending on the frequency as seen in Figure 5.33.

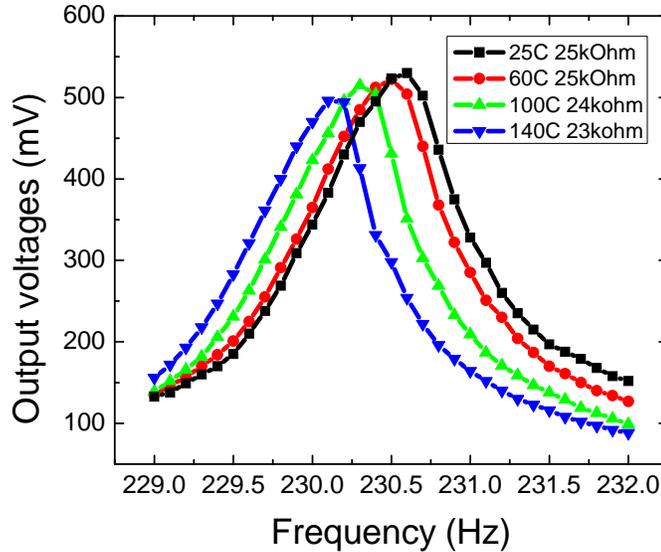


Figure 5.39 Output voltages vs. frequency with different temperature

As the temperature increases, the peak voltages shift to lower values. The gap between the peak frequencies are just 0.1~0.2 Hz, and it means that the piezoelectric MEMS energy harvester has high reliability in matching resonant frequency in high temperature environment.

The peak voltage also decreased with the temperature. It is expected that the displacement of the resonated cantilever is the same as the temperature. The piezoelectric voltage constant decreases as the temperature increases. The piezoelectric voltage constant can be estimated by following Equation (5.11) [90]:

$$g_{31} = \frac{d_{31}}{\epsilon_0 \epsilon^T} \quad (5.11)$$

where  $g_{31}$  is the transverse piezoelectric voltage constant,  $d_{31}$  is the piezoelectric strain constant,  $\epsilon_0$  is the permittivity of vacuum, and  $\epsilon^T$  is the relative dielectric constant at certain temperature.

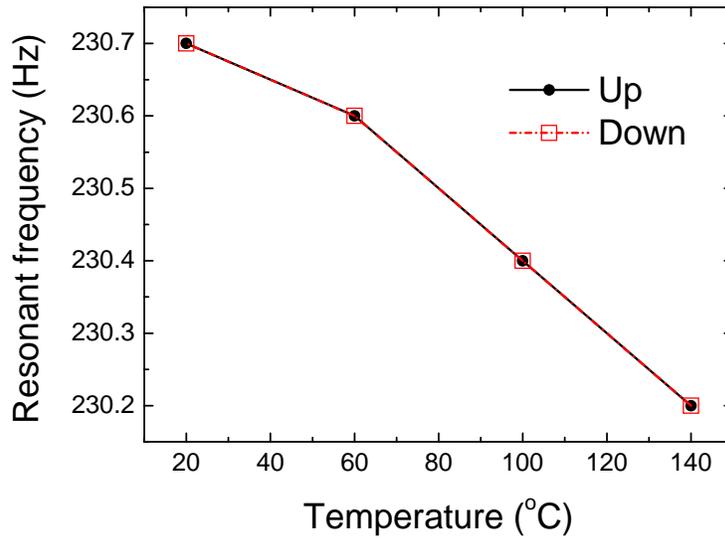


Figure 5.40 Resonant frequency's shift as temperature

The change of resonant frequency was inspected according to temperature in Figure 5.34. When the temperature increased up to 140 °C and then decreased to room temperature, the frequency where the peak voltage existed was measured in both ways. The cantilever has been resonated from start to end of experiments. The shift in resonant frequency caused from the long-term cyclic load was neglected based on the results in the previous section. This experiment was performed to investigate any structural damage of films and their interfaces like dilamination

during exposure to high temperatures. The shift of resonant frequency had good agreement with the results in Figure 5.35, and the resonant frequency was perfectly recovered to that at room temperature after being exposed to high temperature.

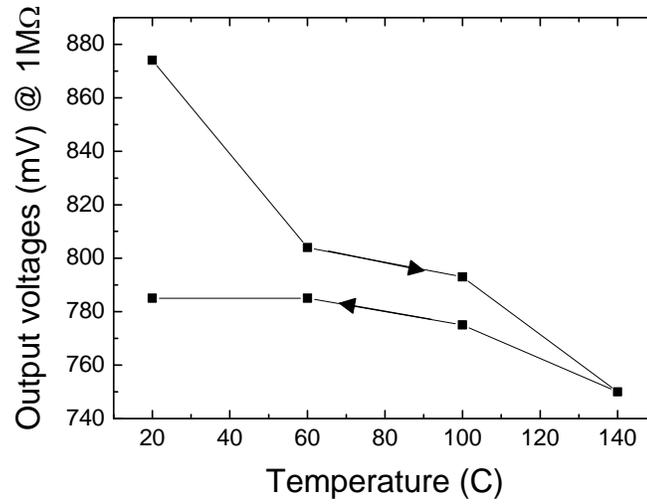


Figure 5.41 The output voltage vs. temperature with open circuit

The output voltage as temperature was inspected, as shown in Figure 5.35. This experiment is performed to investigate the degradation of the piezoelectric properties. The output voltage was measure with 1 MΩ of resistive load during times when the temperature goes up and down. When the temperature increased, the output voltage decreased similar to what was seen in previous results. The linearity between the temperature and output voltage is excluded in this study. On the contrary, the output voltage could not be recovered to the initial value at room temperature during cooling down. It seems that the piezoelectric properties of PZT had degradation during exposure to high temperatures. The maximum temperature of this experiment

is not too high to make a large depolarization. Based on the small decrease of output voltage, a bit of depolarization occurred at that temperature. Followed a poling process, the initial output voltage was recovered.

## CHAPTER 6 CONCLUSIONS AND FUTURE WORK

### 6.1 Summary and conclusions

The present study works for demonstrating and developing the piezoelectric MEMS energy harvester to increase the efficiency of energy conversion. The mechanical vibration energy is converted into electrical energy by successful integration of piezoelectric thin films into microsystems based on silicon. Advance MEMS processes are also developed for robust fabrication. Based on the demonstration of the device, the advanced design is developed to enhance the performance.

Chemical solution deposition with optimal heat treatment was used to deposit PZT thin film, which exhibits the highest piezoelectric coefficient at the composition of morphotropic phase boundary. The substrate of the Pt/adhesive layer largely affected the piezoelectric properties of the PZT thin film. Proper material selection of Ti and thickness control prevented the deterioration of the piezoelectric properties. Further study on thickness of PZT proved that 1  $\mu\text{m}$  is the optimum thickness between the piezoelectric property and squandering deposition process.

The demonstrated single cantilever has dimensions of 4 x 2 x 0.021 mm for the beam and 3 x 2 x 0.5 mm for the silicon proof mass, and the total volume was 3.294mm<sup>3</sup>. Under 0.5 g acceleration and 127 Hz conditions, the device generated 4.52  $\mu\text{W}$ , 538 mVpk-pk, and the

maximum power was  $1372 \mu\text{W}/\text{cm}^3$ . To improve the conversion efficiency of piezoelectric effect, the stress on PZT thin film was widely distributed by using a trapezoidal-shaped cantilever. The trapezoidal shape has increased the power 39% higher than the rectangular shape due to the nonlinear property of piezoelectricity. The local signals near anchor, middle, and mass were first extracted from the MEMS energy harvester, and then they were analyzed comparing simulated data in FEM modeling.

The design of the cantilever array was demonstrated with parallel and serial connection between the cantilevers that have the same dimensions to multiply the current or voltage as the number of cantilevers. The resonant frequency of the energy harvester was very close to the estimation in the range of the ambient vibration range (100 ~ 300 Hz) with less than 2% deviation. Advanced etching techniques and the buried etching stop enabled this fine tuning of resonant frequency, and the cantilever array could multiply the raw signals without any circuitries or loss of electricity. The results show the feasibility that a higher voltage or current can be obtained by integrating multiple cantilevers on one chip.

Reliability is a major issue for most MEMS devices. The structural damage and degradation of piezoelectric properties of PZT under cyclic load and high temperature were investigated. Depolarization seems to begin after  $5 \times 10^6$  cycles, although no microstructural defect existed at that acceleration force. The decrease caused from depolarization was saturated after  $2 \times 10^7$  cycles and was recovered by following the cycles with a poling process. At high temperatures up to  $140^\circ\text{C}$ , the resonant frequency of the device was shifted but fully recovered by cooling down. The output voltage also decreased at high temperatures, but it did not recover to the initial output voltage after cooling down. Depolarization seems to occur, since the initial output voltage was obtained after the poling process that followed. The damage of PZT thin film

after plasma etching was also inspected.

## 6.2 Future work

Piezoelectric MEMS energy harvester still have insufficient energy conversion for the target application of a VLSI wireless sensor or biosensor. Although the conversion efficiency of piezoelectric thin film might be improved, it is restricted in the current thin film technology of metal oxide films. PZT and PZT-based composite materials will be developed to increase the conversion efficiency due to their excellent piezoelectric properties. Furthermore, the advanced design will be developed to enhance the power density of energy harvester. Since the power enhancement though the stress distribution was verified in this study, the future design will be focused on stress distribution under an acceleration force. The quantitative estimation of generated power would be accompanied as the shape change.

The Si-based MEMS energy harvest has insufficient mechanical strength and limitation of design considering various environmental conditions due to the stiffness of Si. On the contrary, the Si-based process is not avoidable considering the cost and heat treatment at high temperature. Although the compound of metal oxide and polymer is actively studied in this field to increase the mechanical toughness, the piezoelectric properties are not sufficient yet. Replacing the silicon membrane with polymer material is proposed for future work. Since the stiffness of polymer materials are relatively easy to control, the multilayer structure of PZT and polymer might enhance both the mechanical strength and displacement under a low acceleration force.

A more efficient mode of a piezoelectric system might be used for the MEMS energy harvester. A higher piezoelectric constant can be obtained in the 33 or 15 modes as compared to the 31 mode. If the configuration of the MEMS structure and electrode design of IDT are considered, it is uncertain that the 33 mode generates a higher power than the 31 mode in spite of its two times higher piezoelectric constant. The 15 mode (with about a four times higher piezoelectric constant than the 31 mode) has not been developed due to its complex design in MEMS structural configuration. The current study for the piezoelectric energy harvest is working on developing the 33 mode, and advanced design for the 15 mode is still a challenge for this device.

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